

## Charge Timer

### Description

The U2403B is a monolithic, integrated-bipolar circuit which can be used in applications for time-controlled, constant-current charge. Selection of charge current versus timing is carried out by using the external circuit at Pins 2, 3 and 4. For high current requirement, an external transistor is recommended in series with the battery. To protect the IC against high power loss

(typically  $> 140^{\circ}\text{C}$ ), the oscillator is shut down when the reference voltage is switched off (0 V). The latter also takes place when there is a saturation caused by collector voltage at Pin 1. When the overtemperature has disappeared and the collector voltage at Pin 1 has exceeded the supply voltage ( $V_1 > V_S$ ), charge time operation continues (see flow chart in figure 3).

### Features

- Easy-to-run autonomous dual rate charger
- Constant charge current
- 3 h – 24 h charge time programmable
- Low cost dc regulator
- Overtemperature protection
- Charge-mode indication
- Operation starts at the moment of battery insertion
- Fast charge time-test mode

### Applications

- Cordless telephones
- Low-cost battery-charge timer
- Entertainment

**Package:** DIP8, SO8

### Block Diagram

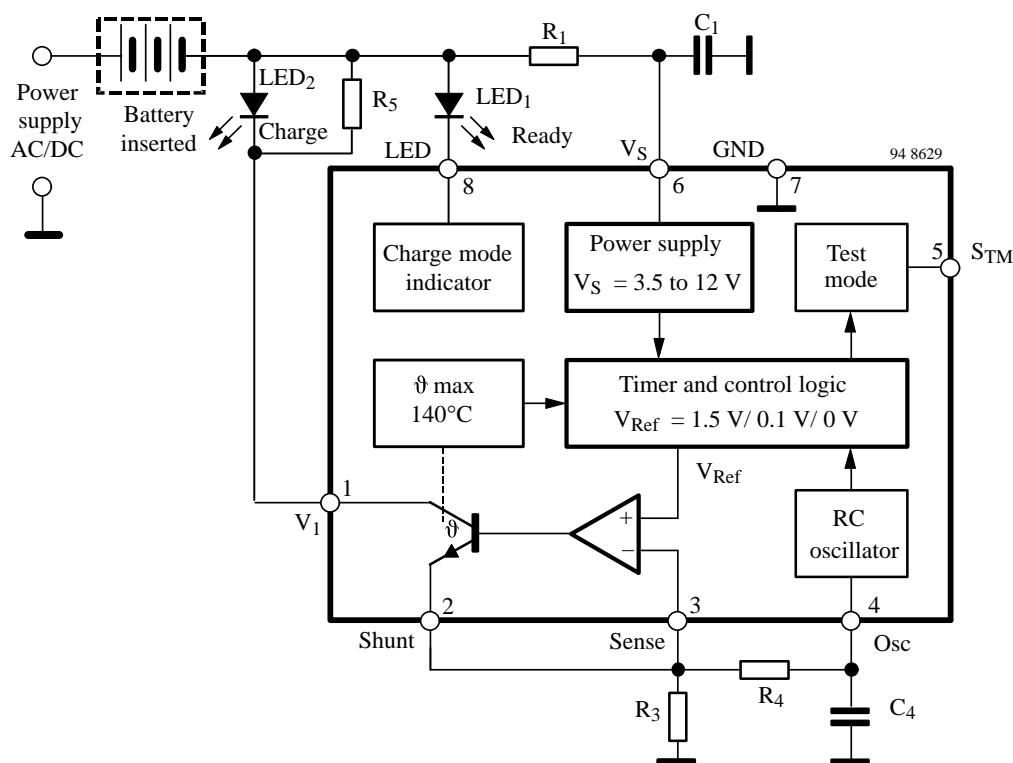
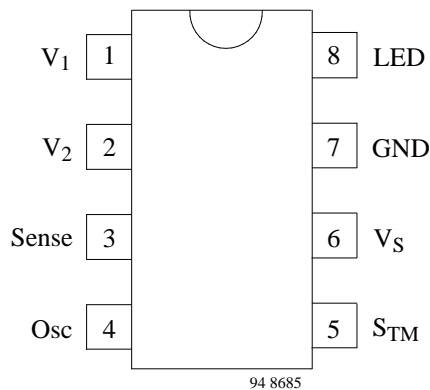


Figure 1. Block diagram with external circuit

## Pin Description



Pin	Symbol	Function
1	V <sub>1</sub>	Collector terminal
2	V <sub>2</sub>	Shunt emitter terminal
3	V <sub>3</sub>	Amplifier sense input
4	Osc	Oscillator input
5	S <sub>TM</sub>	Test mode switch
6	V <sub>S</sub>	Supply voltage
7	GND	Reference point, GND
8	LED	Charge mode indicator

### Pin 1, Collector Voltage V<sub>1</sub>

Pin 1 is an open collector output. When  $V_1 \leq 3$  V, the charge cycle is switched off until it is above the supply voltage, as shown in figure 6.

### Pin 2, Shunt Emitter

The constant current source is supplied by the internal operational amplifier. The voltage across R<sub>3</sub> is determined via the internal reference source.

$$I_{ch} = V_3/R_3 \quad (V_3 = V_{sense})$$

### Pin 3, Amplifier Sense Input (Inverted)

The voltage-regulated current source has a closed loop at Pin 2, Pin 3, and resistor R<sub>3</sub>.

### Pin 4, Oscillator Input R<sub>4</sub>, C<sub>4</sub>

Selection of current charge versus timing is carried out by using the external circuit at Pins 2, 3, and 4. Typical values are given in charge characteristics (see table next page).

### Pin 5, Test-Mode Switch for Charging Time

The charging time,  $t_{ch}$ , is given by the following equation.

$$t_{ch} = \frac{1}{f_{osc}} \times 2^n$$

where:

$f_{osc}$  = oscillator frequency (see figure 2)

n = frequency divider  
 = 26, if S<sub>TM</sub> open  
 = 17, if S<sub>TM</sub> = GND  
 = 8, if S<sub>TM</sub> = V<sub>S</sub>

The first eight divider stages can be tested directly. 256 input tact signals at Pin 4 create one tact signal at Pin 5.

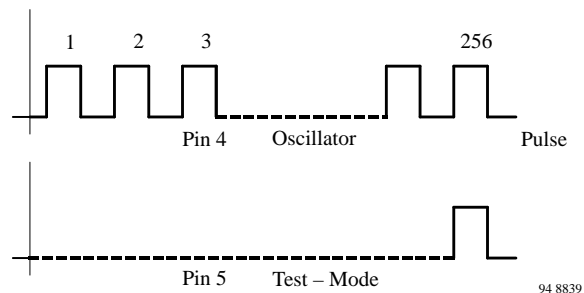


Figure 2. Quick test timer 1/3

### Example

Assume a charge time of 6 h.  
 Select the values of R<sub>4</sub> and C<sub>4</sub> from the tables next page.

For example:  $R_4 = 470 \text{ k}\Omega$   
 $C_4 = 680 \text{ pF}$

There is a frequency of approximately 3100 Hz at Pin 4. It is possible to test the charge time of 6 h by running through the charge cycle for a very short time. By connecting Pin 5 with GND, the test time is 42 s. By connecting Pin 5 with Pin 1 (V<sub>1</sub>), the test time is reduced to about 82.4 ms. R<sub>5</sub> is connected in parallel to the LED<sub>2</sub> and provides a protective bypass function for the LED (see figure 1).

### Pin 6, Supply Voltage, V<sub>S</sub>

$V_S \approx 3.1$  V power-on reset release (turn-on)  
 $V_S \approx 2.9$  V under-voltage reset  
 $V_S \approx 13$  V supply voltage limitation

### Pin 7, Ground

## Pin 8, Charge Mode Indicator

An open-collector output supplies constant current to LED<sub>1</sub> after the active charge phase has been terminated.  $\vartheta_{\max}$  controls the function temperature for the final stage range. This is when the temperature is above 140°C and the charge function is therefore switched off.

## Trickle Charge

The trickle charge starts after the charge has been terminated. In this case, the internal reference voltage is reduced from 1.5 V to approximately 0.1 V. This means the charge current is decreased by the factor:

$$K = 1.5 \text{ V} / 0.1 \text{ V} = 15.$$

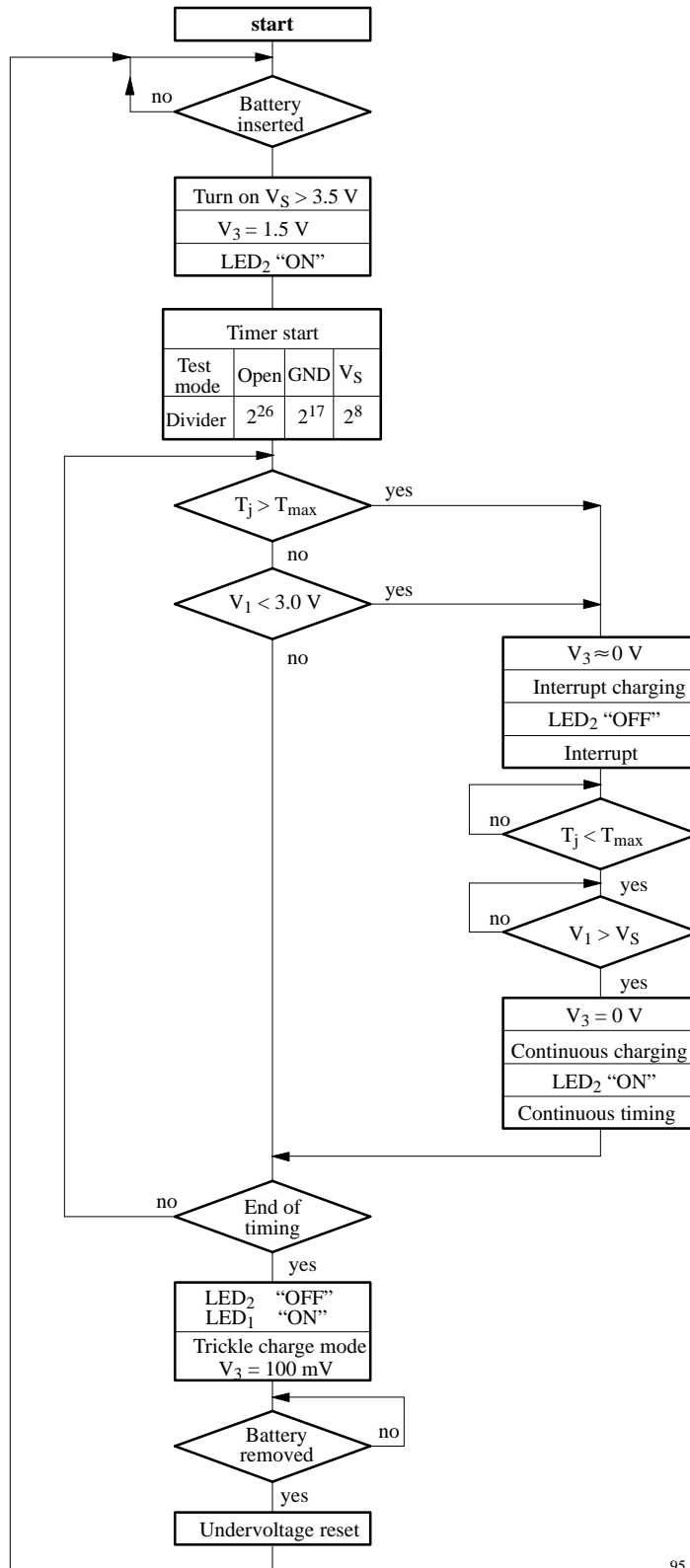
$$\text{Trickle current} = I_{\text{ch}} / 15 + I_6 \text{ (supply current)} + I_8$$

It is possible to reduce the trickle charge with resistor R<sub>6</sub>, as shown in figures 6 and 7.

## Charge Characteristics

### Charge Time

Test time/ Test-Mode Switch S <sub>TM</sub>			Oscillator Components		Frequency
Open	V <sub>S</sub>	GND	R <sub>4</sub> (KΩ)	C <sub>4</sub> (pF)	f <sub>osc</sub> (Hz)
3 h	41.2 ms	21 s	510	270	6213
			430	330	
			300	470	
4 h	54.9 ms	28 s	620	330	4660
			430	470	
			300	680	
5 h	68.6 ms	35 s	510	470	3728
			390	680	
			300	1000	
6 h	82.4 ms	42 s	620	470	3105
			470	680	
			360	1000	
7 h	96.1 ms	49 s	560	680	2663
			430	1000	
			220	2200	
8 h	109.8 ms	56 s	620	680	2330
			470	1000	
			200	2200	
9 h	123.6 ms	1 min 3 s	750	680	2071
			510	1000	
			240	2200	
10 h	137.3 ms	1 min 10 s	620	820	1864
			270	2200	
			130	4700	
12 h	164.8 ms	1 min 24 s	390	2200	1553
			150	4700	
16 h	219.7 ms	1 min 56 s	470	2200	1165
			200	4700	



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Figure 3. Flow chart

## Absolute Maximum Ratings

Reference point Pin 7 (GND), unless otherwise specified.

Parameters	Symbol	Value	Unit
Supply current $t \leq 100 \mu\text{s}$	Pin 6 $I_S$	20	mA
	$i_s$	100	mA
Currents	Pin 1 $I_1$	300	mA
	Pin 2 $-I_2$	310	mA
	Pin 3 $I_3$	1	$\mu\text{A}$
	Pin 4 $I_4$	15	mA
	Pin 5 $I_5$	-75 to +120	$\mu\text{A}$
	Pin 8 $I_8$	8	mA
Voltages	Pins 1, 3, 5, 6 and 8 $V$	13.5	V
	Pin 2 $V_2$	1.6	
	Pin 4 $V_4$	1.5	
Junction temperature	$T_j$	150	$^{\circ}\text{C}$
Ambient temperature	$T_{\text{amb}}$	10 to 85	$^{\circ}\text{C}$
Storage temperature range	$T_{\text{stg}}$	-50 to +150	$^{\circ}\text{C}$

## Thermal Resistance

Parameters	Symbol	Value	Unit
Junction ambient	$R_{\text{thJA}}$	DIP8	120
		SO8 on PC-board	220
		SO8 on ceramic	140
		SO8 on ceramic with thermal compound	80

## Electrical Characteristics

$V_S = 6\text{ V}$ ,  $T_{\text{amb}} = 25^\circ\text{C}$ , reference point Pin 7 (GND), unless otherwise specified.

Parameters	Test Conditions / Pins	Symbol	Min.	Typ.	Max.	Unit
Supply voltage limitation	Pin 6 $I_S = 4\text{ mA}$ $I_S = 20\text{ mA}$	$V_S$	12.5		13.5	V
			12.6		13.7	
Supply current	$V_S = 6\text{ V}$	$I_S$	1.4		2.2	mA
<b>Voltage monitoring</b> Pin 6						
Turn-on threshold		$V_{\text{TON}}$	2.8		3.5	V
Turn-off threshold		$V_{\text{TOFF}}$	2.5		3.2	
<b>Charge-mode indicator (LED)</b> Pin 8						
LED current		$I_8$	3.0		6.0	mA
LED saturation voltage	$I_8 = 3.7\text{ mA}$	$V_8$			960	mV
Leakage current		$I_{\text{lk}}$	-0.35		1.1	$\mu\text{A}$
<b>Collector terminal, Figure 5</b> Pin 1						
Open collector current		$I_{\text{CO}}$	15		55	$\mu\text{A}$
Saturation threshold	$V_S = 6\text{ V}$	$V_{\text{TON}}$	2.55	3.0	3.35	V
		$V_{\text{TOFF}}$	$V_S - 1\text{V}$	$V_S$	$V_S - 0.4\text{V}$	
Shunt emitter current	$R_3 = 5.6\ \Omega$ Pin 2	$I_2$	250		285	mA
<b>Operational sense amplifier, Figure 1</b> Pin 3						
Input current	$V_3 = 0\text{ V}$	$I_3$	-0.6		0.08	$\mu\text{A}$
Input voltage	$V_{\text{Ref}} = 1.5\text{ V}$ $V_{\text{Ref}} = 100\text{ mV}$ $V_{\text{Ref}} = 0\text{ V}$	$V_3$	1.42	1.5	1.58	V
			40	70	100	mV
			-0.4		40	mV
<b>Oscillator</b> Pin 4						
Leakage current	$V_4 = 0\text{ to }0.85\text{ V}$	$I_{\text{lk}}$	-0.5		0.1	$\mu\text{A}$
Threshold voltage	Upper	$V_{\text{T(u)}}$	875		985	mV
Oscillator frequency	$R_4 = 160\text{ k}\Omega$ , $C_4 = 2.2\text{ nF}$ $R_4 = 680\text{ k}\Omega$ , $C_4 = 4.7\text{ nF}$	$f_{\text{osc}}$	2700		3050	Hz
			305		345	
<b>Test mode switch (STM)</b> Pin 5						
Input current	$V_5 = 6\text{ V}$	$I_5$	40		120	$\mu\text{A}$
	$V_5 = 0\text{ V}$		-75		-20	
Output voltage	High	$V_{0(\text{H})}$	1.7		2.5	V
	Low	$V_{0(\text{L})}$	0.5		1.0	

**Internal Temperature Switch**

The internal temperature monitoring is active if the chip temperature rises above 140°C. Above this temperature the voltage at Pin 3 goes to zero. Similarly, the charge current,  $I_{ch}$ , reduces according to the equation:

$$I_{ch} = V_3 / R_3$$

where  $I_{ch} = 1$  to 2 mA (IC supply current)

The oscillator is connected to GND via Pin 3 ( $V_3$ ) which holds the present time status. When the chip temperature decreases below the transition value, all functions are released and the charge time is continued. The process is reversible. If there is a higher power dissipation in the circuit ( $T_j > 140^\circ\text{C}$ ), the temperature monitoring remains permanently activated (ON). The total cycle time is prolonged according to the interrupt-time duration, see figure 4.

**Automatic Control Protection**

To reduce the design costs, it is possible to select the transformer which requires minimum power supply.

The output stage of the control is selected so that it is switched off before saturation is achieved ( $V_{CEsat} = 3.0\text{ V}$ ). In this case, the voltage at Pin 3 is kept at a value of zero. The charge current is also zero and the transformer is now an open circuit impedance. The system becomes active again if  $V_1 \geq V_S$ .

The advantage of the system is that if sags of short duration appear on the mains voltage or if the transformers used are too small, the charge duration is increased, but the charge capacity remains the same (see figure 5).

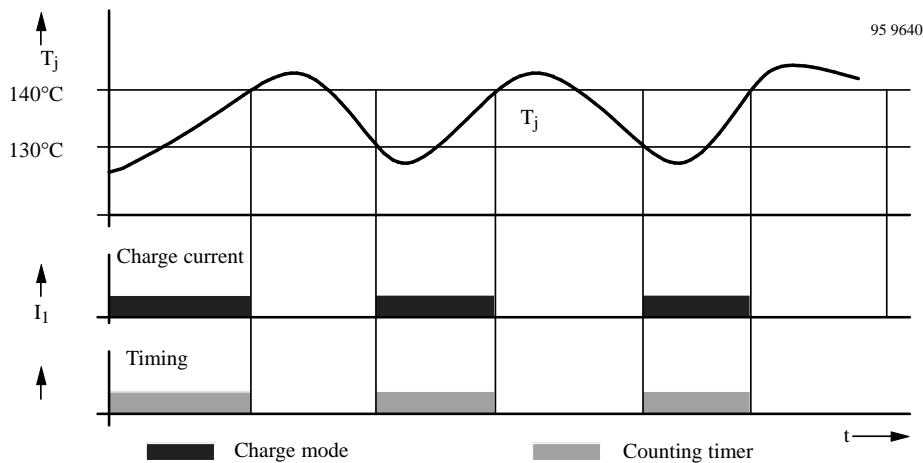


Figure 4. Charge duration – overtemperature

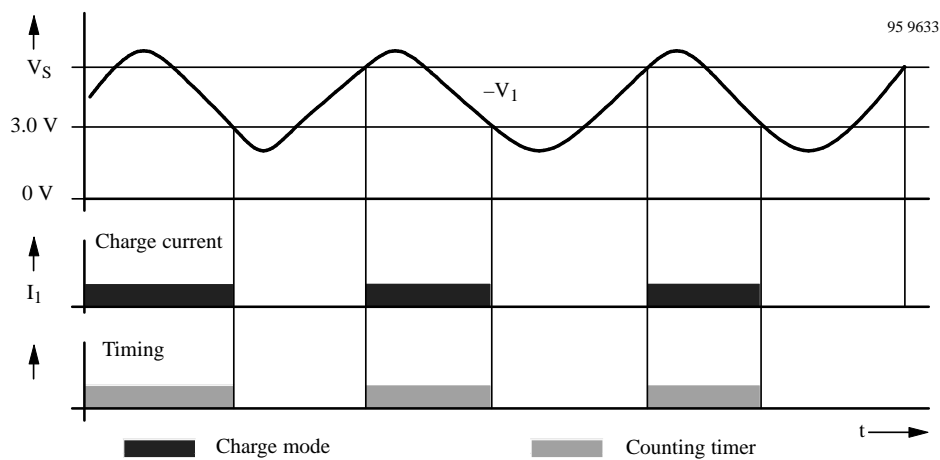


Figure 5. Charge duration –  $V_1$

## Standard Applications

### Basic Example

NiCd battery 750 mAh	$R_1 = 510 \Omega$ , 1/8 W
Charging time: 3 h	$C_1 = 47 \mu\text{F}$ / 16 V
Charge current: 240 mA, 1/3 C	$R_3 = 6.2 \Omega$ , 1/2 W
Trickle charge: 19 mA < 1/40 C	$R_4 = 300 \text{ k}\Omega$
	$C_4 = 470 \text{ pF}$
	$R_5 = 8.2 \Omega$ , 1/2 W

### Minimum Supply Voltage

No of Cells	DC Supply Minimum
1	6.8 V
2	8.3 V
3	9.8 V
4	11.3 V
5	12.8 V

### Special Requirements of Different Charge Times

$R_4$ ,  $C_4$  values for different charging times

	2 h	4 h	6 h	7 h	12 h
$R_4$	300 k $\Omega$	430 k $\Omega$	470 k $\Omega$	470 k $\Omega$	390 k $\Omega$
$C_4$	330 pF	470 pF	680 pF	1 nF	2.2 nF

### Special Requirements for Different Charge Current

$R_3$ ,  $R_5$  values for different charge current

	240 mA	150 mA	100 mA	50 mA
$R_3$	6.2 $\Omega$	10 $\Omega$	15 $\Omega$	30 $\Omega$
$R_5$	8.2 $\Omega$	15 $\Omega$	22 $\Omega$	68 $\Omega$

## Basic Equations

$$R_1 = 0.5 \text{ V} / I_S$$

$$I_S = 1.8 \text{ mA}$$

$$R_5 = V_5 / (I_{\text{ch}} - 20 \text{ mA})$$

### Nominal Charge Current:

$$I_{\text{ch}} = V_3 / R_3 \text{ where } V_3 = 1.48 \text{ V (typ.)}$$

### Trickle Current:

$$I_{\text{ch}} = V_3 / R_3 + I_8 + I_S$$

Typical values are:

$$V_3 = 100 \text{ mV}, I_8 = 4.5 \text{ mA}$$

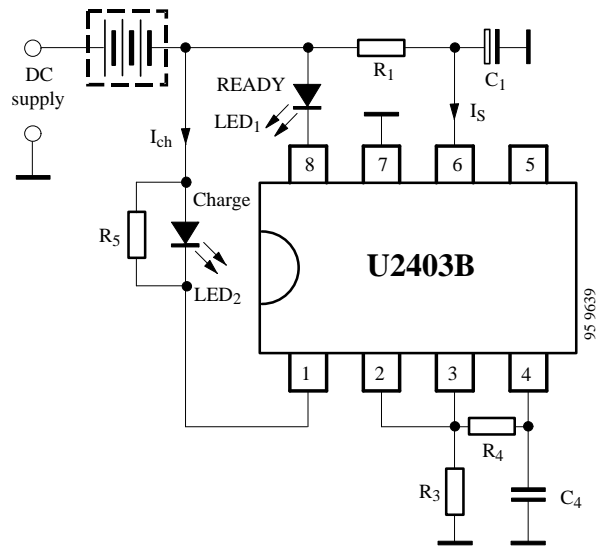


Figure 6. Standard application



## Booster and Trickle Charge Reduction

### Basic Example

NiCd battery 1000 mAh	$R_1 = 510 \Omega$ , 1/8 W
Charging time: 2 h	$C_1 = 100 \mu\text{F}$ / 16 V
Charge current: 500 mA	$R_3 = 3 \Omega$ / 1 W
Trickle charge: 22 mA < 1/22 C	$R_4 = 300 \text{ k}\Omega$ $C_4 = 330 \text{ pF}$ $R_5 = 3.9 \Omega$ / 1 W $C_2 = 1 \mu\text{F}$

### Supply Voltage

No of Cells	DC Supply Minimum
1	$V_S = 6.5 \text{ V}$
2	8.0 V
3	9.5 V
4	11.0 V
5	12.5 V

### Special Requirements for Different Charge Times

$R_4$ ,  $C_4$  values for different charge times

	2 h	4 h	6 h	7 h	12 h
$R_4$	300 k $\Omega$	430 k $\Omega$	470 k $\Omega$	470 k $\Omega$	390 k $\Omega$
$C_4$	330 pF	470 pF	680 pF	1 nF	2.2 nF

### Special Requirements for Different Charge Current

$R_3$ ,  $R_5$  values for different charge currents

	616 mA	493 mA	411 mA	296 mA
$R_3$	2.4 $\Omega$	3 $\Omega$	3.6 $\Omega$	5 $\Omega$
$R_5$	3 $\Omega$	3.9 $\Omega$	4.7 $\Omega$	6.8 $\Omega$

$R_6 = 560 \Omega$ , reduced trickle charge

### Basic Equations

$$R_1 = 0.5 \text{ V} / I_S$$

$$R_5 = V(\text{LED}_2) / (I_{\text{ch}} - 20 \text{ mA})$$

### Nominal Charge Current:

$$I_{\text{ch}} = V_3 / R_3$$

$V_3 = 1.48 \text{ V}$ , typically

### Trickle Current:

$$I_{\text{ch}} = V_3 / R_3 + I_{\text{LED1}} + I_S - I_6$$

Typical values:  
 $V_3 = 100 \text{ mV}$   
 $I_{\text{LED1}} = 4.5 \text{ mA}$   
 $I_S = 1.8 \text{ mA}$

### Trickle-Charge Reduction ( $I_6$ )

$$I_6 = (V_{\text{Batt}} + V_{\text{D1}}) / R_6 \quad V_{\text{D1}} = 0.75 \text{ V}$$

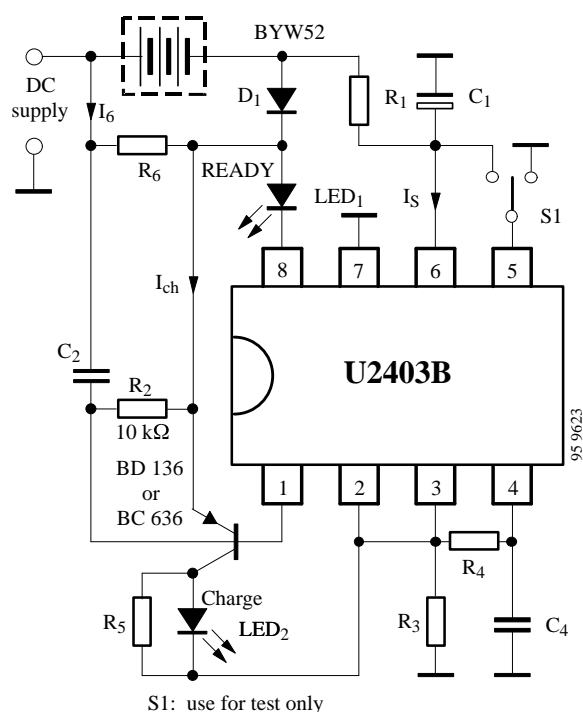


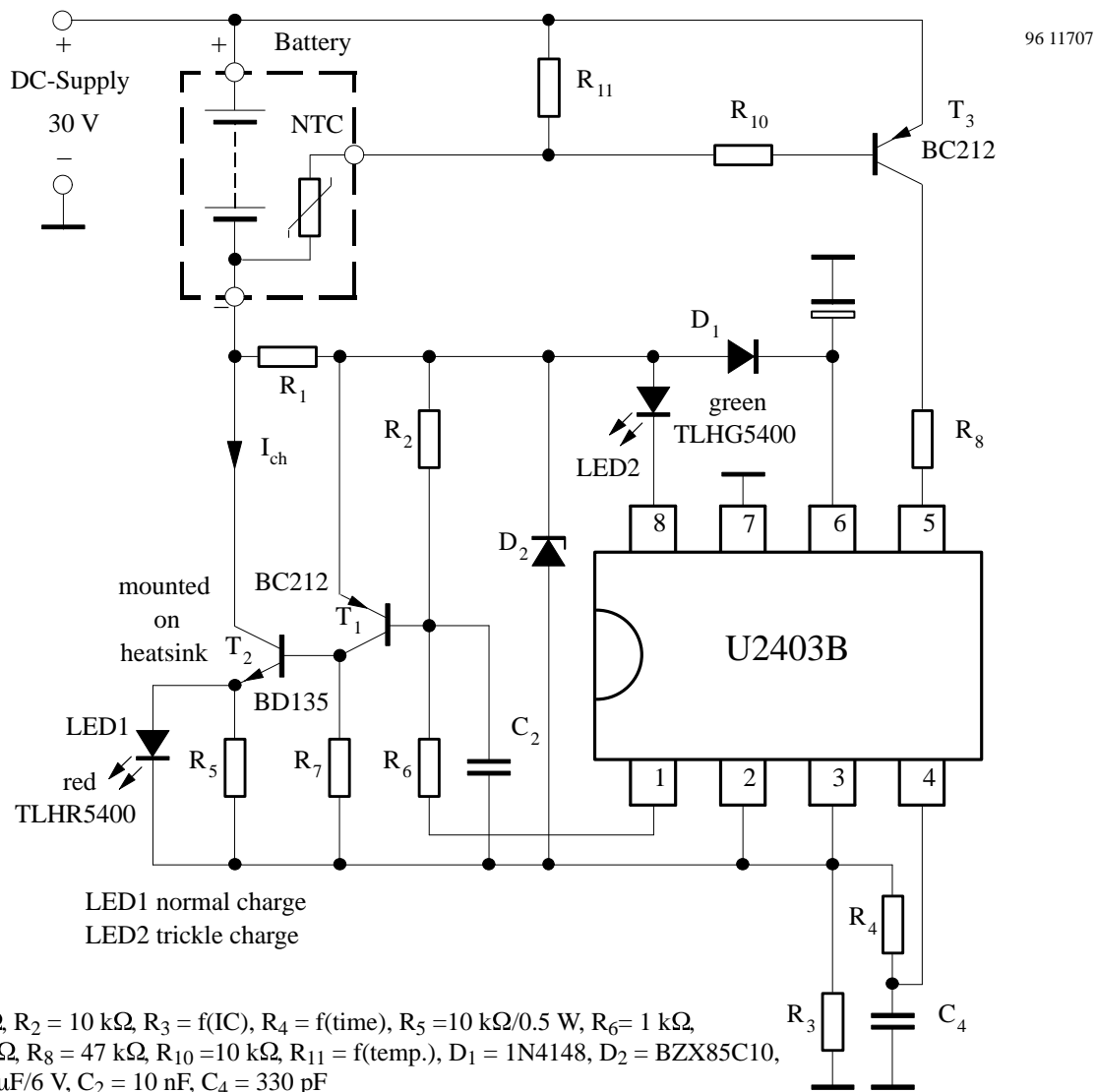
Figure 7. Application for charge current > 250 mA

To fulfill requirements of higher charge current an external booster transistor can be used (see figure 7). As the temperature cannot be monitored in this case a heat sink with a reasonable size should be used for safe operation. Test mode switch S1 can be used for accelerated production check.

## Charge System at Higher Voltage of 30 V

Charge systems with higher voltages than  $V_{Smax}$  can be realized with the additional expander circuitry, as shown in figure 8. This circuit contains a simple temperature monitoring function. When the temperature level is

reached, the transistor,  $T_3$ , is switched on. If  $T_3$  is switched on and there is current flow into Pin 5, normal charge is terminated.



$R_1 = 1\text{ k}\Omega$ ,  $R_2 = 10\text{ k}\Omega$ ,  $R_3 = f(IC)$ ,  $R_4 = f(\text{time})$ ,  $R_5 = 10\text{ k}\Omega/0.5\text{ W}$ ,  $R_6 = 1\text{ k}\Omega$ ,  
 $R_7 = 10\text{ k}\Omega$ ,  $R_8 = 47\text{ k}\Omega$ ,  $R_{10} = 10\text{ k}\Omega$ ,  $R_{11} = f(\text{temp.})$ ,  $D_1 = 1N4148$ ,  $D_2 = BZX85C10$ ,  
 $C_1 = 100\text{ }\mu\text{F}/6\text{ V}$ ,  $C_2 = 10\text{ nF}$ ,  $C_4 = 330\text{ pF}$   
 $R_{11} = f(\text{temp.})$  depends on number of cells

Figure 8. U2403B for higher supply voltage up to 30 V with integrated temperature monitoring

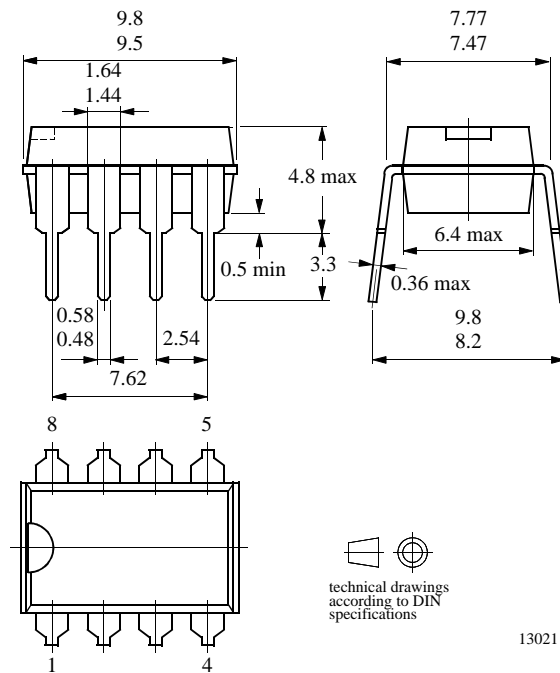
No of Cells	$R_{11}$
2	13 k $\Omega$
3	8.2 k $\Omega$
4	6.2 k $\Omega$
5	4.7 k $\Omega$

NTC Value	
25°C	6.8 k $\Omega$
40°C	3.9 k $\Omega$
50°C	2.8 k $\Omega$

**Package Information**

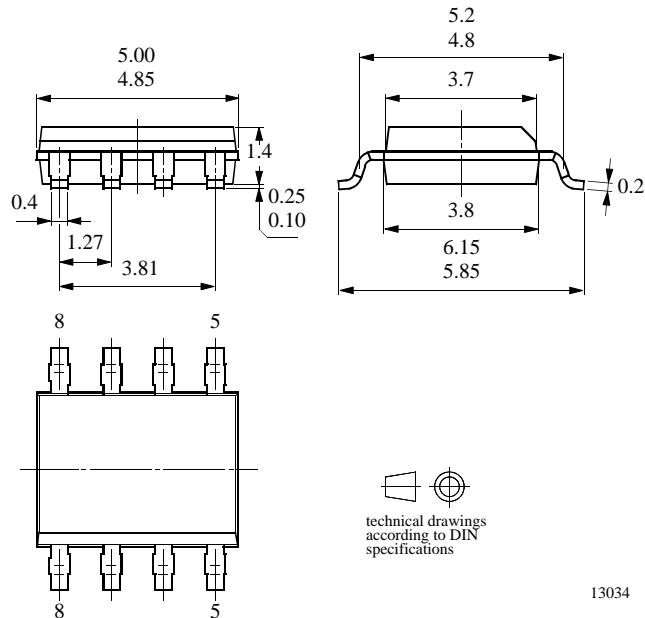
Package DIP8

Dimensions in mm



Package SO8

Dimensions in mm



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