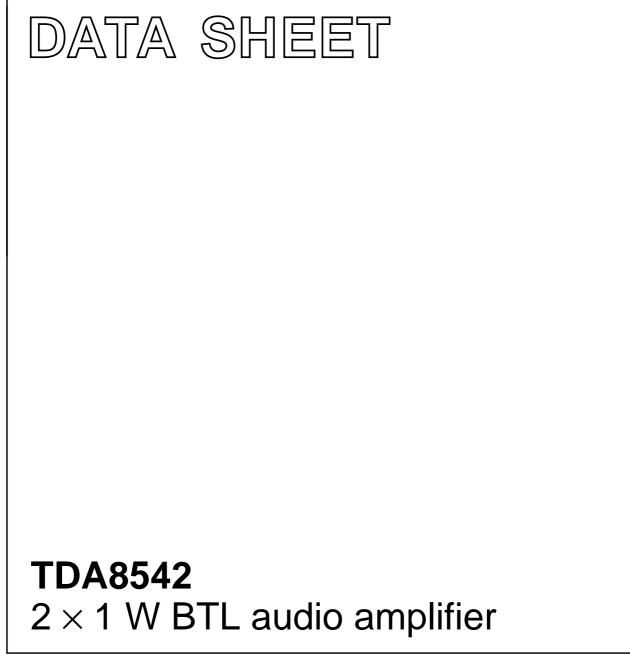
INTEGRATED CIRCUITS



Product specification Supersedes data of 1997 Feb 19 File under Integrated Circuits, IC01 1998 Apr 01



TDA8542

FEATURES

- Flexibility in use
- Few external components
- Low saturation voltage of output stage
- Gain can be fixed with external resistors
- Standby mode controlled by CMOS compatible levels
- Low standby current
- No switch-on/switch-off plops
- High supply voltage ripple rejection
- Protected against electrostatic discharge
- Outputs short-circuit safe to ground, V_{CC} and across the load
- Thermally protected.

QUICK REFERENCE DATA

APPLICATIONS

- Portable consumer products
- Personal computers
- Motor-driver (servo).

GENERAL DESCRIPTION

The TDA8542(T) is a two channel audio power amplifier for an output power of 2×1 W with an 8 Ω load at a 5 V supply. The circuit contains two BTL amplifiers with a complementary PNP-NPN output stage and standby/mute logic. The TDA8542T comes in a 16 pin SO package and the TDA8542 in a 16 pin DIP package.

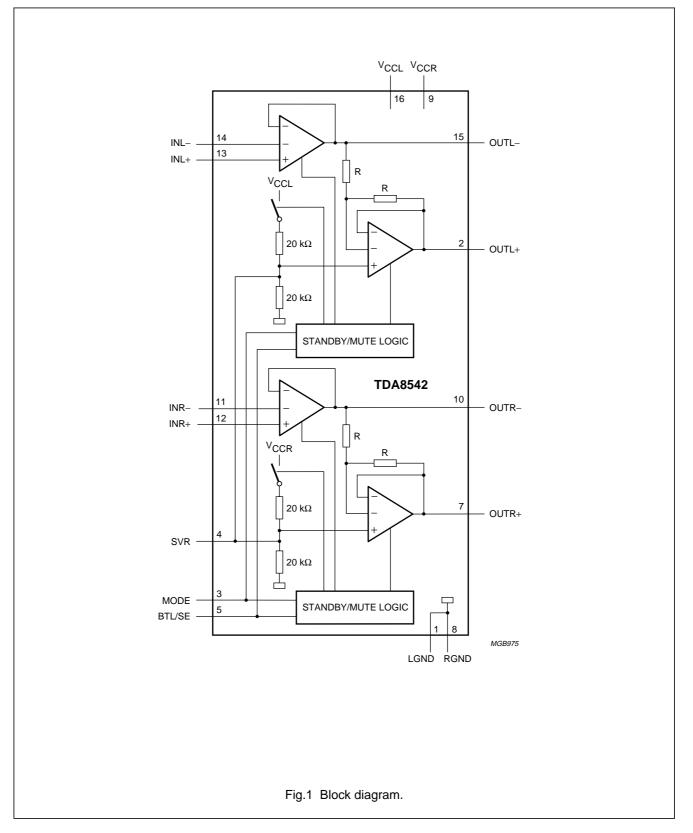
SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V _{CC}	supply voltage		2.2	5	18	V
lq	quiescent current	$V_{CC} = 5 V$	-	15	22	mA
I _{stb}	standby current		—	—	10	μA
Po	output power	THD = 10%; $R_L = 8 \Omega$; $V_{CC} = 5 V$	1	1.2	-	W
THD	total harmonic distortion	$P_{o} = 0.5 W$	-	0.15	-	%
SVRR	supply voltage ripple rejection		50	-	_	dB

ORDERING INFORMATION

TYPE	PACKAGE		
NUMBER	NAME DESCRIPTION VERS		VERSION
TDA8542T	SO16L	D16L plastic small outline package; 16 leads; body width 7.5 mm	
TDA8542	DIP16	plastic dual in-line package; 16 leads (300 mil); long body	SOT38-1

TDA8542

BLOCK DIAGRAM

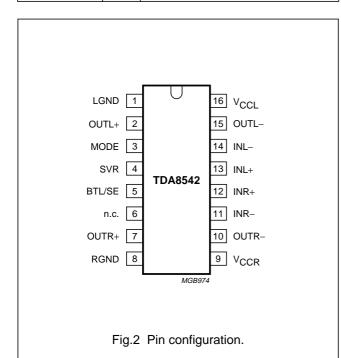


TDA8542

2×1 W BTL audio amplifier

PINNING

SYMBOL	PIN	DESCRIPTION
LGND	1	ground, left channel
OUTL+	2	positive loudspeaker terminal, left channel
MODE	3	operating mode select (standby, mute, operating)
SVR	4	half supply voltage, decoupling ripple rejection
BTL/SE	5	BTL loudspeaker or SE headphone operation
n.c.	6	not connected
OUTR+	7	positive loudspeaker terminal, right channel
RGND	8	ground, right channel
V _{CCR}	9	supply voltage, right channel
OUTR-	10	negative loudspeaker terminal, right channel
INR-	11	negative input, right channel
INR+	12	positive input, right channel
INL+	13	positive input, left channel
INL-	14	negative input, left channel
OUTL-	15	negative loudspeaker terminal, left channel
V _{CCL}	16	supply voltage, left channel



FUNCTIONAL DESCRIPTION

The TDA8542(T) is a 2 × 1 W BTL audio power amplifier capable of delivering 2 × 1 W output power to an 8 Ω load at THD = 10% using a 5 V power supply. Using the MODE pin the device can be switched to standby and mute condition. The device is protected by an internal thermal shutdown protection mechanism. The gain can be set within a range from 6 dB to 30 dB by external feedback resistors.

Power amplifier

The power amplifier is a Bridge Tied Load (BTL) amplifier with a complementary PNP-NPN output stage. The voltage loss on the positive supply line is the saturation voltage of a PNP power transistor, on the negative side the saturation voltage of a NPN power transistor. The total voltage loss is <1 V and with a 5 V supply voltage and an 8 Ω loudspeaker an output power of 1 W can be delivered.

Mode select pin

The device is in the standby mode (with a very low current consumption) if the voltage at the MODE pin is $>(V_{CC} - 0.5 \text{ V})$, or if this pin is floating. At a MODE voltage level of less than 0.5 V the amplifier is fully operational. In the range between 1.5 V and $V_{CC} - 1.5 \text{ V}$ the amplifier is in mute condition. The mute condition is useful to suppress plop noise at the output caused by charging of the input capacitor.

Headphone connection

A headphone can be connected to the amplifier using two coupling capacitors for each channel. The common GND pin of the headphone is connected to the ground of the amplifier (see Fig.13). In this case the BTL/SE pin must be either on a logic HIGH level or not connected at all.

The two coupling capacitors can be omitted if it is allowed to connect the common GND pin of the headphone jack not to ground, but to a voltage level of $\frac{1}{2}V_{CC}$ (see Fig.13). In this case the BTL/SE pin must be either on a logic LOW level or connected to ground. If the BTL/SE pin is on a LOW level, the power amplifier for the positive loudspeaker terminal is always in mute condition.

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LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V _{CC}	supply voltage	operating	-0.3	+18	V
VI	input voltage		-0.3	V _{CC} + 0.3	V
I _{ORM}	repetitive peak output current		-	1	А
T _{stg}	storage temperature	non-operating	-55	+150	°C
T _{amb}	operating ambient temperature		-40	+85	°C
V _{psc}	AC and DC short-circuit safe voltage		-	10	V
P _{tot}	total power dissipation	SO16L	_	1.2	W
		DIP16	_	2.2	W

QUALITY SPECIFICATION

In accordance with "SNW-FQ-611-E". The number of the quality specification can be found in the "Quality Reference Handbook". The handbook can be ordered using the code 9397 750 00192.

THERMAL CHARACTERISTICS

SYMBOL	PARAMETER		UNIT
R _{th j-a}	thermal resistance from junction to ambient in free air:		
	TDA8542T (SO16L)		K/W
	TDA8542 (DIP16)	55	K/W

TDA8542

DC CHARACTERISTICS

 V_{CC} = 5 V; T_{amb} = 25 °C; R_L = 8 Ω ; V_{MODE} = 0 V; measured in test circuit Fig.3; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V _{CC}	supply voltage	operating	2.2	5	18	V
lq	quiescent current	R _L = ∞; note 1	_	15	22	mA
I _{stb}	standby current	$V_{MODE} = V_{CC}$	-	-	10	μA
Vo	DC output voltage	note 2	_	2.2	-	V
V _{OUT+} – V _{OUT-}	differential output voltage offset		_	_	50	mV
I _{IN+} , I _{IN-}	input bias current		_	_	500	nA
V _{MODE}	input voltage mode select	operating	0	_	0.5	V
		mute	1.5	_	V _{CC} – 1.5	V
		standby	V _{CC} - 0.5	_	V _{CC}	V
I _{MODE}	input current mode select	$0 < V_{MODE} < V_{CC}$	_	_	20	μA
V _{BS}	input voltage BTL/SE pin	single-ended	0	_	0.6	V
		BTL	2	_	V _{CC}	V
I _{BS}	input current BTL/SE pin	$V_{BS} = 0$	_	_	100	μA

Notes

- 1. With a load connected at the outputs the quiescent current will increase, the maximum of this increase being equal to the DC output offset voltage divided by R_L .
- 2. The DC output voltage with respect to ground is approximately $0.5 \times V_{CC}.$

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AC CHARACTERISTICS

V_{CC} = 5 V; T_{amb} = 25 °C; R_L = 8 Ω; f = 1 kHz; V_{MODE} = 0 V; measured in test circuit Fig.3; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Po	output power	THD = 10%	1	1.2	-	W
		THD = 0.5%	0.6	0.9	-	W
THD	total harmonic distortion	P _o = 0.5 W	-	0.15	0.3	%
Gv	closed loop voltage gain	note 1	6	-	30	dB
Zi	differential input impedance		-	100	-	kΩ
V _{no}	noise output voltage	note 2	-	-	100	μV
SVRR	supply voltage ripple rejection	note 3	50	-	-	dB
		note 4	40	-	-	dB
Vo	output voltage in mute condition	note 5	-	-	200	μV
α_{cs}	channel separation		40	_	_	dB

Notes

- 1. Gain of the amplifier is $2 \times R2/R1$ in test circuit of Fig.3.
- 2. The noise output voltage is measured at the output in a frequency range from 20 Hz to 20 kHz (unweighted), with a source impedance of $R_S = 0 \Omega$ at the input.
- 3. Supply voltage ripple rejection is measured at the output, with a source impedance of $R_S = 0 \Omega$ at the input. The ripple voltage is a sine wave with a frequency of 1 kHz and an amplitude of 100 mV (RMS), which is applied to the positive supply rail.
- 4. Supply voltage ripple rejection is measured at the output, with a source impedance of $R_S = 0 \Omega$ at the input. The ripple voltage is a sine wave with a frequency between 100 Hz and 20 kHz and an amplitude of 100 mV (RMS), which is applied to the positive supply rail.
- 5. Output voltage in mute position is measured with a 1 V (RMS) input voltage in a bandwidth of 20 kHz, so including noise.

TDA8542

TEST AND APPLICATION INFORMATION

Test conditions

Because the application can be either Bridge-Tied Load (BTL) or Single-Ended (SE), the curves of each application are shown separately.

The thermal resistance = 55 K/W for the DIP16; the maximum sine wave power dissipation for T_{amb} = 25 °C is:

$$\frac{150-25}{55} = 2.3 \text{ W}$$

For $T_{amb} = 60 \text{ °C}$ the maximum total power dissipation is:

 $\frac{150-60}{55} = 1.7 \text{ W}$

BTL application

 $T_{amb} = 25^{\circ}C$ if not specially mentioned, $V_{CC} = 5 V$, f = 1 kHz, $R_L = 8 \Omega$, $G_v = 20 dB$, audio band-pass 22 Hz to 22 kHz.

The BTL application diagram is illustrated in Fig.3.

The quiescent current has been measured without any load impedance. The total harmonic distortion as a function of frequency was measured with a low-pass filter of 80 kHz. The value of capacitor C3 influences the behaviour of the SVRR at low frequencies, increasing the value of C3 increases the performance of the SVRR. The figure of the mode select voltage (V_{ms}) as a function of the supply voltage shows three areas; operating, mute and standby. It shows, that the DC-switching levels of the mute and standby respectively depends on the supply voltage level.

SE application

 $\label{eq:Tamb} \begin{array}{l} \mathsf{T}_{amb} = 25^\circ\text{C} \text{ if not specially mentioned, } \mathsf{V}_{CC} = 7.5 \text{ V}, \\ \mathsf{f} = 1 \text{ kHz}, \ \mathsf{R}_{\mathsf{L}} = 4 \ \Omega, \ \mathsf{G}_{\mathsf{v}} = 20 \text{ dB}, \ \text{audio band-pass} \\ 22 \text{ Hz to } 22 \text{ kHz}. \end{array}$

The SE application diagram is illustrated in Fig.14.

If the BTL/SE pin (pin 5) is connected to ground, the positive outputs (pins 2 and 7) will be in mute condition with a DC level of $1_2'V_{CC}$. When a headphone is used ($R_L \geq 25~\Omega$) the SE headphone application can be used without output coupling capacitors; load between negative output and one of the positive outputs (e.g. pin 2) as common pin.

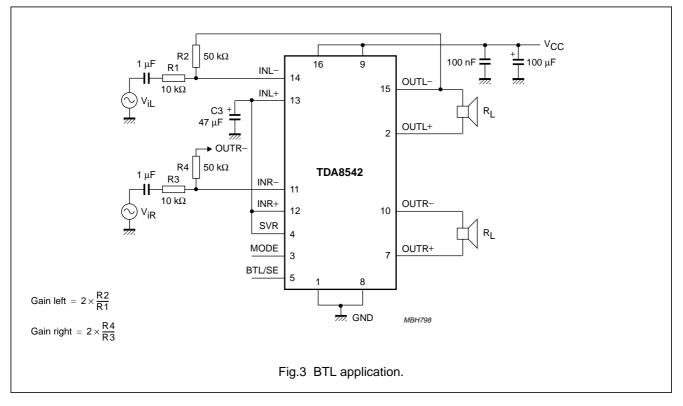
Increasing the value of electrolytic capacitor C3 will result in a better channel separation. Because the positive output is not designed for high output current ($2 \times I_o$) at low load impedance ($\leq 16 \Omega$), the SE application with output capacitors connected to ground is advised. The capacitor value of C4/C5 in combination with the load impedance determines the low frequency behaviour. The THD as a function of frequency was measured using a low-pass filter of 80 kHz. The value of capacitor C3 influences the behaviour of the SVRR at low frequencies, increasing the value of C3 increases the performance of the SVRR.

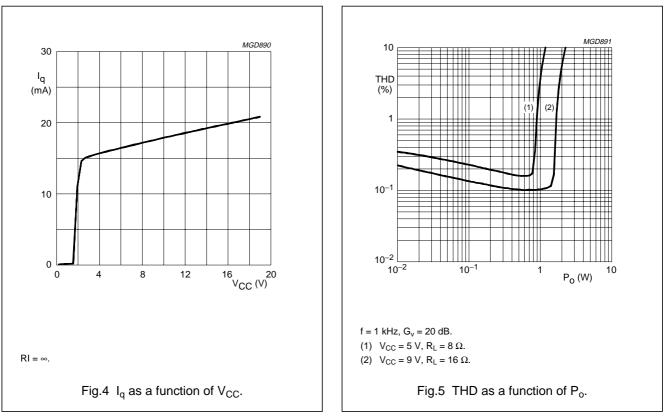
General remark

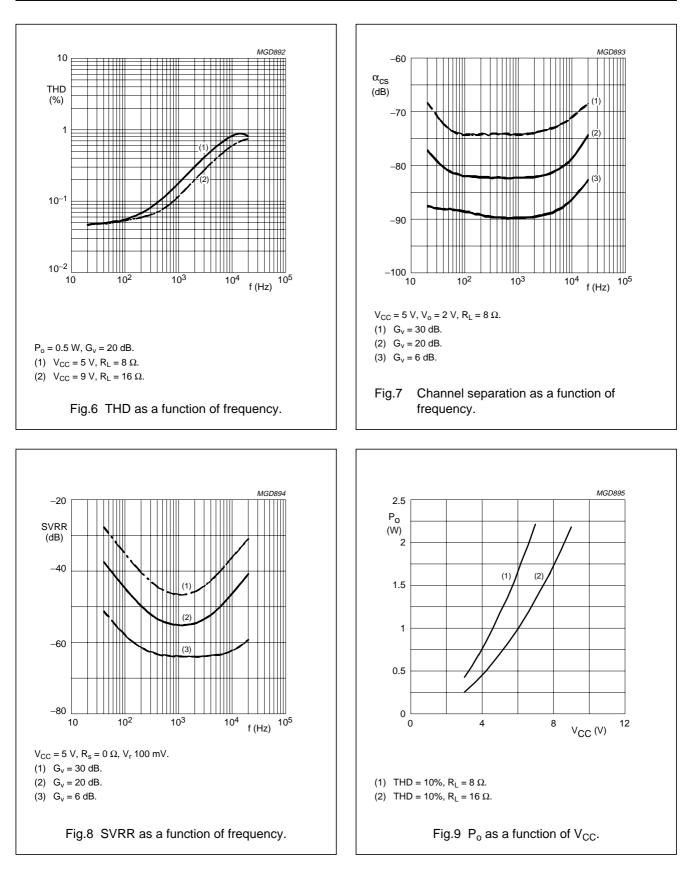
The frequency characteristic can be adapted by connecting a small capacitor across the feedback resistor. To improve the immunity of HF radiation in radio circuit applications, a small capacitor can be connected in parallel with the feedback resistor (56 k Ω); this creates a low-pass filter.

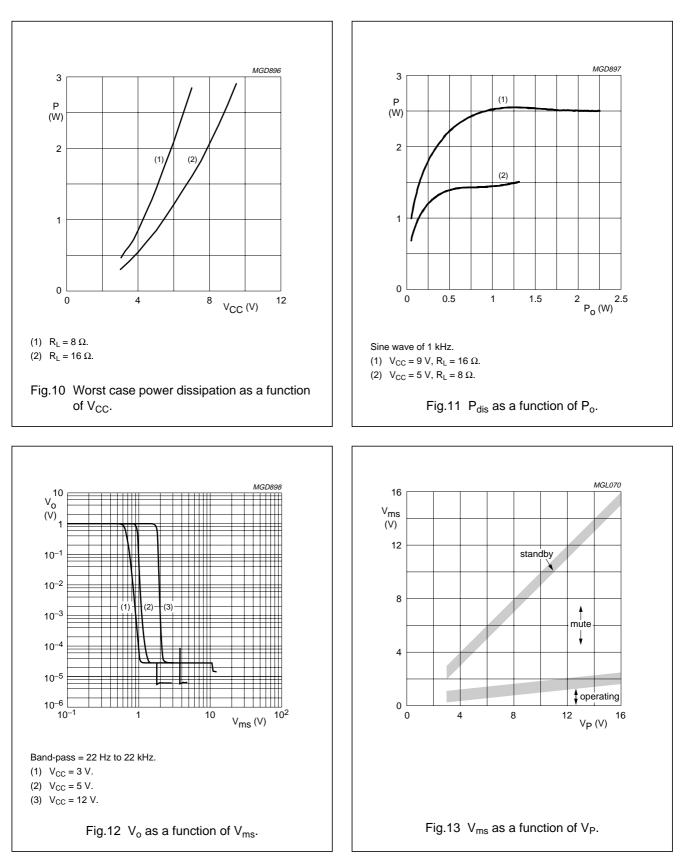
TDA8542

BTL APPLICATION



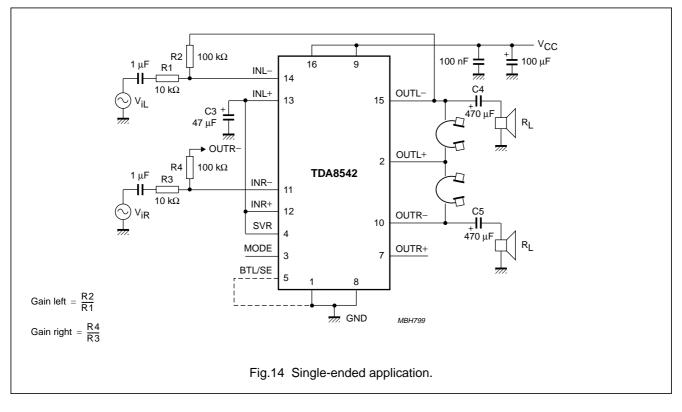


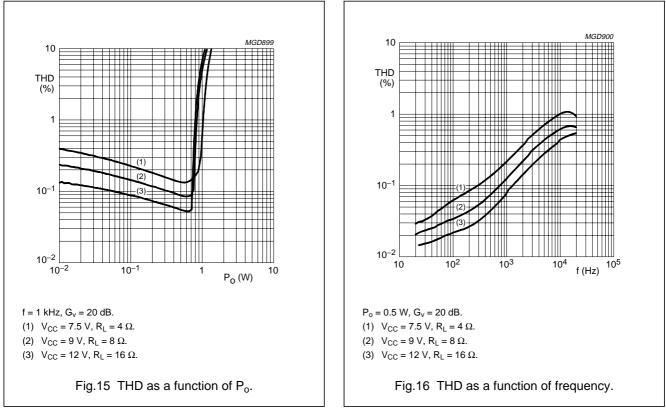


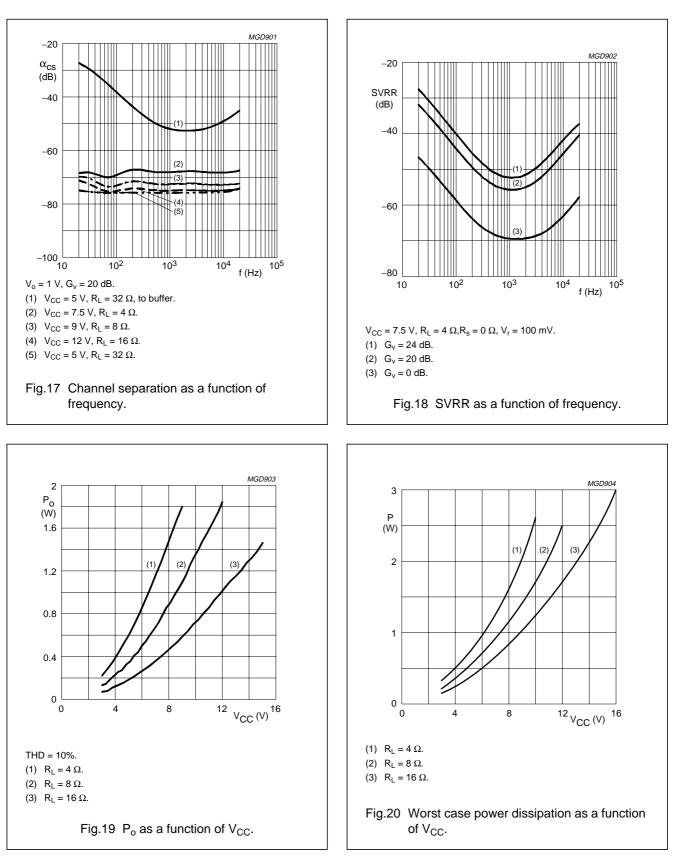


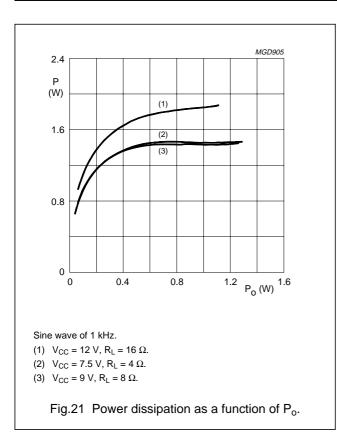
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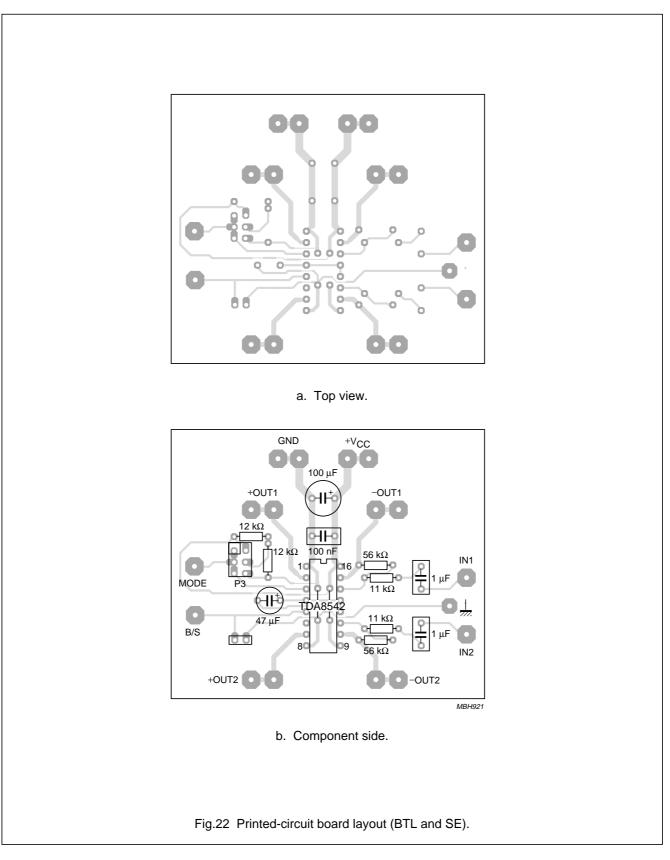
SE APPLICATION











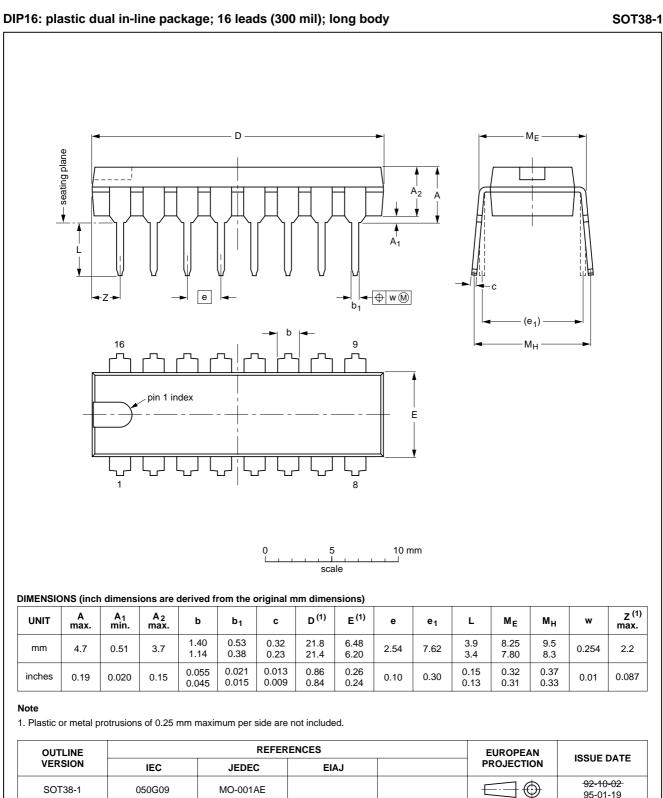
TDA8542

2×1 W BTL audio amplifier

PACKAGE OUTLINES

SO16: plastic small outline package; 16 leads; body width 7.5 mm D Α Х = v 🕅 A H_{E} 🛛 у 16 Q A pin 1 index Lp - 1 Π Π Г 8 detail X ► e + **(** w M bp 10 mm 5 0 scale DIMENSIONS (inch dimensions are derived from the original mm dimensions) Α z⁽¹⁾ D⁽¹⁾ E⁽¹⁾ ${\sf H}_{\sf E}$ UNIT **A**₁ L Q θ С v w A₂ A_3 bp е Lp у max 0.30 2.45 0.49 0.32 10.5 7.6 10.65 1.1 1.1 0.9 2.65 mm 0.25 1.27 1.4 0.25 0.25 0.1 2.25 0.36 0.23 7.4 10.00 0.4 1.0 0.4 0.10 10.1 8° 0° 0.035 0.043 0.043 0.012 0.096 0.019 0.013 0.41 0.30 0.419 inches 0.10 0.01 0.050 0.055 0.01 0.01 0.004 0.004 0.089 0.014 0.009 0.40 0.29 0.394 0.016 0.039 0.016 Note 1. Plastic or metal protrusions of 0.15 mm maximum per side are not included. REFERENCES EUROPEAN OUTLINE ISSUE DATE PROJECTION VERSION IEC JEDEC EIAJ 95-01-24 97-05-22 SOT162-1 075E03 MS-013AA

SOT162-1



1998 Apr 01

TDA8542

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TDA8542

SOLDERING

Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our *"IC Package Databook"* (order code 9398 652 90011).

DIP

SOLDERING BY DIPPING OR BY WAVE

The maximum permissible temperature of the solder is 260 °C; solder at this temperature must not be in contact with the joint for more than 5 seconds. The total contact time of successive solder waves must not exceed 5 seconds.

The device may be mounted up to the seating plane, but the temperature of the plastic body must not exceed the specified maximum storage temperature ($T_{stg\,max}$). If the printed-circuit board has been pre-heated, forced cooling may be necessary immediately after soldering to keep the temperature within the permissible limit.

REPAIRING SOLDERED JOINTS

Apply a low voltage soldering iron (less than 24 V) to the lead(s) of the package, below the seating plane or not more than 2 mm above it. If the temperature of the soldering iron bit is less than $300 \,^{\circ}$ C it may remain in contact for up to 10 seconds. If the bit temperature is between 300 and 400 $^{\circ}$ C, contact may be up to 5 seconds.

SO

REFLOW SOLDERING

Reflow soldering techniques are suitable for all SO packages.

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement. Several techniques exist for reflowing; for example, thermal conduction by heated belt. Dwell times vary between 50 and 300 seconds depending on heating method. Typical reflow temperatures range from 215 to 250 °C.

Preheating is necessary to dry the paste and evaporate the binding agent. Preheating duration: 45 minutes at 45 $^{\circ}$ C.

WAVE SOLDERING

Wave soldering techniques can be used for all SO packages if the following conditions are observed:

- A double-wave (a turbulent wave with high upward pressure followed by a smooth laminar wave) soldering technique should be used.
- The longitudinal axis of the package footprint must be parallel to the solder flow.
- The package footprint must incorporate solder thieves at the downstream end.

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Maximum permissible solder temperature is 260 °C, and maximum duration of package immersion in solder is 10 seconds, if cooled to less than 150 °C within 6 seconds. Typical dwell time is 4 seconds at 250 °C.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

REPAIRING SOLDERED JOINTS

Fix the component by first soldering two diagonallyopposite end leads. Use only a low voltage soldering iron (less than 24 V) applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 °C. When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and 320 °C.

TDA8542

DEFINITIONS

Data sheet status				
Objective specification	This data sheet contains target or goal specifications for product development.			
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.			
Product specification	This data sheet contains final product specifications.			
Limiting values				
more of the limiting values r of the device at these or at a	accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or nay cause permanent damage to the device. These are stress ratings only and operation any other conditions above those given in the Characteristics sections of the specification imiting values for extended periods may affect device reliability.			
Application information				

Where application information is given, it is advisory and does not form part of the specification.

LIFE SUPPORT APPLICATIONS

These products are not designed for use in life support appliances, devices, or systems where malfunction of these products can reasonably be expected to result in personal injury. Philips customers using or selling these products for use in such applications do so at their own risk and agree to fully indemnify Philips for any damages resulting from such improper use or sale.

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Netherlands: Postbus 90050, 5600 PB EINDHOVEN, Bldg. VB, Tel. +31 40 27 82785, Fax. +31 40 27 88399 New Zealand: 2 Wagener Place, C.P.O. Box 1041, AUCKLAND, Tel. +64 9 849 4160, Fax. +64 9 849 7811

Norway: Box 1, Manglerud 0612, OSLO, Tel. +47 22 74 8000, Fax. +47 22 74 8341

Pakistan: see Singapore

Philippines: Philips Semiconductors Philippines Inc., 106 Valero St. Salcedo Village, P.O. Box 2108 MCC, MAKATI, Metro MANILA, Tel. +63 2 816 6380, Fax. +63 2 817 3474

Poland: Ul. Lukiska 10, PL 04-123 WARSZAWA, Tel. +48 22 612 2831, Fax. +48 22 612 2327

Portugal: see Spain

Romania: see Italy

Russia: Philips Russia, Ul. Usatcheva 35A, 119048 MOSCOW, Tel. +7 095 755 6918, Fax. +7 095 755 6919

Singapore: Lorong 1, Toa Payoh, SINGAPORE 319762,

Tel. +65 350 2538, Fax. +65 251 6500

Slovakia: see Austria Slovenia: see Italy

South Africa: S.A. PHILIPS Pty Ltd., 195-215 Main Road Martindale, 2092 JOHANNESBURG, P.O. Box 7430 Johannesburg 2000, Tel. +27 11 470 5911, Fax. +27 11 470 5494

South America: Al. Vicente Pinzon, 173, 6th floor, 04547-130 SÃO PAULO, SP, Brazil, Tel. +55 11 821 2333, Fax. +55 11 821 2382

Spain: Balmes 22, 08007 BARCELONA, Tel. +34 3 301 6312, Fax. +34 3 301 4107

Sweden: Kottbygatan 7, Akalla, S-16485 STOCKHOLM, Tel. +46 8 5985 2000, Fax. +46 8 5985 2745

Switzerland: Allmendstrasse 140, CH-8027 ZÜRICH, Tel. +41 1 488 2741 Fax. +41 1 488 3263

Taiwan: Philips Semiconductors, 6F, No. 96, Chien Kuo N. Rd., Sec. 1, TAIPEI, Taiwan Tel. +886 2 2134 2865, Fax. +886 2 2134 2874

Thailand: PHILIPS ELECTRONICS (THAILAND) Ltd., 209/2 Sanpavuth-Bangna Road Prakanong, BANGKOK 10260, Tel. +66 2 745 4090, Fax. +66 2 398 0793

Turkey: Talatpasa Cad. No. 5, 80640 GÜLTEPE/ISTANBUL, Tel. +90 212 279 2770, Fax. +90 212 282 6707

Ukraine: PHILIPS UKRAINE, 4 Patrice Lumumba str., Building B, Floor 7, 252042 KIEV, Tel. +380 44 264 2776, Fax. +380 44 268 0461

United Kingdom: Philips Semiconductors Ltd., 276 Bath Road, Hayes, MIDDLESEX UB3 5BX, Tel. +44 181 730 5000, Fax. +44 181 754 8421 United States: 811 East Arques Avenue, SUNNYVALE, CA 94088-3409, Tel. +1 800 234 7381

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