## 8-BIT MCUs WITH 16K ROM/OTP/EPROM, 512 BYTES RAM, ADC, DAC (PWM), TIMER, RC AND SCI

- User Program Memory (ROM/OTP/EPROM): 16K bytes
- Data RAM: 512 bytes, including 256 bytes of stack
- Master Reset and Power-On Reset
- Run, Wait, Slow, Halt and RAM Retention modes
- 23 to 32 I/O lines:
-4 to 5 programmable interrupt inputs
-6 to 8 high sink outputs
- 4 to 8 analog alternate inputs
- 10 to 14 alternate functions
- EMI filtering
- Programmable watchdog (WDG)
- 16-bit Timer, featuring:
- 2 Input Captures
- 2 Output Compares (with 1 output pin)
- PWM and Pulse Generator modes
- 8-bit Analog-to-Digital converter $(4$ to 8 channels)
- Four 10-bit and one 12-bit Digital to Analog Converter Channels with PWM output
- Fast ${ }^{2} \mathrm{C}$ Multimaster Interface
- Serial Communications Interface (SCI) (ST72371N only)
- 8-bit Data Manipulation
- 63 Basic Instructions
- 17 main Addressing Modes
- $8 \times 8$ Unsigned Multiply Instruction
- True Bit Manipulation
- Complete Development Support on PC/DOSWINDOWS ${ }^{\text {TM }}$ Real-Time Emulator
- Full Software Package on DOS/WINDOWSM (C-Compiler, Cross-Assembler, Debugger)


Device Summary

| Features | ST72372J4 | ST72371N4 |
| :---: | :---: | :---: |
| Program Memory - bytes | 16K |  |
| RAM (stack) - bytes | 512 (256) |  |
| 10-Bit D/A Converter | 4 channels |  |
| 12-Bit D/A Converter | 1 channel |  |
| A/D Converter | 4 channels | 8 channels |
| 16-Bit Timer | 1 |  |
| $\mathrm{I}^{2} \mathrm{C}$ Bus | 1 multimaster |  |
| SCI | No | Yes |
| I/Os | 30 | 39 |
| Operating Supply | 4.0 to 5.5 V |  |
| CPU Frequency | $8 \mathrm{MHz} \max (24 \mathrm{MHz}$ quartz) |  |
| Temperature Range | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ |  |
| Package | SDIP42 | SDIP56TQFP64 |

Rev. 1.1

[^0]
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## 1 GENERAL DESCRIPTION

### 1.1 INTRODUCTION

The ST72371/ST72372 HCMOS Microcontroller Units are members of the ST7 family. These devices are based on an industry standard 8-bit core and feature an enhanced instruction set. The processor runs with an external clock up to 24 MHz with a 5.5 V supply. Under software control the ST72371/ST72372 can be placed in WAIT, SLOW or HALT modes thus reducing power consumption. The enhanced instruction set and addressing modes afford real programming potential.

In addition to standard 8-bit data management the ST7 features true bit manipulation, $8 \times 8$ unsigned multiplication and indirect addressing modes on the whole memory.
The devices include an on-chip oscillator, CPU, 16K bytes program memory, 512 bytes RAM, 23 to 32 I/O lines, a Timer with 2 Input Captures and 2 Output Compares (with 1 output pin), a 4 to 8 channel A/D Converter, an FC multi Master, an SCI ${ }^{(1)}$ Serial Communications Interface, a Watchdog Reset, four 10-bit and one 12-bit D/A Converter channels with PWM output.

Figure 1.ST72371/ST72372 Block Diagram

${ }^{(1)}$ ST72371N only

### 1.2 PIN DESCRIPTION

Figure 2. ST72371N 64-Pin QFP Pinout


Figure 3. ST72371J 42-Pin SDIP Pinout


Figure 4. ST72371N 56-Pin SDIP Pinout
DA0
DA1
DA2

Note: Several pins of the I/O ports assume software programmable alternate functions as shown in the pin description.

## PIN DESCRIPTION (Cont'd)

Table 1. ST72371N Pin Description

| $\begin{aligned} & \text { Pin } n^{\circ} \\ & \text { QFP64 } \end{aligned}$ | Pin $n^{\circ}$ SDIP56 | Pin Name | Type | Description | Remarks |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 1 |  | NC |  | Not Connected |  |
| 2 | 22 | PD6 | I/O | Port D6 |  |
| 3 | 23 | PD5 | I/O | Port D5 | External Interrupt: EIO |
| 4 | 24 | PD4 | I/O | Port D4 | External Interrupt: EI1 |
| 5 | 25 | PD3 | I/O | Port D3 | External Interrupt: El2 |
| 6 | 26 | PD2 | 1/O | Port D2 |  |
| 7 | 27 | PD1 | I/O | Port D1 |  |
| 8 | 28 | PD0 | I/O | Port D0 |  |
| 9 | 29 | $\mathrm{V}_{\text {SS }}$ | S | Ground |  |
| 10 | 30 | ICAP2 | I | Timer Input Capture 2 with 256 prescaler | Not for general purpose I/O |
| 11 | 31 | $V_{\text {DD }}$ | S | Main power supply |  |
| 12 | 32 | PC0/OCMP | I/O | Port C0 or Timer Output Compare |  |
| 13 | 33 | PC1 | 1/0 | Port C1 |  |
| 14 | 34 | PC2 | I/O | Port C2 | External Interrupt: El4 |
| 15 | 35 | PC3 | I/O | Port C3 |  |
| 16 |  | NC |  | Not Connected |  |
| 17 |  | NC |  | Not Connected |  |
| 18 | 36 | PC4/SCLI | I/O | Port C4 or ${ }^{2} \mathrm{C}$ Serial Clock |  |
| 19 | 37 | PC5/SDAI | 1/0 | Port C5 or ${ }^{2} \mathrm{C}$ Serial Data |  |
| 20 | 38 | PC6/RDI | I/O | Port C6, or SCI Receive Data Input |  |
| 21 | 39 | PC7/TDO | I/O | Port C7 or SCI Transmit Data Output |  |
| 22 | 40 | NU |  | Non User pin. Must be left unconnected |  |
| 23 | 41 | NU |  | Non User pin. Must be left unconnected |  |
| 24 | 42 | NU |  | Non User pin. Must be left unconnected |  |
| 25 | 43 | NU |  | Non User pin. Must be left unconnected |  |
| 26 | 44 | OSCOUT | 0 | Input/Output Oscillator pin.These pins connect a parallel-resonant crystal, or an external source to the on-chip oscillator. |  |
| 27 | 45 | OSCIN | I |  |  |
| 28 | 46 | PA7 | 1/0 | Port A7, High Sink |  |
| 29 | 47 | PA6 | I/O | Port A6, High Sink |  |
| 30 | 48 | PA5 | I/O | Port A5, High Sink |  |
| 31 | 49 | PA4 | I/O | Port A4, High Sink |  |
| 32 |  | NC |  | Not Connected |  |
| 33 |  | NC |  | Not Connected |  |
| 34 | 50 | PA3 | 1/O | Port A3, High Sink |  |
| 35 | 51 | PA2 | I/O | Port A2, High Sink |  |
| 36 | 52 | PA1 | I/O | Port A1, High Sink |  |
| 37 | 53 | PA0 | I/O | Port A0, High Sink |  |
| 38 | 54 | RESET | I/O | Bidirectional. Active low. Top priority non maskable interrupt. | Can be used to reset external peripherals. |
| 39 | 55 | NU | I/O | Non User pin. Must be left unconnected |  |


| Pin $\mathbf{n}^{\circ}$ <br> QFP64 | Pin $\mathbf{n}^{\circ}$ <br> SDP56 | Pin Name | Type | Description |  |
| :--- | :--- | :--- | :---: | :--- | :--- |

Table 2. ST72372J Pin Description

| Pin | Pin Name | Type | Description | Remarks |
| :---: | :---: | :---: | :---: | :---: |
| 1 | DA1 | $\bigcirc$ | 10-bit D/A (PWM output) |  |
| 2 | DA2 | $\bigcirc$ | 10-bit D/A (PWM output) | For analog controls, after external |
| 3 | DA3 | 0 | 10-bit D/A (PWM output) | filtering |
| 4 | DA4 | 0 | 10-bit D/A (PWM output) |  |
| 5 | NU |  | Non User pin. Must be left unconnected |  |
| 6 | NU |  | Non User pin. Must be left unconnected |  |
| 7 | NU |  | Non User pin. Must be left unconnected |  |
| 8 | NU |  | Non User pin. Must be left unconnected |  |
| 9 | $V_{\text {SSA }}$ | S | Ground for analog peripheral (ADC) | Must be connected externally to $\mathrm{V}_{\text {S }}$ |
| 10 | $\mathrm{V}_{\text {DDA }}$ | S | Power Supply for analog peripheral (ADC) | Must be connected externally to $\mathrm{V}_{\mathrm{DD}}$ |
| 11 | PB7/AIN3 | I/O | Port B7 or ADC analog input 3 |  |
| 12 | PB2/AIN2 | I/O | Port B2 or ADC analog input 2 |  |


| Pin | Pin Name | Type | Description | Remarks |
| :---: | :---: | :---: | :---: | :---: |
| 13 | PB1/AIN1 | I/O | Port B1 or ADC analog input 1 |  |
| 14 | PB0/AIN0 | 1/O | Port B0 or ADC analog input 0 |  |
| 15 | ICAP1 | I | Timer Input Capture 1 | Not for general purpose I/O |
| 16 | PD6 | I/O | Port D6 |  |
| 17 | PD5 | I/O | Port D5 | External Interrupt: EIO |
| 18 | PD4 | I/O | Port D4 | External Interrupt: El1 |
| 19 | PD3 | I/O | Port D3 | External Interrupt: El2 |
| 20 | PD2 | I/O | Port D2 |  |
| 21 | PD1 | 1/0 | Port D1 |  |
| 22 | PD0 | I/O | Port D0 |  |
| 23 | $\mathrm{V}_{\text {SS }}$ | S | Ground |  |
| 24 | ICAP2 | I | Timer Input Capture 2 with 256 prescaler | Not for general purpose I/O |
| 25 | $V_{\text {DD }}$ | S | Main power supply |  |
| 26 | PC0/OCMP | I/O | Port C0 or Timer Output Compare |  |
| 27 | PC2 | I/O | Port C2 | External Interrupt: EI4 |
| 28 | PC3 | I/O | Port C3 |  |
| 29 | PC4/SCLI | I/O | Port C4 or ${ }^{2} \mathrm{C}$ Serial Clock |  |
| 30 | PC5/SDAI | I/O | Port C5 or ${ }^{2} \mathrm{C}$ Serial Data |  |
| 31 | PC6 | I/O | Port C6, High Sink |  |
| 32 | OSCOUT | 0 | Input/Output Oscillator pin. These pins connect a parallel-resonant crystal, or an external source to the on-chip oscillator. |  |
| 33 | OSCIN | I |  |  |
| 34 | PA7 | I/O | Port A7, High Sink |  |
| 35 | PA6 | I/O | Port A6, High Sink |  |
| 36 | PA5 | I/O | Port A5, High Sink |  |
| 37 | PA4 | I/O | Port A4, High Sink |  |
| 38 | PA3 | I/O | Port A3, High Sink |  |
| 39 | PA1 | I/O | Port A1, High Sink |  |
| 40 | RESET | I/O | Bidirectional. Active low signal. Top priority non maskable interrupt. | Can be used to reset external periph erals. |
| 41 | TEST/V ${ }_{\text {PP }}$ | S | Test mode pin or EPROM programming voltage. In EPROM devices acts as programming voltage input $V_{\text {PP. }}$. | This pin should be tied low in user mode |
| 42 | DAO | O | 12-bit DAC (PWM output) | For analog controls, after external filtering |

Note: S= Supply

### 1.3 MEMORY MAP

Figure 5. Program Memory Map


Table 3. Interrupt Vector Map

| Vector Address | Description | Remarks |
| :---: | :---: | :---: |
| FFE0-FFE1h | Reserved |  |
| FFE2-FFE3h | SCI Interrupt Vector | Internal Interrupts |
| FFE4-FFE5h | IC Interrupt Vector | " |
| FFE6-FFE7h | Timer Overflow Interrupt Vector | " |
| FFE8-FFE9h | Timer Output Compare Interrupt Vector | " |
| FFEA-FFEBh | Timer Input Capture Interrupt Vector | " |
| FFEC-FFEDh | Reserved |  |
| FFEE-FFEFh | El4 Interrupt Vector | External Interrupts |
| FFF0-FFF1h | EI0 Interrupt Vector | " |
| FFF2-FFF3h | EI1 Interrupt Vector | " |
| FFF4-FFF5h | EI2 Interrupt Vector | " |
| FFF6-FFF7h | EI3 Interrupt Vector |  |
| FFF8-FFF9h | Reserved |  |
| FFFA-FFFBh | Reserved |  |
| FFFC-FFFDh | TRAP Interrupt Vector | Software Interrupt |
| FFFE-FFFFh | RESET Vector | CPU Interrupt |

## MEMORY MAP (Cont'd)

Table 4. Hardware Register Memory Map

| Address | Block | Register Label | Register Name | Reset Status | Remarks |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { 0000h } \\ & 0001 \mathrm{~h} \end{aligned}$ | Port A | $\begin{array}{\|l\|} \hline \text { PADR } \\ \text { PADDR } \end{array}$ | Port A Data Register <br> Port A Data Direction Register | $\begin{array}{\|l\|} \hline 00 \mathrm{~h} \\ 00 \mathrm{~h} \end{array}$ | $\begin{aligned} & \hline R / W \\ & R / W \end{aligned}$ |
| $\begin{aligned} & 0002 \mathrm{~h} \\ & 0003 \mathrm{~h} \end{aligned}$ | Port C | $\begin{aligned} & \hline \text { PCDR } \\ & \text { PCDDR } \end{aligned}$ | Port C Data Register <br> Port C Data Direction Register | $\begin{aligned} & \text { 00h } \\ & 00 \mathrm{~h} \end{aligned}$ | $\begin{aligned} & \hline R / W \\ & R / W \end{aligned}$ |
| $\begin{aligned} & 0004 \mathrm{~h} \\ & 0005 \mathrm{~h} \end{aligned}$ | Port D | $\begin{aligned} & \hline \text { PDDR } \\ & \text { PDDDR } \end{aligned}$ | Port D Data Register <br> Port D Data Direction Register | $\begin{array}{\|l\|} \hline 00 \mathrm{~h} \\ 00 \mathrm{~h} \end{array}$ | $\begin{aligned} & \hline R / W \\ & R / W \end{aligned}$ |
| $\begin{aligned} & \hline 0006 \mathrm{~h} \\ & 0007 \mathrm{~h} \\ & 0008 \mathrm{~h} \end{aligned}$ | Port B | $\begin{aligned} & \hline \text { PBDR } \\ & \text { PBDDR } \\ & \text { PBOR } \end{aligned}$ | Port B Data Register <br> Port B Data Direction Register <br> Port B Option Register | $\begin{array}{\|l\|} \hline 00 \mathrm{~h} \\ 00 \mathrm{~h} \\ 00 \mathrm{~h} \end{array}$ | $\begin{aligned} & \hline R / W \\ & R / W \\ & R / W \end{aligned}$ |
| 0009h |  | MISCR | Miscellaneous Register | 00h | R/W |
| $\begin{aligned} & \text { 000Ah } \\ & \text { 000Bh } \end{aligned}$ | ADC | $\begin{aligned} & \text { ADCDR } \\ & \text { ADCCSR } \end{aligned}$ | ADC Data Register ADC Control Status register | $\begin{aligned} & \hline 00 \mathrm{~h} \\ & 00 \mathrm{~h} \end{aligned}$ | Read only R/W |
| 000Ch | WDG | WDGCR | Watchdog Control Register | 7Fh | R/W |
| $\begin{aligned} & \text { 000Dh } \\ & \text { 000Fh } \end{aligned}$ | Reserved Area (3 bytes) |  |  |  |  |
| 0010h | ITR | ITRFRE | Interrupt Register | 00h | R/W |
| 0011h <br> 0012h <br> 0013h <br> 0014h <br> 0015h <br> 0016h <br> 0017h <br> 0018h <br> 0019h <br> 001Ah <br> 001Bh <br> 001Ch <br> 001Dh <br> 001Eh <br> 001Fh | TIM | TIMCR2 <br> TIMCR1 <br> TIMSR <br> TIMIC1HR <br> TIMIC1LR <br> TIMOC1HR <br> TIMOC1LR <br> TIMCHR <br> TIMCLR <br> TIMACHR <br> TIMACLR <br> TIMIC2HR <br> TIMIC2LR <br> TIMOC2HR <br> TIMOC2LR | Timer Control Register 2 <br> Timer Control Register 1 <br> Timer Status Register <br> Timer Input Capture 1 High Register <br> Timer Input Capture 1 Low Register <br> Timer Output Compare 1 High Register <br> Timer Output Compare 1 Low Register <br> Timer Counter High Register <br> Timer Counter Low Register <br> Timer Alternate Counter High Register <br> Timer Alternate Counter Low Register <br> Timer Input Capture 2 High Register <br> Timer Input Capture 2 Low Register <br> Timer Output Compare 2 High Register <br> Timer Output Compare 2 Low Register | 00h <br> 00h <br> 00h <br> xxh <br> xxh <br> 80h <br> 00h <br> FFh <br> FCh <br> FFh <br> FCh <br> xxh <br> xxh <br> 80h <br> 00h | R/W <br> R/W <br> Read only <br> Read only <br> Read only <br> R/W <br> R/W <br> Read only <br> Read only <br> Read only <br> Read only <br> Read only <br> Read only <br> R/W <br> R/W |
| $\begin{aligned} & \text { 0020h } \\ & 0021 \mathrm{~h} \end{aligned}$ | Reserved Area (2 bytes) |  |  |  |  |
| $\begin{aligned} & \text { 0022h } \\ & 0023 \mathrm{~h} \end{aligned}$ |  | $\begin{aligned} & \hline \text { PWM0 } \\ & \text { BRM0 } \end{aligned}$ | 12-BIT PWM Register 12-BIT BRM Register | $\begin{array}{\|l} \hline 80 \mathrm{~h} \\ \mathrm{COh} \end{array}$ | $\begin{aligned} & \hline R / W \\ & R / W \end{aligned}$ |
| $\begin{aligned} & 0024 \mathrm{~h} \\ & 0025 \mathrm{~h} \\ & 0026 \mathrm{~h} \\ & 0027 \mathrm{~h} \\ & 0028 \mathrm{~h} \\ & 0029 \mathrm{~h} \end{aligned}$ | DAC | PWM1 BRM21 <br> PWM2 <br> PWM3 <br> BRM43 <br> PWM4 | 10-BIT PWM / BRM Registers | 80h <br> 00h <br> 80h <br> 80h <br> 00h <br> 80h | $\begin{aligned} & \hline R / W \\ & R / W \\ & R / W \\ & R / W \\ & R / W \\ & R / W \end{aligned}$ |
| 002Ah to 002Fh | Reserved Area (6 bytes) |  |  |  |  |


| Address | Block | Register Label | Register Name | Reset Status | Remarks |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \hline 0030 \mathrm{~h} \\ & 0031 \mathrm{~h} \\ & 0032 \mathrm{~h} \\ & 0033 \mathrm{~h} \\ & 0034 \mathrm{~h} \end{aligned}$ | SCI | SCISR SCIDR SCIBRR SCICR1 SCICR2 | SCI Status Register SCI Data Register SCI Baud Rate Register SCI Control Register 1 SCI Control Register 2 | C0h xxh $00 x x$ xxxxb xxh $00 h$ | $\begin{aligned} & \hline \text { Read only } \\ & \text { R/W } \\ & \text { R/W } \\ & \text { R/W } \\ & \text { R/W } \end{aligned}$ |
| $\begin{array}{\|l\|} \hline 0035 \mathrm{~h} \text { to } \\ 0042 \mathrm{~h} \end{array}$ | Reserved Area (14 bytes) |  |  |  |  |
| 0043h | TIM | CONFIG | ICAP Pin Configuration Warning: Write 0Ch in this register to use the ICAP1 and ICAP2 functions. | 08h | R/W |
| 0044h to 0058h | Reserved Area (21 bytes) |  |  |  |  |
| $\begin{aligned} & \text { 0059h } \\ & \text { 005Ah } \\ & 005 \mathrm{Bh} \\ & 005 \mathrm{Ch} \\ & \text { 005Dh } \\ & \text { 005Eh } \\ & 005 \mathrm{Fh} \end{aligned}$ | $1^{2} \mathrm{C}$ | $\begin{aligned} & \text { I2CDR } \\ & \text { I2COAR } \\ & \text { I2CCCR } \\ & \text { I2CSR2 } \\ & \text { I2CSR1 } \\ & \text { I2CCR } \end{aligned}$ | ${ }^{2}{ }^{2} \mathrm{C}$ Data Register Reserved ${ }^{2}{ }^{2} \mathrm{C}$ (7 Bits) Slave Address Register ${ }^{2}{ }^{2} \mathrm{C}$ Clock Control Register $\mathrm{I}^{2} \mathrm{C}$ Status Register 2 $I^{2} \mathrm{C}$ Status Register 1 $1^{2} \mathrm{C}$ Control Register | $\begin{aligned} & 00 \mathrm{~h} \\ & 00 \mathrm{~h} \\ & 00 \mathrm{~h} \\ & 00 \mathrm{~h} \\ & 00 \mathrm{~h} \\ & 00 \mathrm{~h} \end{aligned}$ | R/W <br> R/W <br> R/W <br> Read only <br> Read only <br> R/W |
| $\begin{array}{\|l} \hline \text { 0060h to } \\ \text { 007Fh } \end{array}$ | Reserved Area (32 bytes) |  |  |  |  |

## 2 CENTRAL PROCESSING UNIT

### 2.1 Introduction

This CPU has a full 8-bit architecture and contains six internal registers allowing efficient 8-bit data manipulation.

### 2.2 Main Features

- Enable executing 63 basic instructions
- Fast 8-bit by 8-bit multiply
- 17 main addressing modes (with indirect addressing mode)
- Two 8-bit index registers
- 16-bit stack pointer
- 8 MHz CPU internal frequency
- Low power modes
- Maskable hardware interrupts
- Non-maskable software interrupt


### 2.3 CPU Registers

The 6 CPU registers shown in Figure 1 are not present in the memory mapping and are accessed by specific instructions.

## Accumulator (A)

The Accumulator is an 8-bit general purpose register used to hold operands and the results of the arithmetic and logic calculations and to manipulate data.

## Index Registers (X and Y)

In indexed addressing modes, these 8-bit registers are used to create either effective addresses or temporary storage areas for data manipulation. (The Cross-Assembler generates a precede instruction (PRE) to indicate that the following instruction refers to the Y register.)
The Y register is not affected by the interrupt automatic procedures (not pushed to and popped from the stack).

## Program Counter (PC)

The program counter is a 16-bit register containing the address of the next instruction to be executed by the CPU. It is made of two 8 -bit registers PCL (Program Counter Low which is the LSB) and PCH (Program Counter High which is the MSB).

Figure 6. CPU Registers


## CENTRAL PROCESSING UNIT(Cont'd)

 Condition Code Register (CC)Read/Write

Reset Value: 111x1010

| 7 |
| :--- | |  |  | 0 |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 1 | 1 | 1 | $H$ | I | N | Z | C |

The 8 -bit Condition Code register contains the interrupt mask and four flags representative of the result of the instruction just executed. This register can also be handled by the PUSH and POP instructions.

These bits can be individually tested and/or controlled by specific instructions.

## Bit 4 = H Half carry.

This bit is set by hardware when a carry occurs between bits 3 and 4 of the ALU during an ADD or ADC instruction. It is reset by hardware during the same instructions.
0 : No half carry has occurred.
1: A half carry has occurred.
This bit is tested using the JRH or JRNH instruction. The H bit is useful in BCD arithmetic subroutines.

## bit 3 = I Interrupt mask.

This bit is set by hardware when entering in interrupt or by software to disable all interrupts except the TRAP software interrupt. This bit is cleared by software.
0: Interrupts are enabled.
1: Interrupts are disabled.
This bit is controlled by the RIM, SIM and IRET instructions and is tested by the JRM and JRNM instructions.

Note: Interrupts requested while I is set are latched and can be processed when I is cleared. By default an interrupt routine is not interruptable because the I bit is set by hardware when you en-
ter it and reset by the IRET instruction at the end of the interrupt routine. If the I bit is cleared by software in the interrupt routine, pending interrupts are serviced regardless of the priority level of the current interrupt routine.

## Bit $2=\mathbf{N}$ Negative.

This bit is set and cleared by hardware. It is representative of the result sign of the last arithmetic, logical or data manipulation. It is a copy of the $t^{\text {th }}$ bit of the result.
0 : The result of the last operation is positive or null.
1: The result of the last operation is negative
(i.e. the most significant bit is a logic 1 ).

This bit is accessed by the JRMI and JRPL instructions.

## Bit $1=\mathbf{Z}$ Zero.

This bit is set and cleared by hardware. This bit indicates that the result of the last arithmetic, logical or data manipulation is zero.
0 : The result of the last operation is different from zero.
1: The result of the last operation is zero.
This bit is accessed by the JREQ and JRNE test instructions.

## Bit $0=\mathbf{C}$ Carry/borrow.

This bit is set and cleared by hardware and software. It indicates an overflow or an underflow has occurred during the last arithmetic operation.
0: No overflow or underflow has occurred.
1: An overflow or underflow has occurred.
This bit is driven by the SCF and RCF instructions and tested by the JRC and JRNC instructions. It is also affected by the "bit test and branch", shift and rotate instructions.

## CENTRAL PROCESSING UNIT(Cont'd)

## Stack Pointer (SP)

Read/Write
Reset Value: 01 FFh

$$
15 \quad 8
$$

| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

7

| SP7 | SP6 | SP5 | SP4 | SP3 | SP2 | SP1 | SP0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

The Stack Pointer is a 16-bit register which is always pointing to the next free location in the stack. It is then decremented after data has been pushed onto the stack and incremented before data is popped from the stack (see Figure 7).
Since the stack is 256 bytes deep, the most significant byte is forced by hardware. Following an MCU Reset, or after a Reset Stack Pointer instruction (RSP), the Stack Pointer contains its reset value (the SP7 to SP0 bits are set) which is the stack higher address.

The least significant byte of the Stack Pointer can be directly accessed by a LD instruction.
Note: When the lower limit is exceeded, the Stack Pointer wraps around to the stack upper limit, without indicating the stack overflow. The previously stored information is then overwritten and therefore lost. The stack also wraps in case of an underflow.

The stack is used to save the return address during a subroutine call and the CPU context during an interrupt. The user may also directly manipulate the stack by means of the PUSH and POP instructions. In the case of an interrupt, the PCL is stored at the first location pointed to by the SP. Then the other registers are stored in the next locations as shown in Figure 7.

- When an interrupt is received, the SP is decremented and the context is pushed on the stack.
- On return from interrupt, the SP is incremented and the context is popped from the stack.
A subroutine call occupies two locations and an interrupt five locations in the stack area.

Figure 7. Stack Manipulation Example


[^1]
## 3 CLOCKS, RESET, INTERRUPTS \& POWER SAVING MODES

### 3.1 CLOCK SYSTEM

### 3.1.1 General Description

The MCU accepts either a Crystal or Ceramic resonator, or an external clock signal to drive the internal oscillator. The internal clock (Łpu) is derived from the external oscillator frequency ( $\delta s c$ ).
The external Oscillator clock is first divided by 3, and an additional division factor of 2 can be applied if Slow Mode is selected by resetting the SMS bit in the Miscellaneous Register. This reduces the frequency of the frpu; the clock signal is also routed to the on-chip peripherals. The CPU clock signal consists of a square wave with a duty cycle of $50 \%$.
The internal oscillator is designed to operate with an AT-cut parallel resonant quartz crystal resonator in the frequency range specified for $f_{s c}$. The circuit shown in Figure 8 is recommended when using a crystal, and Table 5 lists the recommended capacitance and feedback resistance values. The crystal and associated components should be mounted as close as possible to the input pins in order to minimize output distortion and start-up stabilisation time.
Use of an external CMOS oscillator is recommended when crystals outside the specified frequency ranges are to be used.

Table 5. Recommended Crystal Values

| $\mathbf{2 4} \mathbf{~ M h z}$ |  |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{R}_{\mathrm{SMAX}}$ | 70 | 25 | 20 | Ohms |
| $\mathrm{C}_{\mathrm{L} 1}$ | 22 | 47 | 56 | pf |
| $\mathrm{C}_{\mathrm{L} 2}$ | 22 | 47 | 56 | pf |

Legend:
$C_{L 1}, C_{L 2}=$ Maximum total capacitance on pins OSCIN and OSCOUT (the value includes the external capacitance tied to the pin plus the parasitic capacitance of the board and of the device).
$R_{\text {SMAX }}=$ Maximum series parasitic resistance of the quartz allowed.

Note: The tables relate to the quartz crystal only (not ceramic resonator).

### 3.1.2 External Clock

An external clock may be applied to the OSCIN input with the OSCOUT pin not connected. The あxov specifications does not apply when using an external clock input. The equivalent specification of the external clock source should be used instead of toxov (see Electrical Characteristics).

Figure 8. Crystal/Ceramic Resonator


Figure 9. Clock Prescaler Block Diagram


### 3.2 RESET

### 3.2.1 Introduction

There are three sources of Reset:

- RESET pin (external source)
- Power-On Reset (Internal source)
- WATCHDOG (Internal Source)

The Reset Service Routine vector is located at address FFFEh-FFFFh.

### 3.2.2 External Reset

The RESET pin is both an input and an open-drain output with integrated pull-up resistor. When one of the internal Reset sources is active, the Reset pin is driven low to reset the whole application.

### 3.2.3 Reset Operation

The duration of the Reset condition, which is also reflected on the output pin, is fixed at 4096 internal CPU Clock cycles. A Reset signal originating from an external source must have a duration of at least 1.5 internal CPU Clock cycles in order to be recognised. At the end of the Power-On Reset cycle, the MCU may be held in the Reset condition by an External Reset signal. The RESET pin may thus be used to ensure $V_{D D}$ has risen to a point where the MCU can operate correctly before the user program is run. Following a Power-On Reset event, or
after exiting Halt mode, a 4096 CPU Clock cycle delay period is initiated in order to allow the oscillator to stabilise and to ensure that recovery has taken place from the Reset state.
During the Reset cycle, the device Reset pin acts as an output that is pulsed low. In its high state, an internal pull-up resistor of about $300 \mathrm{~K} \Omega$ is connected to the Reset pin. This resistor can be pulled low by external circuitry to reset the device.

### 3.2.4 Power-on Reset

This circuit detects the ramping up of $V_{D D}$, and generates a pulse that is used to reset the application (at approximately $V_{D D}=2 \mathrm{~V}$ ).
Power-On Reset is designed exclusively to cope with power-up conditions, and should not be used in order to attempt to detect a drop in the power supply voltage.
Caution: to re-initialize the Power-On Reset, the power supply must fall below approximately 0.8 V (Vtn), prior to rising above 2V. If this condition is not respected, on subsequent power-up the Reset pulse may not be generated. An external Reset pulse may be required to correctly reactivate the circuit.

Figure 10. Reset Block Diagram


RESET (Cont'd)
Table 6. List of sections affected by RESET, WAIT and HALT (Refer to 3.6 for Wait and Halt Modes)

| Section | RESET | WAIT | HALT |
| :--- | :--- | :--- | :--- |
| CPU clock running at 4 MHz | X |  |  |
| Timer Prescaler reset to zero | X |  |  |
| Timer Counter set to FFFCh | X |  |  |
| All Timer enable bits set to 0 (disabled) | X |  |  |
| Data Direction Registers set to 0 (as Inputs) | X |  |  |
| Set Stack Pointer to 01FFh | X |  |  |
| Force Internal Address Bus to restart vector FFFEh, FFFFh | X |  |  |
| Set Interrupt Mask Bit (l-Bit, CC) to 1 (Interrupt disable) | X |  |  |
| Set Interrupt Mask Bit (I-Bit, CC) to 0 (Interrupt enable) |  | X | X |
| Reset HALT latch | X |  |  |
| Reset WAIT latch | X |  |  |
| Disable Oscillator (for 4096 cycles) | X |  |  |
| Set Timer Clock to 0 | X |  | X |
| Watchdog counter reset | X |  |  |
| Watchdog register reset | X |  |  |
| Port data registers reset | X |  |  |
| Other on-chip peripherals: registers reset | X |  |  |

Figure 11. Reset Timing Diagram


Note: Refer to Electrical Characteristics for values of $\ddagger D R$ and $t_{0 x O v}$

### 3.3 INTERRUPTS

The ST7 core may be interrupted by one of two different methods: maskable hardware interrupts as listed in Table 7 and a non-maskable software interrupt (TRAP). The Interrupt processing flowchart is shown in Figure 12.

The maskable interrupts must be enabled clearing the I bit in order to be serviced. However, disabled interrupts may be latched and processed when they are enabled (see external interrupts subsection).
When an interrupt has to be serviced:

- Normal processing is suspended at the end of the current instruction execution.
- The PC, X, A and CC registers are saved onto the stack.
- The I bit of the CC register is set to prevent additional interrupts.
- The PC is then loaded with the interrupt vector of the interrupt to service and the first instruction of the interrupt service routine is fetched (refer to Table 7 for vector addresses).
The interrupt service routine should finish with the IRET instruction which causes the contents of the saved registers to be recovered from the stack.
Note: As a consequence of the IRET instruction, the I bit will be cleared and the main program will resume.


## Priority management

By default, a servicing interrupt can not be interrupted because the I bit is set by hardware entering in interrupt routine.
In the case several interrupts are simultaneously pending, an hardware priority defines which one will be serviced first (seeTable 7).

## Non Maskable Software Interrupts

This interrupt is entered when the TRAP instruction is executed regardless of the state of the I bit. It will be serviced according to the flowchart on Figure 12.

## Interrupts and Low power mode

All interrupts allow the processor to leave the Wait low power mode. Only external and specific men-
tioned interrupts allow the processor to leave the Halt low power mode (refer to the "Exit from HALT" column in Table 7).

## External Interrupts

External interrupt vectors can be loaded in the PC register if the corresponding external interrupt occurred and if the I bit is cleared. These interrupts allow the processor to leave the Halt low power mode.
The external interrupt polarity can be selected through the Miscellaneous register or Interrupt register (if available) (seeSection 3.4.5).
An external interrupt triggered on edge will be latched and the interrupt request automatically cleared on entering the interrupt service routine.
More than one input pin can be connected to the same interrupt request (depending on the device). In this case, all inputs configured as interrupt are logically ORed.
Warning: The type of polarity defined in the Miscellaneous or Interrupt register (if available) applies to the El source. In case of an ORed source, a low level on an I/O pin configured as input with interrupt, masks the interrupt request even in case of rising-edge polarity.

## Peripheral Interrupts

Different peripheral interrupt flags in the status register are able to cause an interrupt when they are active if both:

- The I bit of the CC register is cleared.
- The corresponding enable bit is set in the control register.
If any of these two conditions is false, the interrupt is latched and thus remains pending.
Clearing an interrupt request is done by:
- writing " 0 " to the corresponding bit in the status register or
- an access to the status register while the flag is set followed by a read or write of an associated register.
Note: the clearing sequence resets the internal latch. A pending interrupt (i.e. waiting for being enabled) will therefore be lost if the clear sequence is executed.


## INTERRUPTS (Cont'd)

Figure 12. Interrupt Processing Flowchart


## INTERRUPTS (Cont'd)

Table 7. Interrupt Mapping

| Source Block | Description | Register Label | Flag | $\begin{aligned} & \text { Exit } \\ & \text { from } \end{aligned}$ HALT | Vector Address | Priority Order |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESET | Reset | N/A | N/A | yes | FFFEh-FFFFh | Highest Priority |
| TRAP | Software Interrupt | N/A | N/A | no | FFFCh-FFFDh |  |
| NOT USED |  |  |  |  | FFFAh-FFFBh |  |
|  |  |  |  |  | FFF8h-FFF9h |  |
| El3 | Ext. Interrupt PD7, rising edge | ITRFRE | EI3F | yes | FFF6h-FFF7h |  |
| EI2 | Ext. Interrupt PD3, falling edge | ITRFRE | El2F |  | FFF4h-FFF5h |  |
| El1 | Ext. Interrupt PD4, falling edge | ITRFRE | El1F |  | FFF2h-FFF3h |  |
| EIO | Ext. Interrupt PD5, falling edge | ITRFRE | EIOF |  | FFF0h-FFF1h |  |
| El4 | Ext. Interrupt PC2, falling edge | MISCR | EI4F |  | FFEEh-FFEFh |  |
| NOT USED |  |  |  |  | FFECh-FFEDh |  |
| TIMER | Input Capture 1 | TIMSR | ICF1 | no | FFEAh-FFEBh |  |
|  | Input Capture 2 |  | ICF2 |  | FFEAn-FFEBh |  |
|  | Output Compare 1 |  | OCF1 |  | FFE8h-FFE9h |  |
|  | Output Compare 2 |  | OCF2 |  | FFE8h-FFE9h |  |
|  | Timer Overflow |  | TOF |  | FFE6h-FFE7h |  |
| I2C | I2C Interface Interrupt | $\begin{aligned} & \hline \text { I2CSR1 } \\ & \text { I2CSR2 } \end{aligned}$ | * |  | FFE4h-FFE5h |  |
| SCI | Transmit Buffer Empty | SCISR | TDRE |  | FFE2h-FFE3h | Lowest Priority |
|  | Transmit Complete |  | TC |  |  |  |
|  | Receive Buffer Full |  | RDRF |  |  |  |
|  | Idle Line Detect |  | IDLE |  |  |  |
|  | Overrun |  | OR |  |  |  |
| NOT USED |  |  |  |  | FFEOh-FFE1h |  |

* Many flags can cause an interrupt: see peripheral interrupt status register description.


### 3.4 POWER SAVING MODES

### 3.4.1 Introduction

There are three Power Saving modes. Slow Mode is selected by setting the relevant bits in the Miscellaneous register. Wait and Halt modes may be entered using the WFI and HALT instructions.

### 3.4.2 Slow Mode

In Slow mode, the oscillator frequency can be divided by a value defined in the Miscellaneous Register. The CPU and peripherals are clocked at this lower frequency. Slow mode is used to reduce power consumption, and enables the user to adapt clock frequency to available supply voltage.
Note: On reset, Slow mode is selected by default ( $\mathrm{Fosc}^{16}$ ).

### 3.4.3 Wait Mode

Wait mode places the MCU in a low power consumption mode by stopping the CPU. All peripherals remain active. During Wait mode, the I bit (CC Register) is cleared, so as to enable all interrupts. All other registers and memory remain unchanged. The MCU will remain in Wait mode until an Interrupt or Reset occurs, whereupon the Program Counter branches to the starting address of the Interrupt or Reset Service Routine.
The MCU will remain in Wait mode until a Reset or an Interrupt occurs, causing it to wake up.
Refer to Figure 13 below.

Figure 13. WAIT Flow Chart


Note: Before servicing an interrupt, the CC register is pushed on the stack. The I-Bit is set during the interrupt routine and cleared when the CC register is popped.

## POWER SAVING MODES(Cont'd)

### 3.4.4 Halt Mode

The Halt mode is the MCU lowest power consumption mode. The Halt mode is entered by executing the HALT instruction. The internal oscillator is then turned off, causing all internal processing to be stopped, including the operation of the on-chip peripherals. The Halt mode cannot be used when the watchdog is enabled, if the HALT instruction is executed while the watchdog system is enabled, a watchdog reset is generated thus resetting the entire MCU.
When entering Halt mode, the I bit in the CC Register is cleared so as to enable External Interrupts. If an interrupt occurs, the CPU becomes active.
The MCU can exit the Halt mode upon reception of an interrupt or a reset. Refer to the Interrupt Mapping Table. The oscillator is then turned on and a stabilization time is provided before releasing CPU operation. The stabilization time is 4096 CPU clock cycles.
After the start up delay, the CPU continues operation by servicing the interrupt which wakes it up or by fetching the reset vector if a reset wakes it up.

Figure 14. HALT Flow Chart


* or some specific interrupts

Note: Before servicing an interrupt, the CC register is pushed on the stack. The l-Bit is set during the interrupt routine and cleared when the CC register is popped.

### 3.4.5 Register Description MISCELLANEOUS REGISTER (MISCR)

Read/Write
Reset Value: 00000000 (00h)


Bit 7 = EI4F Falling Edge Detector Flag.
This bit is set by hardware when a falling edge occurs on the pin assigned to EI4. An interrupt is generated if EI4ITE=1 It is cleared by software.
0 : No falling edge detected on EI4
1: Falling edge detected on EI4

Bit 6 = EI4ITE EI4 Interrupt Enable.
This bit is set and cleared by software.
0 : El4 interrupt disabled
1: El4 interrupt enabled

## Bit 5 = SMS Slow Mode Select.

This bit is set and cleared by software. It is used to select the slow or fast mode CPU frequency.
0 : $\mathrm{f}_{\mathrm{CPu}}=$ Oscillator frequency $/ 6$ (slow mode)
1: f frpu $=$ Oscillator frequency / 3 (normal mode)

## Bit 4: 1 = Reserved

Bit $0=$ POCO PWM/BRM Output Configuration Bit This bits is set and cleared by software. They select the PWM/BRM output configuration for pins DA1-DA4.
0 : Push-pull
1: Open drain
Note. DAO is only Push-Pull Output.

## INTERRUPT REGISTER (ITRFRE)

Read/Write
Reset Value: 00000000 (00h)

$$
\begin{array}{ll}
7 & 0
\end{array}
$$



Bit 7:5 = EIOF, El1F, EI2F Falling Edge Detector Flags.
These bits are set by hardware when a falling edge occurs on the pins assigned to EIO, El1 or El2. They are cleared by software. When any of these bits are set, an interrupt is generated if the corresponding ITE bit $=1$ and the I bit in the CC register $=0$.
0 : No falling edge detected
1: Falling edge detected

Bit 4 = El3F Rising Edge Detector Flag.
This bit is set by hardware when a rising edge occurs on the pin assigned to EI3. It is cleared by software. When EI3F is set an interrupt is generated if $\mathrm{EI} 3 \mathrm{ITE}=1$ and the I bit in the CC register $=0$.
0: No rising edge detected on El3
1: Rising edge detected on EI3
Bit 3:0 = EIOITE, EI1ITE, EI2ITE, EI3ITE Interrupt Enable Bits.
These bits are set and cleared by software.
0 : Interrupt disabled
1: Interrupt enabled

## 4 ON-CHIP PERIPHERALS

### 4.1 I/O PORTS

### 4.1.1 Introduction

The I/O ports offer different functional modes:

- transfer of data through digital inputs and outputs and for specific pins:
- analog signal input (ADC)
- alternate signal input/output for the on-chip peripherals.
- external interrupt generation

An I/O port is composed of up to 8 pins. Each pin can be programmed independently as digital input (with or without interrupt generation) or digital output.

### 4.1.2 Functional Description

Each port is associated to 2 main registers:

- Data Register (DR)
- Data Direction Register (DDR)
and some of them to an optional register:
- Option Register (OR)

Each I/O pin may be programmed using the corresponding register bits in DDR and OR registers: bit $X$ corresponding to pin $X$ of the port. The same correspondence is used for the DR register.
The following description takes into account the OR register, however some specific ports do not provide this register. The generic I/O block diagram is shown on Figure 15.

### 4.1.2.1 Input Modes

The input configuration is selected by clearing the corresponding DDR register bit.
In this case, reading the DR register returns the digital value applied to the external I/O pin.

Different input modes can be selected by software through the OR register.

## Notes:

1. All the inputs are triggered by a CMOS Schmitt trigger.
2. When switching from input mode to output mode, the DR register should be written first to
output the correct value as soon as the port is configured as an output.

### 4.1.2.2 External Interrupt Generation

An I/O can be used to generate an external Interrupt request to the CPU. External Interrupts are enabled and their polarity selected using the OR, MISC and ITRFRE registers (where available).
Each external interrupt vector is linked to a dedicated group of I/O port pins (see Interrupts section). If more than one input pin is selected simultaneously as interrupt source, this is logically ORed. For this reason if one of the interrupt pins is tied low, it masks the other ones.

### 4.1.2.3 Output Mode

The pin is configured in output mode by setting the corresponding DDR register bit.
In this mode, writing " 0 " or " 1 " to the DR register applies this digital value to the I/O pin through the latch. Then reading the DR register returns the previously stored value.
Note: In this mode, the interrupt function is disabled.

### 4.1.2.4 Digital Alternate Function

When an on-chip peripheral is configured to use a pin, the alternate function is automatically selected. This alternate function takes priority over standard I/O programming. When the signal is coming from an on-chip peripheral, the I/O pin is automatically configured in output mode (push-pull or open drain according to the peripheral).
When the signal is going to an on-chip peripheral, the I/O pin has to be configured in input mode. In this case, the pin's state is also digitally readable by addressing the DR register.
Note: When the on-chip peripheral uses a pin as input and output, this pin must be configured as an input (DDR = 0).
Warning: The alternate function must not be activated as long as the pin is configured as input with interrupt, in order to avoid generating spurious interrupts.

## I/O PORTS (Cont'd)

### 4.1.2.5 Analog Alternate Function

When the pin is used as an ADC input the I/O must be configured as input, floating. The analog multiplexer (controlled by the ADC registers) switches the analog voltage present on the selected pin to the common analog rail which is connected to the ADC input.
It is recommended not to change the voltage level or loading on any port pin while conversion is in progress. Furthermore it is recommended not to have clocking pins located close to a selected analog pin.
Warning: The analog input voltage level must be within the limits stated in the Absolute Maximum Ratings.

I/O PORTS (Cont'd)
Figure 15. I/O Block Diagram


Table 8. Port Mode Configuration

| Configu ration Mode | Pull-up | P-buffer |
| :--- | :---: | :---: |
| Floating | 0 | 0 |
| Pull-up | 1 | 0 |
| Push-pull | 0 | 1 |
| True Open Drain | not present | not present |
| Open Drain (logic level) | not present | 0 |

## Legend:

$\begin{array}{ll}0- & \text { present, not activated } \\ 1- & \text { present and activated }\end{array}$

## Notes:

- No OR Register on some ports (see register map).
- ADC Switch on ports with analog alternate functions.


## I/O PORTS (Cont'd)

### 4.1.2.6 Device Specific Configurations

Table 9. ST72372N Port Configuration

| Port | Pin name | Input (DDR=0) |  | Output (DDR=1) |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | OR=0* | OR=1 | OR=0 | OR=1 |
| Port A | PA0:PA7 | floating |  | true open drain, high sink capability |  |
| Port B | PB0:PB7 | pull-up | floating <br> (for analog <br> conversion only) | push-pull |  |
| Port C | PC0:PC1 | pull-up |  | push-pull |  |
|  | PC2:PC5 | floating |  | open drain |  |
|  | PC6 | pull-up |  | true open drain, high sink capability |  |
|  | PC7 | pull-up |  | push-pull |  |
| Port D | PD0:PD7 | pull-up |  | push-pull |  |

*Reset state.
Table 10. ST72372J Port Configuration

| Port | Pin name | Input (DDR=0) |  | Output (DDR=1) |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | OR=0* | OR=1 | OR=0 | OR=1 |
| Port A | PA1, PA3:7 | floating |  | true open drain, high sink capability |  |
| Port B | PB0:2, PB7 | pull-up | floating <br> (for analog <br> conversion only) | push-pull |  |
|  | PC0 | PC2:PC5 | pull-up | push-pull |  |
|  | PC6 | floating | open drain |  |  |
|  | Port D | PD0:PD6 | pull-up | push-pull |  |

*Reset state.

Note: The DA1-DA4 output pins are configurable as push pull or open drain using the POC0 Bit in the Miscellaneous Register.

I/O PORTS (Cont'd)

### 4.1.3 Register Description

## DATA REGISTERS (DR)

Read/Write
Reset Value: 00000000 (00h)

7

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7:0 = D7-D0 Data Register 8 bits.
The behaviour of the DR register depends on the selected input/output configuration. Writing the DR register is always taken in account even if the pin is configured as an input. Reading the DR register returns either the DR register latch content (pin configured as output) or the digital value applied to the I/O pin (pin configured as input).

## DATA DIRECTION REGISTERS (DDR)

Read/Write
Reset Value: 00000000 (00h) (input mode)
7

| DD7 | DD6 | DD5 | DD4 | DD3 | DD2 | DD1 | DD0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7:0 = DD7-DD0 Data Direction Register 8 bits.
The DDR register gives the input/output direction configuration of the pins. Each bit is set and cleared by software.
0 : Input mode
1: Output mode

## OPTION REGISTER (OR)

Read/Write
Reset Value: 00000000 (00h)

| AD7 | AD6 | AD5 | AD4 | AD3 | AD2 | AD1 | AD0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7:0 = AD[7:0] Port B Digital/Analog Input Configuration Bits.
0 : The pull-up is connected and pin configured as digital input (reset condition)
1: The pull-up is disconnected and the pin is configured as analog input.

I/O PORTS (Cont'd)
Table 11. I/O Port Register Map and Reset Values

| Address <br> (Hex.) | Register Label | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0000h | PADR <br> Reset Value | $\begin{gathered} \text { D7 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D6 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D5 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D4 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D3 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D2 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D1 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DO } \\ 0 \end{gathered}$ |
| 0001h | PADDR <br> Reset Value | $\begin{gathered} \text { DD7 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD6 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD5 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD4 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD3 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD2 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD1 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DDO } \\ 0 \end{gathered}$ |
| 0002h | PCDR <br> Reset Value | $\begin{gathered} \text { D7 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D6 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D5 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D4 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D3 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D2 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D1 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D0 } \\ 0 \end{gathered}$ |
| 0003h | PCDDR <br> Reset Value | $\begin{gathered} \text { DD7 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD6 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD5 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD4 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD3 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD2 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD1 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DDO } \\ 0 \end{gathered}$ |
| 0004h | PDDR <br> Reset Value | $\begin{gathered} \text { D7 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D6 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D5 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D4 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D3 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D2 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D1 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D0 } \\ 0 \end{gathered}$ |
| 0005h | PDDDR <br> Reset Value | $\begin{gathered} \text { DD7 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD6 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD5 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD4 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD3 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD2 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD1 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DDO } \\ 0 \end{gathered}$ |
| 0006h | PBDR <br> Reset Value | $\begin{gathered} \text { D7 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D6 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D5 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D4 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D3 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D2 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { D1 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DO } \\ 0 \end{gathered}$ |
| 0007h | PBDDR <br> Reset Value | $\begin{gathered} \text { DD7 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD6 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD5 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD4 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD3 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD2 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DD1 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { DDO } \\ 0 \end{gathered}$ |
| 0008h | PBOR <br> Reset Value | $\begin{gathered} \text { AD7 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { AD6 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { AD5 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { AD4 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { AD3 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { AD2 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { AD1 } \\ 0 \end{gathered}$ | $\begin{gathered} \text { ADO } \\ 0 \end{gathered}$ |

### 4.2 WATCHDOG TIMER (WDG)

### 4.2.1 Introduction

The Watchdog timer is used to detect the occurrence of a software fault, usually generated by external interference or by unforeseen logical conditions, which causes the application program to abandon its normal sequence. The Watchdog circuit generates an MCU reset on expiry of a programmed time period, unless the program refreshes the counter's contents before the T6 bit becomes cleared.

### 4.2.2 Main Features

- Programmable timer (64 increments of 49,152 CPU cycles)
- Programmable reset
- Reset (if watchdog activated) after a HALT instruction or when the T6 bit reaches zero

Figure 16. Watchdog Block Diagram


## WATCHDOG TIMER (Cont'd)

### 4.2.3 Functional Description

The counter value stored in the CR register (bits T6:T0), is decremented every 49,152 machine cycles, and the length of the timeout period can be programmed by the user in 64 increments.
If the watchdog is activated (the WDGA bit is set) and when the 7-bit timer (bits T6:T0) rolls over from 40h to 3Fh (T6 become cleared), it initiates a reset cycle pulling low the reset pin for typically 500ns.
The application program must write in the CR register at regular intervals during normal operation to prevent an MCU reset. The value to stored in the CR register must be between FFh and C0h (see Table 1):

- The WDGA bit is set (watchdog enabled)
- The T6 bit is set to prevent generating an immediate reset
- The T5:T0 bits contain the number of increments which represents the time delay before the watchdog produces a reset.

Table 12. Watchdog Timing ( $\mathrm{f}_{\mathrm{CPU}}=8 \mathrm{MHz}$ )

|  | CR Register <br> initial value | WDG timeout period <br> $(\mathbf{m s})$ |
| :---: | :---: | :---: |
| Max | FFh | 393.216 |
| Min | COh | 6.144 |

The T6 bit can be used to generate a software reset (the WDGA bit is set and the T6 bit is cleared). If the watchdog is activated, the HALT instruction will generate a Reset.

### 4.2.4 Register Description CONTROL REGISTER (CR)

Read/Write

Reset Value: 01111111 (7Fh)

| 7 | 0 |  |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| WDGA | T6 | T5 | T4 | T3 | T2 | T1 | T0 |

Bit 7=WDGA Activation bit.
This bit is set by software and only cleared by hardware after a reset. When WDGA $=1$, the watchdog can generate a reset.
0: Watchdog disabled
1: Watchdog enabled

Bit 6:0 = T[6:0] 7-bit timer (MSB to $L S B$ ).
These bits contain the decremented value. A reset is produced when it rolls over from 40h to 3Fh (T6 become cleared) if $\mathrm{WDGA}=1$.

Notes: Following a reset, the watchdog is disabled. Once activated it cannot be disabled, except by a reset.

## Table 13. WDG Register Map

| Address <br> (Hex.) | Register <br> Name | $\mathbf{7}$ | $\mathbf{6}$ | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| OC <br> Reset Value | CR | WDGA <br> 0 | T6 <br> 1 | T5 <br> 1 | T 4 <br> 1 | T 3 <br> 1 | T 2 <br> 1 | T 1 <br> 1 | T0 <br> 1 |

### 4.3 16-BIT TIMER

### 4.3.1 Introduction

The timer consists of a 16 -bit free-running counter driven by a programmable prescaler.
It may be used for a variety of purposes, including pulse length measurement of up to two input signals (input capture) or generation of up to two output waveforms (output compare and PWM).
Pulse lengths and waveform periods can be modulated from a few microseconds to several milliseconds using the timer prescaler and the CPU clock prescaler.

### 4.3.2 Main Features

- Programmable prescaler: tpu $^{\text {divided by } 2,4 \text { or } 8 .}$
- Overflow status flag and maskable interrupt
- External clock input (must be at least 4 times slower than the CPUclock speed) with the choice of active edge
- Output compare functions with
- 2 dedicated 16-bit registers
- 2 dedicated programmable signals
- 2 dedicated status flags
- 1 dedicated maskable interrupt
- Input capture functions with
- 2 dedicated 16-bit registers
- 2 dedicated active edge selection signals
- 2 dedicated status flags
- 1 dedicated maskable interrupt
- Pulse width modulation mode (PWM)
- One pulse mode
- 5 alternate functions on I/O ports*

The Block Diagram is shown inFigure 1.
*Note: Some external pins are not available on all devices. Refer to the device pin out description.
When reading an input signal which is not available on an external pin, the value will always be ' 1 '.

### 4.3.3 Functional Description

### 4.3.3.1 Counter

The principal block of the Programmable Timer is a 16 -bit free running increasing counter and its associated 16 -bit registers:

## Counter Registers

- Counter High Register (CHR) is the most significant byte (MSB).
- Counter Low Register (CLR) is the least significant byte (LSB).
Alternate Counter Registers
- Alternate Counter High Register (ACHR) is the most significant byte (MSB).
- Alternate Counter Low Register (ACLR) is the least significant byte (LSB).
These two read-only 16 -bit registers contain the same value but with the difference that reading the ACLR register does not clear the TOF bit (overflow flag), (see note at the end of paragraph titled 16-bit read sequence).
Writing in the CLR register or ACLR register resets the free running counter to the FFFCh value.

The timer clock depends on the clock control bits of the CR2 register, as illustrated inTable 1. The value in the counter register repeats every 131.072, 262.144 or 524.288 internal processor clock cycles depending on the CC1 and CC0 bits.

## 16-BIT TIMER (Cont'd)

Figure 17. Timer Block Diagram


## 16-BIT TIMER (Cont'd)

16-bit read sequence: (from either the Counter Register or the Alternate Counter Register).


## Sequence completed

The user must read the MSB first, then the LSB value is buffered automatically.
This buffered value remains unchanged until the 16 -bit read sequence is completed, even if the user reads the MSB several times.
After a complete reading sequence, if only the CLR register or ACLR register are read, they return the LSB of the count value at the time of the read.
An overflow occurs when the counter rolls over from FFFFh to 0000h then:

- The TOF bit of the SR register is set.
- A timer interrupt is generated if:
- TOIE bit of the CR1 register is set and
- I bit of the CC register is cleared.

If one of these conditions is false, the interrupt remains pending to be issued as soon as they are both true.

Clearing the overflow interrupt request is done in two steps:

1. Reading the SR register while the TOF bit is set.
2. An access (read or write) to the CLR register.

Notes: The TOF bit is not cleared by accesses to ACLR register. This feature allows simultaneous use of the overflow function and reads of the free running counter at random times (for example, to measure elapsed time) without the risk of clearing the TOF bit erroneously.
The timer is not affected by WAIT mode.
In HALT mode, the counter stops counting until the mode is exited. Counting then resumes from the previous count (MCU awakened by an interrupt) or from the reset count (MCU awakened by a Reset).

### 4.3.3.2 External Clock

The external clock (where available) is selected if $C C 0=1$ and CC1=1 in CR2 register.
The status of the EXEDG bit determines the type of level transition on the external clock pin EXTCLK that will trigger the free running counter.
The counter is synchronised with the falling edge of the internal CPU clock.

At least four falling edges of the CPU clock must occur between two consecutive active edges of the external clock; thus the external clock frequency must be less than a quarter of the CPU clock frequency.

## 16-BIT TIMER (Cont'd)

Figure 18. Counter Timing Diagram, internal clock divided by 2


Figure 19. Counter Timing Diagram, internal clock divided by 4


Figure 20. Counter Timing Diagram, internal clock divided by 8


## 16-BIT TIMER (Cont'd)

### 4.3.3.3 Input Capture

In this section, the index, $i$, may be 1 or 2.
The two input capture 16 -bit registers (IC1R and IC2R) are used to latch the value of the free running counter after a transition detected by the ICAPi pin (see figure 5).


ICi Rregister is a read-only register.
The active transition is software programmable through the IEDGi bit of the Control Register (CRi).

Timing resolution is one count of the free running counter: ( ${ }^{(\mathrm{CPU}} /(\mathrm{CC} 1 . \mathrm{CC0})$ ).

## Procedure

To use the input capture function select the following in the CR2 register:

- Select the timer clock (CC1-CC0) (seeTable 1).
- Select the edge of the active transition on the ICAP2 pin with the IEDG2 bit.
And select the following in the CR1 register:
- Set the ICIE bit to generate an interrupt after an input capture.
- Select the edge of the active transition on the ICAP1 pin with the IEDG1 bit.
When an input capture occurs:
- ICFi bit is set.
- The ICiR register contains the value of the free running counter on the active transition on the ICAPi pin (see Figure 6).
- A timer interrupt is generated if the ICIE bit is set and the I bit is cleared in the CC register. Otherwise, the interrupt remains pending until both conditions become true.
Clearing the Input Capture interrupt request is done in two steps:

1. Reading the SR register while the ICF bit is set.
2. An access (read or write) to the IOLR register.

Note: After reading the ICiHR register, transfer of input capture data is inhibited until the IOLR register is also read.
The ICiR register always contains the free running counter value which corresponds to the most recent input capture.
During HALT mode, if at least one valid input capture edge occurs on the ICAPi pin, the input capture detection circuitry is armed. This does not set any timer flags, and does not "wake-up" the MCU. If the MCU is awoken by an interrupt, the input capture flag will become active, and data corresponding to the first valid edge during HALT mode will be present.

## 16-BIT TIMER (Cont'd)

Figure 21. Input Capture Block Diagram


Figure 22. Input Capture Timing Diagram


Note: Active edge is rising edge.

## 16-BIT TIMER (Cont'd)

### 4.3.3.4 Output Compare

In this section, the index, $i$, may be 1 or 2 .
This function can be used to control an output waveform or indicating when a period of time has elapsed.
When a match is found between the Output Compare register and the free running counter, the output compare function:

- Assigns pins with a programmable value if the OCIE bit is set
- Sets a flag in the status register
- Generates an interrupt if enabled

Two 16-bit registers Output Compare Register 1 (OC1R) and Output Compare Register 2 (OC2R) contain the value to be compared to the free running counter each timer clock cycle.


These registers are readable and writable and are not affected by the timer hardware. A reset event changes the OCiR value to 8000 h .
Timing resolution is one count of the free running counter: (fCPU/(CC1.CC0)).

## Procedure

To use the output compare function, select the following in the CR2 register:

- Set the OCIE bit if an output is needed then the OCMPi in is dedicated to the output compare $i$ function.
- Select the timer clock (CC1-CC0) (seeTable 1).

And select the following in the CR1 register:

- Select the OLVLibit to applied to the OCMPipins after the match occurs.
- Set the OCIE bit to generate an interrupt if it is needed.
When match is found:
- OCFi bit is set.
- The OCMPi pin takes OLVLi bit value (OCMPi pin latch is forced low during reset and stays low until valid compares change it to a high level).
- A timer interrupt is generated if the OCIE bit is set in the CR2 register and the I bit is cleared in the CC register (CC).

Clearing the output compare interrupt request is done by:
3. Reading the SR register while the OCF bit is set.
4. An access (read or write) to the OCLR register.

Note: After a processor write cycle to the OOHR register, the output compare function is inhibited until the OCLR register is also written.
If the OCIE bit is not set, the OCMPi pin is a general I/O port and the OLVLi bit will not appear when match is found but an interrupt could be generated if the OCIE bit is set.
The value in the 16 -bit OCR register and the OLVi bit should be changed after each successful comparison in order to control an output waveform or establish a new elapsed timeout.
The OCiR register value required for a specific timing application can be calculated using the following formula:

$$
\Delta \mathrm{OCRR}=\frac{\Delta t * f_{\mathrm{CPU}}}{\mathrm{t}_{\text {PRESC }}}
$$

Where:
$\Delta \mathrm{t} \quad=$ Desired output compare period (in seconds)
$\mathrm{f}_{\mathrm{CPU}}=$ Internal clock frequency
$t_{\text {PRESC }}=$ Timer clock prescaler (CC1-CC0 bits, see Table 1)
The following procedure is recommended to prevent the OCFi bit from being set between the time it is read and the write to the OOR register:

- Write to the OCHR register (further compares are inhibited).
- Read the SR register (first step of the clearance of the OCFi bit, which may be already set).
- Write to the OCILR register (enables the output compare function and clears the OCF bit).


## 16-BIT TIMER (Cont'd)

Figure 23. Output Compare Block Diagram


Figure 24. Output Compare Timing Diagram, Internal Clock Divided by 2


## 16-BIT TIMER (Cont'd)

### 4.3.3.5 Forced Compare Mode

In this section $i$ may represent 1 or 2 .
The following bits of the CR1 register are used:


When the FOLVi bit is set, the OLVLi bit is copied to the OCMPi in. The OLVi bit has to be toggled in order to toggle the OCMPi pin when it is enabled (OCIE bit=1).
The OCFi bit is not set, and thus no interrupt request is generated.

### 4.3.3.6 One Pulse Mode

One Pulse mode enables the generation of a pulse when an external event occurs. This mode is selected via the OPM bit in the CR2 register.
The one pulse mode uses the Input Capture1 function and the Output Compare1 function.

## Procedure

To use one pulse mode:

1. Load the OC1R register with the value corresponding to the length of the pulse (see the formula in Section 0.1.3.7).
2. Select the following in the the CR1 register:

- Using the OLVL1 bit, select the level to be applied to the OCMP1 pin after the pulse.
- Using the OLVL2 bit, select the level to be applied to the OCMP1 pin during the pulse.
- Select the edge of the active transition on the ICAP1 pin with the IEDG1 bit

3. Select the following in the CR2 register:

- Set the OC1E bit, the OCMP1 pin is then dedicated to the Output Compare 1 function.
- Set the OPM bit.
- Select the timer clock CC1-CC0 (seeTable 1).


Then, on a valid event on the ICAP1 pin, the counter is initialized to FFFCh and OLVL2 bit is loaded on the OCMP1 pin. When the value of the counter is equal to the value of the contents of the OC1R register, the OLVL1 bit is output on the OCMP1 pin, (See Figure 9).

Note: The OCF1 bit cannot be set by hardware in one pulse mode but the OCF2 bit can generate an Output Compare interrupt.
The ICF1 bit is set when an active edge occurs and can generate an interrupt if the ICIE bit is set. When the Pulse Width Modulation (PWM) and One Pulse Mode (OPM) bits are both set, the PWM mode is the only active one.

Figure 25. One Pulse Mode Timing


## 16-BIT TIMER (Cont'd)

### 4.3.3.7 Pulse Width Modulation Mode

Pulse Width Modulation mode enables the generation of a signal with a frequency and pulse length determined by the value of the OC1R and OC2R registers.
The pulse width modulation mode uses the complete Output Compare 1 function plus the OC2R register.

## Procedure

To use pulse width modulation mode:

1. Load the OC2R register with the value corresponding to the period of the signal.
2. Load the OC1R register with the value corresponding to the length of the pulse if (OLVL1=0 and OLVL2=1).
3. Select the following in the CR1 register:

- Using the OLVL1 bit, select the level to be applied to the OCMP1 pin after a successful comparison with OC1R register.
- Using the OLVL2 bit, select the level to be applied to the OCMP1 pin after a successful comparison with OC2R register.

4. Select the following in the CR2 register:

- Set OC1E bit: the OCMP1 pin is then dedicated to the output compare 1 function.
- Set the PWM bit.
- Select the timer clock (CC1-CC0) (seeTable 1).

If OLVL1=1 and OLVL2=0 the length of the pulse is the difference between the OC2R and OC1R registers.
The OCiR register value required for a specific timing application can be calculated using the following formula:

$$
\text { OCiR Value }=\frac{t * f_{\mathrm{CPU}}}{\mathrm{t}_{\mathrm{PRESC}}}-5
$$

Where:
$-t=$ Desired output compare period (seconds)
$-\mathrm{f}_{\mathrm{CPU}}=$ Internal clock frequency (see Miscellaneous register)
$-t_{\text {PRESC }}=$ Timer clock prescaler (CC1-CC0 bits, see Table 1)
The Output Compare 2 event causes the counter to be initialized to FFFCh (SeeFigure 10).


Note: After a write instruction to the OCHR register, the output compare function is inhibited until the OCiLR register is also written.
The ICF1 bit is set by hardware when the counter reaches the OC2R value and can produce a timer interrupt if the ICIE bit is set and the I bit is cleared. Therefore the Input Capture 1 function is inhibited but the Input Capture 2 is available.
The OCF1 and OCF2 bits cannot be set by hardware in PWM mode therefore the Output Compare interrupt is inhibited.
When the Pulse Width Modulation (PWM) and One Pulse Mode (OPM) bits are both set, the PWM mode is the only active one.

Figure 26. Pulse Width Modulation Mode Timing


## 16-BIT TIMER (Cont'd)

### 4.3.4 Register Description

Each Timer is associated with three control and status registers, and with six pairs of data registers (16-bit values) relating to the two input captures, the two output compares, the counter and the alternate counter.

## CONTROL REGISTER 1 (CR1)

Read/Write
Reset Value: 00000000 (00h)


Bit 7 = ICIE Input Capture Interrupt Enable.
0 : Interrupt is inhibited.
1: A timer interrupt is generated whenever the ICF1 or ICF2 bit of the SR register is set.

Bit $6=$ OCIE Output Compare Interrupt Enable. 0 : Interrupt is inhibited.
1: A timer interrupt is generated whenever the OCF1 or OCF2 bit of the SR register is set.

Bit 5 = TOIE Timer Overflow Interrupt Enable.
0 : Interrupt is inhibited.
1: A timer interrupt is enabled whenever the TOF bit of the SR register is set.

Bit 4 = FOLV2 Forced Output Compare 2.
This bit is set and cleared by software.
0 : No effect on the OCMP2 pin.
1:Forces the OLVL2 bit to be copied to the OCMP2 pin.

Bit 3 = FOLV1 Forced Output Compare 1. This bit is set and cleared by software.
0 : No effect on the OCMP1 pin.
1: Forces OLVL1 to be copied to the OCMP1 pin.

Bit 2 = OLVL2 Output Level 2.
This bit is copied to the OCMP2 pin whenever a successful comparison occurs with the OC2R register and OCxE is set in the CR2 register. This value is copied to the OCMP1 pin in One Pulse Mode and Pulse Width Modulation mode.

Bit 1 = IEDG1 Input Edge 1.
This bit determines which type of level transition on the ICAP1 pin will trigger the capture.
0 : A falling edge triggers the capture.
1: A rising edge triggers the capture.

Bit $0=$ OLVL1 Output Level 1.
The OLVL1 bit is copied to the OCMP1 pin whenever a successful comparison occurs with the OC1R register and the OC1E bit is set in the CR2 register.

## 16-BIT TIMER (Cont'd)

CONTROL REGISTER 2 (CR2)
Read/Write
Reset Value: 00000000 (00h)

$$
\begin{array}{ll}
7 & 0
\end{array}
$$



Bit $7=$ OC1E Output Compare 1 Enable.
0 : Output Compare 1 function is enabled, but the OCMP1 pin is a general I/O.
1: Output Compare 1 function is enabled, the OCMP1 pin is dedicated to the Output Compare 1 capability of the timer.

Bit $6=$ OC2E Output Compare 2 Enable.
0 : Output Compare 2 function is enabled, but the OCMP2 pin is a general I/O.
1: Output Compare 2 function is enabled, the OCMP2 pin is dedicated to the Output Compare 2 capability of the timer.

Bit 5 = OPM One Pulse Mode.
0 : One Pulse Mode is not active.
1: One Pulse Mode is active, the ICAP1 pin can be used to trigger one pulse on the OCMP1 pin; the active transition is given by the IEDG1 bit. The length of the generated pulse depends on the contents of the OC1R register.

Bit 4 = PWM Pulse Width Modulation.
0 : PWM mode is not active.
1: PWM mode is active, the OCMP1 pin outputs a programmable cyclic signal; the length of the pulse depends on the value of OC1R register; the period depends on the value of OC2R register.

Bit 3, 2 = CC1-CC0 Clock Control.
The value of the timer clock depends on these bits:
Table 14. Clock Control Bits

| Timer Clock | CC1 | CC0 |
| :---: | :---: | :---: |
| $\mathrm{f}_{\mathrm{CPU}} / 4$ | 0 | 0 |
| $\mathrm{f}_{\mathrm{CPU}} / 2$ | 0 | 1 |
| $\mathrm{f}_{\mathrm{CPU}} / 8$ | 1 | 0 |
| External Clock (where <br> available) | 1 | 1 |

Bit 1 = IEDG2 Input Edge 2.
This bit determines which type of level transition on the ICAP2 pin will trigger the capture.
0 : A falling edge triggers the capture.
1: A rising edge triggers the capture.
Bit 0 = EXEDG External Clock Edge.
This bit determines which type of level transition on the external clock pin EXTCLK will trigger the free running counter.
0 : A falling edge triggers the free running counter. 1: A rising edge triggers the free running counter.

## 16-BIT TIMER (Cont'd) <br> STATUS REGISTER (SR)

Read Only
Reset Value: 00000000 (00h)
The three least significant bits are not used.


Bit 7 = ICF1 Input Capture Flag 1.
0 : No input capture (reset value).
1: An input capture has occurred or the counter has reached the OC2R value in PWM mode. To clear this bit, first read the SR register, then read or write the low byte of the IC1R (IC1LR) register.

Bit $6=$ OCF1 Output Compare Flag 1.
0 : No match (reset value).
1: The content of the free running counter has matched the content of the OC1R register. To clear this bit, first read the SR register, then read or write the low byte of the OC1R (OC1LR) register.

Bit $5=$ TOF Timer Overflow.
0 : No timer overflow (reset value).
1:The free running counter rolled over from FFFFh to 0000 h . To clear this bit, first read the SR register, then read or write the low byte of the CR (CLR) register.
Note: Reading or writing the ACLR register does not clear TOF.

Bit 4 = ICF2 Input Capture Flag 2.
0 : No input capture (reset value).
1: An input capture has occurred.To clear this bit, first read the SR register, then read or write the low byte of the IC2R (IC2LR) register.

Bit 3 = OCF2 Output Compare Flag 2.
0 : No match (reset value).
1: The content of the free running counter has matched the content of the OC2R register. To clear this bit, first read the SR register, then read or write the low byte of the OC2R (OC2LR) register.

Bit 2-0 = Reserved, forced by hardware to 0 .

## INPUT CAPTURE 1 HIGH REGISTER (IC1HR)

Read Only
Reset Value: Undefined
This is an 8 -bit read only register that contains the high part of the counter value (transferred by the input capture 1 event).

7 0


## INPUT CAPTURE 1 LOW REGISTER (IC1LR)

Read Only
Reset Value: Undefined
This is an 8 -bit read only register that contains the low part of the counter value (transferred by the input capture 1 event).


Read/Write
Reset Value: 10000000 (80h)
This is an 8 -bit register that contains the high part of the value to be compared to the CHR register.

7
0


OUTPUT COMPARE 1 LOW REGISTER (OC1LR)

## Read/Write

Reset Value: 00000000 (00h)
This is an 8-bit register that contains the low part of the value to be compared to the CLR register.

7
0


16-BIT TIMER (Cont'd)
OUTPUT COMPARE 2 HIGH REGISTER ALTERNATE COUNTER HIGH REGISTER (OC2HR)
Read/Write
Reset Value: 10000000 (80h)
This is an 8-bit register that contains the high part of the value to be compared to the CHR register.


OUTPUT COMPARE 2 LOW REGISTER (OC2LR)
Read/Write
Reset Value: 00000000 (00h)
This is an 8-bit register that contains the low part of the value to be compared to the CLR register.

7 0


## COUNTER HIGH REGISTER (CHR)

Read Only
Reset Value: 11111111 (FFh)
This is an 8-bit register that contains the high part of the counter value.


## COUNTER LOW REGISTER (CLR)

Read Only
Reset Value: 11111100 (FCh)
This is an 8-bit register that contains the low part of the counter value. A write to this register resets the counter. An access to this register after accessing the SR register clears the TOF bit.

(ACHR)
Read Only
Reset Value: 11111111 (FFh)
This is an 8-bit register that contains the high part of the counter value.

7
0


## ALTERNATE COUNTER LOW REGISTER (ACLR)

Read Only
Reset Value: 11111100 (FCh)
This is an 8-bit register that contains the low part of the counter value. A write to this register resets the counter. An access to this register after an access to SR register does not clear the TOF bit in SR register.

7 0


## INPUT CAPTURE 2 HIGH REGISTER (IC2HR)

Read Only
Reset Value: Undefined
This is an 8-bit read only register that contains the high part of the counter value (transferred by the Input Capture 2 event).

$$
\begin{array}{ll}
7 & 0
\end{array}
$$



## INPUT CAPTURE 2 LOW REGISTER (IC2LR)

Read Only
Reset Value: Undefined
This is an 8-bit read only register that contains the low part of the counter value (transferred by the Input Capture 2 event).

7 0


Table 15. 16-Bit Timer Register Map

| Address <br> (Hex.) | Register Name | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 11 | CR2 | OC1E | OC2E | OPM | PWM | CC1 | CCO | IEDG2 | EXEDG |
| 12 | CR1 | ICIE | OCIE | TOIE | FOLV2 | FOLV1 | OLVL2 | IEDG1 | OLVL1 |
| 13 | SR | ICF1 | OCF1 | TOF | ICF2 | OCF2 | 0 | 0 | 0 |
| 14 | IC1HR | MSB |  |  |  |  |  |  | LSB |
| 15 | IC1LR | MSB |  |  |  |  |  |  | LSB |
| 16 | OC1HR | MSB |  |  |  |  |  |  | LSB |
| 17 | OC1LR | MSB |  |  |  |  |  |  | LSB |
| 18 | CHR | MSB |  |  |  |  |  |  | LSB |
| 19 | CLR | MSB |  |  |  |  |  |  | LSB |
| 1A | ACHR | MSB |  |  |  |  |  |  | LSB |
| 1B | ACLR | MSB |  |  |  |  |  |  | LSB |
| 1 C | IC2HR | MSB |  |  |  |  |  |  | LSB |
| 1D | IC2LR | MSB |  |  |  |  |  |  | LSB |
| 1E | OC2HR | MSB |  |  |  |  |  |  | LSB |
| 1F | OC2LR | MSB |  |  |  |  |  |  | LSB |
| 43 | CONFIG <br> Reset Value | $0$ | $\bar{o}$ | $\bar{o}$ | $\overline{0}$ | $\overline{1}$ | $\begin{gathered} \text { ICAP } \\ 0 \end{gathered}$ | $\bar{o}$ | 0 |

Warning: Write OCh in the CONFIG register to use the ICAP1 and ICAP2 pins (set bits 3 and 2).

### 4.4 I C BUS INTERFACE (I2C)

### 4.4.1 Introduction

The $I^{2} C$ Bus Interface serves as an interface between the microcontroller and the serial ${ }^{1} \mathrm{~F}$ bus. It provides both multimaster and slave functions, and controls all ${ }^{2} \mathrm{C}$ bus-specific sequencing, protocol, arbitration and timing. It supports fast IC mode ( 400 kHz ).

### 4.4.2 Main Features

- Parallel bus $/{ }^{2} \mathrm{C}$ protocol converter
- Multi-Master capability
- Interrupt generation
- Standard ${ }^{2} \mathrm{C}$ mode/Fast $\mathrm{P}^{2} \mathrm{C}$ mode
- 7-bit Addressing
- $I^{2}$ C Slave Mode
- Start bit detection flag
- Detection of misplaced Start or Stop condition
- Transfer problem detection
- Address Matched detection
- Default Address detection
- End of byte transmission flag
- Transmitter/Receiver flag
- Stop bit Detection
- $\mathrm{I}^{2} \mathrm{C}$ Master Mode
- $I^{2} \mathrm{C}$ bus busy flag
- Arbitration lost flag
- End of byte transmission flag
- Transmitter/Receiver flag
- Clock generation


### 4.4.3 General Description

In addition to receiving and transmitting data, this interface converts it from serial to parallel format and vice versa, using either an interrupt or polled
handshake. The interrupts are enabled or disabled by software. The interface is connected to the ${ }^{\text {F C }}$ bus by a data pin (SDAI) and by a clock pin (SCLI). It can be connected both with a standard ${ }^{\mathrm{F} C}$ bus and a Fast ${ }^{1} \mathrm{C}$ bus. This selection is made by software.

## Mode Selection

The interface can operate in the four following modes:

- Slave transmitter/receiver
- Master transmitter/receiver

By default, it operates in slave mode.
The interface automatically switches from slave to master after it generates a START condition and from master to slave in case of arbitration loss or a STOP generation, this allows Multi-Master capability.

## Communication Flow

In Master mode, it initiates a data transfer and generates the clock signal. A serial data transfer always begins with a start condition and ends with a stop condition. Both start and stop conditions are generated in master mode by software.
In Slave mode, the interface is capable of recognising its own address (7-bit), and the General Call address. The General Call address detection may be enabled or disabled by software.
Data and addresses are transferred as 8 -bit bytes, MSB first. The first byte following the start condition is the address byte; it is always transmitted in Master mode.
A 9th clock pulse follows the 8 clock cycles of a byte transfer, during which the receiver must send an acknowledge bit to the transmitter. Refer toFigure 1 .

## Figure $1.1^{2} \mathrm{C}$ BUS Protocol



## I C BUS INTERFACE (Cont'd)

Acknowledge may be enabled and disabled by software.
The $I^{2} C$ interface address and/or general call address can be selected by software.
The speed of the RC interface may be selected between Standard $(0-100 \mathrm{KHz})$ and Fast FC ( $100-$ 400 KHz ).

## SDA/SCL Line Control

Transmitter mode: the interface holds the clock line low before transmission to wait for the microcontroller to write the byte in the Data Register.
Receiver mode: the interface holds the clock line low after reception to wait for the microcontroller to read the byte in the Data Register.

The SCL frequency ( $\mathrm{F}_{\mathrm{SCL}}$ ) is controlled by a programmable clock divider which depends on the $\mathrm{I}^{2} \mathrm{C}$ bus mode.
When the $I^{2} C$ cell is enabled, the SDA and SCL ports must be configured as floating open-drain output or floating input. In this case, the value of the external pull-up resistance used depends on the application.
When the ${ }^{2}$ C cell is disabled, the SDA and SCL ports revert to being standard I/O port pins.

Figure 2. $I^{2} \mathrm{C}$ Interface Block Diagram


## I C BUS INTERFACE (Cont'd)

### 4.4.4 Functional Description

Refer to the CR, SR1 and SR2 registers inSection 0.1.5. for the bit definitions.

By default the ${ }^{2} C$ interface operates in Slave mode ( $M / S L$ bit is cleared) except when it initiates a transmit or receive sequence.

### 4.4.4.1 Slave Mode

As soon as a start condition is detected, the address is received from the SDA line and sent to the shift register; then it is compared with the address of the interface or the General Call address (if selected by software).
Address not matched: the interface ignores it and waits for another Start condition.
Address matched the interface generates in sequence:

- Acknowledge pulse if the ACK bit is set.
- EVF and ADSL bits are set with an interrupt if the ITE bit is set.
Then the interface waits for a read of the SR1 register, holding the SCL line low (see Figure 3 Transfer sequencing EV1).
Next, read the DR register to determine from the least significant bit if the slave must enter Receiver or Transmitter mode.


## Slave Receiver

Following the address reception and after SR1 register has been read, the slave receives bytes from the SDA line into theDR registervia the internal shift register. After each byte the interface generates in sequence:

- Acknowledge pulse if the ACK bit is set
- EVF and BTF bits are set with an interrupt if the ITE bit is set.
Then the interface waits for a read of the SR1 register followed by a read of the DR register, holding the SCL line low (see Figure 3 Transfer sequencing EV2).


## Slave Transmitter

Following the address reception and after SR1 register has been read, the slave sends bytes from the DR register to the SDA line via the internal shift register.

The slave waits for a read of the SR1 register followed by a write in the DR register, holding the SCL line low (see Figure 3 Transfer sequencing EV3).
When the acknowledge pulse is received:

- The EVF and BTF bits are set by hardware with an interrupt if the ITE bit is set.


## Closing slave communication

After the last data byte is transferred a Stop Condition is generated by the master. The interface detects this condition and sets:

- EVF and STOPF bits with an interrupt if the ITE bit is set.
Then the interface waits for a read of the SR2 register (see Figure 3 Transfer sequencing EV4).


## Error Cases

- BERR: Detection of a Stop or a Start condition during a byte transfer. In this case, the EVF and the BERR bits are set with an interrupt if the ITE bit is set.
If it is a Stop then the interface discards the data, released the lines and waits for another Start condition.
If it is a Start then the interface discards the data and waits for the next slave address on the bus.
- AF: Detection of a non-acknowledge bit. In this case, the EVF and AF bits are set with an interrupt if the ITE bit is set.
Note: In both cases, SCL line is not held low; however, SDA line can remain low due to possible "0» bits transmitted last. It is then necessary to release both lines by software.


## How to release the SDA / SCL lines

Set and subsequently clear the STOP bit while BTF is set. The SDA/SCL lines are released after the transfer of the current byte.

## I C BUS INTERFACE (Cont'd)

### 4.4.4.2 Master Mode

To switch from default Slave mode to Master mode a Start condition generation is needed.

## Start condition and Transmit Slave address

Setting the START bit while the BUSY bit is cleared causes the interface to switch to Master mode (M/SL bit set) and generates a Start condition.
Once the Start condition is sent:

- The EVF and SB bits are set by hardware with an interrupt if the ITE bit is set.
Then the master waits for a read of the SR1 register followed by a write in the DR register with the Slave address byte, holding the SCL line Iow (see Figure 3 Transfer sequencing EV5).

Then the slave address byte is sent to the SDA line via the internal shift register.
After completion of this transfer (and acknowledge from the slave if the ACK bit is set):

- The EVF bit is set by hardware with interrupt generation if the ITE bit is set.
Then the master waits for a read of the SR1 register followed by a write in the CR register (for example set PE bit), holding the SCL line low(see Figure 3 Transfer sequencing EV6).

Next the master must enter Receiver or Transmitter mode.

## Master Receiver

Following the address transmission and after SR1 and CR registers have been accessed, themaster receives bytes from the SDA line into theDR register via the internal shift register. After each byte the interface generates in sequence:

- Acknowledge pulse if if the ACK bit is set
- EVF and BTF bits are set by hardware with an interrupt if the ITE bit is set.
Then the interface waits for a read of the SR1 register followed by a read of the DR register, holding the SCL line low (see Figure 3 Transfer sequencing EV7).
To close the communication: before reading the last byte from the DR register, set the STOP bit to generate the Stop condition. The interface goes automatically back to slave mode (M/SL bit cleared).

Note: In order to generate the non-acknowledge pulse after the last received data byte, the ACK bit must be cleared just before reading the second last data byte.

## Master Transmitter

Following the address transmission and after SR1 register has been read, the master sends bytes from the DR register to the SDA line via the internal shift register.
The master waits for a read of the SR1 register followed by a write in the DR register, holding the SCL line low (see Figure 3 Transfer sequencing EV8).
When the acknowledge bit is received, the interface sets:

- EVF and BTF bits with an interrupt if the ITE bit is set.
To close the communication: after writing the last byte to the DR register, set the STOP bit to generate the Stop condition. The interface goes automatically back to slave mode (M/SL bit cleared).


## Error Cases

- BERR: Detection of a Stop or a Start condition during a byte transfer. In this case, the EVF and BERR bits are set by hardware with an interrupt if ITE is set.
- AF: Detection of a non-acknowledge bit. In this case, the EVF and AF bits are set by hardware with an interrupt if the ITE bit is set. To resume, set the START or STOP bit.
- ARLO: Detection of an arbitration lost condition. In this case the ARLO bit is set by hardware (with an interrupt if the ITE bit is set and the interface goes automatically back to slave mode (the M/SL bit is cleared).
Note: In all these cases, the SCL line is not held low; however, the SDA line can remain low due to possible « 0 » bits transmitted last. It is then necessary to release both lines by software.


## I C BUS INTERFACE (Cont'd)

Figure 3. Transfer Sequencing

## Slave receiver:



Slave transmitter:


Master receiver:


Master transmitter:


## Legend:

S=Start, $\mathrm{P}=$ Stop, $\mathrm{A}=$ Acknowledge, NA=Non-acknowledge
$\mathrm{EVx}=$ Event (with interrupt if ITE=1)
$E V 1: E V F=1, A D S L=1$, cleared by reading SR1 register.
EV2: $E V F=1, B T F=1$, cleared by reading $S R 1$ register followed by reading $D R$ register.
EV3: $\mathrm{EVF}=1, \mathrm{BTF}=1$, cleared by reading SR1 register followed by writing DR register.
$\mathrm{EV} 3-1: \mathrm{EVF}=1, \mathrm{AF}=1$, cleared by reading SR1 register.
EV4: $E V F=1, S T O P F=1$, cleared by reading SR2 register.
EV 5 : $\mathrm{EVF}=1, \mathrm{SB}=1$, cleared by reading SR 1 register followed by writing DR register.
EV6: $\mathrm{EVF}=1$, cleared by reading SR1 register followed by writing CR register (for example $\mathrm{PE}=1$ ).
EV7: $\mathrm{EVF}=1, \mathrm{BTF}=1$, cleared by reading SR1 register followed by reading DR register.
EV : $\mathrm{EVF}=1, \mathrm{BTF}=1$, cleared by reading SR1 register followed by writing DR register.
Figure 4. Event Flags and Interrupt Generation

*EVF can also be set by EV6 or an error from the SR2 register.

## I C BUS INTERFACE (Cont'd)

### 4.4.5 Register Description <br> $I^{2} \mathrm{C}$ CONTROL REGISTER (CR)

Read / Write
Reset Value: 00000000 (00h)

## 7

0

| 0 | 0 | PE | ENGC | START | ACK | STOP | ITE |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7:6 = Reserved. Forced to 0 by hardware.
Bit $5=$ PE Peripheral enable.
This bit is set and cleared by software.
0 : Peripheral disabled
1: Master/Slave capability
Notes:

- When $\mathrm{PE}=0$, all the bits of the CR register and the SR register except the Stop bit are reset. All outputs are released while $\mathrm{PE}=0$
- When $\mathrm{PE}=1$, the corresponding $\mathrm{I} / O$ pins are selected by hardware as alternate functions.
- To enable the ${ }^{2} \mathrm{C}$ interface, write the CR register TWICE with $P E=1$ as the first write only activates the interface (only PE is set).


## Bit 4 = ENGC Enable General Call.

This bit is set and cleared by software. It is also cleared by hardware when the interface is disabled ( $\mathrm{PE}=0$ ). The 00 h General Call address is acknowledged ( 01 h ignored).
0 : General Call disabled
1: General Call enabled
Bit 3 = START Generation of a Start condition This bit is set and cleared by software. It is also cleared by hardware when the interface is disabled ( $\mathrm{PE}=0$ ) or when the Start condition is sent (with interrupt generation if ITE=1).

- In master mode:

0 : No start generation
1: Repeated start generation

- In slave mode:

0 : No start generation
1: Start generation when the bus is free

Bit 2 = ACK Acknowledge enable.
This bit is set and cleared by software. It is also cleared by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
0 : No acknowledge returned
1: Acknowledge returned after an address byte or a data byte is received

Bit 1 = STOP Generation of a Stop condition
This bit is set and cleared by software. It is also cleared by hardware in master mode. Note: This bit is not cleared when the interface is disabled ( $\mathrm{PE}=0$ ).

- In master mode:

0 : No stop generation
1: Stop generation after the current byte transfer or after the current Start condition is sent. The STOP bit is cleared by hardware when the Stop condition is sent.

- In slave mode:

0 : No stop generation
1: Release the SCL and SDA lines after the current byte transfer ( $\mathrm{BTF}=1$ ). In this mode the STOP bit has to be cleared by software.

Bit $0=$ ITE Interrupt enable.
This bit is set and cleared by software and cleared by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
0: Interrupts disabled
1: Interrupts enabled
Refer to Figure 4 for the relationship between the events and the interrupt.
SCL is held low when the SB, BTF or ADSL flags or an EV6 event (See Figure 3) is detected.

## $I^{2} \mathrm{C}$ INTERFACE (Cont'd) <br> $I^{2} \mathrm{C}$ STATUS REGISTER 1 (SR1)

Read Only
Reset Value: 00000000 (00h)

| 7 |  |  |  |  |  |  |  |
| :---: | :--- | :--- | :--- | :--- | :--- | :--- | :---: |

Bit $7=$ EVF Event flag.
This bit is set by hardware as soon as an event occurs. It is cleared by software reading SR2 register in case of error event or as described inFigure 3. It is also cleared by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
0 : No event
1: One of the following events has occurred:

- BTF=1 (Byte received or transmitted)
- ADSL=1 (Address matched in Slave mode while ACK=1)
- SB=1 (Start condition generated in Master mode)
- $\mathrm{AF}=1$ (No acknowledge received after byte transmission if $\mathrm{ACK}=1$ )
- STOPF=1 (Stop condition detected in Slave mode)
- ARLO=1 (Arbitration lost in Master mode)
- BERR=1 (Bus error, misplaced Start or Stop condition detected)
- Address byte successfully transmitted in Master mode.

Bit $6=$ Reserved. Forced to 0 by hardware.

## Bit 5 = TRA Transmitter/Receiver.

When BTF is set, TRA $=1$ if a data byte has been transmitted. It is cleared automatically when BTF is cleared. It is also cleared by hardware after detection of Stop condition (STOPF=1), loss of bus arbitration (ARLO=1) or when the interface is disabled ( $\mathrm{PE}=0$ ).
0 : Data byte received (if BTF=1)
1: Data byte transmitted

## Bit 4 = BUSY Bus busy.

This bit is set by hardware on detection of a Start condition and cleared by hardware on detection of a Stop condition. It indicates a communication in progress on the bus. This information is still updated when the interface is disabled ( $\mathrm{PE}=0$ ).
0 : No communication on the bus
1: Communication ongoing on the bus

Bit 3 = BTF Byte transfer finished.
This bit is set by hardware as soon as a byte is correctly received or transmitted with interrupt generation if ITE=1. It is cleared by software reading SR1 register followed by a read or write of DR register. It is also cleared by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).

- Following a byte transmission, this bit is set after reception of the acknowledge clock pulse. In case an address byte is sent, this bit is set only after the EV6 event (SeeFigure 3). BTF is cleared by reading SR1 register followed by writing the next byte in DR register.
- Following a byte reception, this bit is set after transmission of the acknowledge clock pulse if ACK=1. BTF is cleared by reading SR1 register followed by reading the byte from DR register.
The SCL line is held low while BTF=1.
0 : Byte transfer not done
1: Byte transfer succeeded
Bit 2 = ADSL Address matched (Slave mode). This bit is set by hardware as soon as the received slave address matched with the OAR register content or a general call is recognized. An interrupt is generated if $I T E=1$. It is cleared by software reading SR1 register or by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
The SCL line is held low while ADSL=1.
0 : Address mismatched or not received
1: Received address matched


## Bit 1 = M/SL Master/Slave.

This bit is set by hardware as soon as the interface is in Master mode (writing START=1). It is cleared by hardware after detecting a Stop condition on the bus or a loss of arbitration (ARLO=1). It is also cleared when the interface is disabled ( $\mathrm{PE}=0$ ).
0 : Slave mode
1: Master mode

Bit $0=$ SB Start bit (Master mode).
This bit is set by hardware as soon as the Start condition is generated (following a write START=1). An interrupt is generated if ITE=1. It is cleared by software reading SR1 register followed by writing the address byte in DR register.It is also cleared by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
0: No Start condition
1: Start condition generated
$\mathrm{I}^{2} \mathrm{C}$ INTERFACE (Cont'd)
$I^{2} \mathrm{C}$ STATUS REGISTER 2 (SR2)
Read Only
Reset Value: 00000000 (00h)
7

| 0 | 0 | 0 | AF | STOPF | ARLO | BERR | GCAL |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7:5 = Reserved. Forced to 0 by hardware.

## Bit 4 = AF Acknowledge failure.

This bit is set by hardware when no acknowledge is returned. An interrupt is generated if ITE=1. It is cleared by software reading SR2 register or by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
The SCL line is not held low while $A F=1$.
0 : No acknowledge failure
1: Acknowledge failure

Bit 3 = STOPF Stop detection (Slave mode).
This bit is set by hardware when a Stop condition is detected on the bus after an acknowledge (if ACK=1). An interrupt is generated if ITE=1. It is cleared by software reading SR2 register or by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
The SCL line is not held low while STOPF=1.
0 : No Stop condition detected
1: Stop condition detected

Bit 2 = ARLO Arbitration lost.
This bit is set by hardware when the interface los-
es the arbitration of the bus to another master. An interrupt is generated if ITE=1. It is cleared by software reading SR2 register or by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).

After an ARLO event the interface switches back automatically to Slave mode (M/SL=0).
The SCL line is not held low while $A R L O=1$.
0: No arbitration lost detected
1: Arbitration lost detected

Bit 1 = BERR Bus error.
This bit is set by hardware when the interface detects a misplaced Start or Stop condition. An interrupt is generated if ITE=1. It is cleared by software reading SR2 register or by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
The SCL line is not held low while BERR=1.
0 : No misplaced Start or Stop condition
1: Misplaced Start or Stop condition

Bit $0=$ GCAL General Call (Slave mode).
This bit is set by hardware when a general call address is detected on the bus while $\mathrm{ENGC}=1$. It is cleared by hardware detecting a Stop condition (STOPF=1) or when the interface is disabled ( $\mathrm{PE}=0$ ).
0 : No general call address detected on bus
1: general call address detected on bus

## $I^{2} \mathrm{C}$ INTERFACE (Cont'd) <br> $I^{2} \mathrm{C}$ CLOCK CONTROL REGISTER (CCR)

Read / Write
Reset Value: 00000000 (00h)
7
0

| FM/SM | CC6 | CC5 | CC4 | CC3 | CC2 | CC1 | CC0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit $7=$ FM/SM Fast/Standard PCC mode.
This bit is set and cleared by software. It is not cleared when the interface is disabled ( $\mathrm{PE}=0$ ).
0 : Standard ${ }^{2} \mathrm{C}$ mode
1: Fast $l^{2} \mathrm{C}$ mode

Bit 6:0 = CC6-CC0 7-bit clock divider.
These bits select the speed of the bus ( $F_{\text {SCL }}$ ) depending on the FC mode. They are not cleared when the interface is disabled ( $\mathrm{PE}=0$ ).

- Standard mode (FM/SM=0): $\mathrm{F}_{\text {SCL }}<=100 \mathrm{kHz}$

$$
\mathrm{F}_{\mathrm{SCL}}=\mathrm{f}_{\mathrm{CPU}} /(2 \mathrm{x}([\mathrm{CC} 6 . . \mathrm{CC} 0]+2))
$$

- Fast mode (FM/SM=1): $\mathrm{F}_{\mathrm{SCL}}>100 \mathrm{kHz}$

$$
\mathrm{F}_{\mathrm{SCL}}=\mathrm{f}_{\mathrm{CPU}} /(3 \mathrm{x}([\mathrm{CC} 6 . . \mathrm{CC} 0]+2))
$$

Note: The programmed $\mathrm{F}_{\text {SCL }}$ assumes no load on SCL and SDA lines.

## $I^{2} \mathrm{C}$ DATA REGISTER (DR)

Read / Write
Reset Value: 00000000 (00h)

| 7 |  | 0 |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |

Bit 7:0 = D7-D0 8-bit Data Register.
These bits contains the byte to be received or transmitted on the bus.

- Transmitter mode: Byte transmission start automatically when the software writes in the DR register.
- Receiver mode: the first data byte is received automatically in the DR register using the least significant bit of the address.
Then, the next data bytes are received one by one after reading the DR register.


## $I^{2} \mathrm{C}$ OWN ADDRESS REGISTER (OAR)

Read / Write
Reset Value: 00000000 (00h)


| ADD7 | ADD6 | ADD5 | ADD4 | ADD3 | ADD2 | ADD1 | ADD0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7:1 = ADD7-ADD1 Interface address.
These bits define the ${ }^{2} \mathrm{C}$ bus address of the interface. They are not cleared when the interface is disabled ( $\mathrm{PE}=0$ ).

Bit $0=$ ADDO Address direction bit.
This bit is don't care, the interface acknowledges either 0 or 1 . It is not cleared when the interface is disabled ( $\mathrm{PE}=0$ ).
Note: Address 01 h is always ignored.

I2C INTERFACE (Cont'd)
Table 16. $I^{2} \mathrm{C}$ Register Map

| Address <br> (Hex.) | Register Name | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 5F | CR |  |  | PE | ENGC | START | ACK | STOP | ITE |
| 5E | SR1 | EVF |  | TRA | BUSY | BTF | ADSL | M/SL | SB |
| 5D | SR2 |  |  |  | AF | STOPF | ARLO | BERR | GCAL |
| 5C | CCR | FM/SM | CC6 .. CC0 |  |  |  |  |  |  |
| 5B | OAR | ADD7 .. ADD0 |  |  |  |  |  |  |  |
| 59 | DR | DR7 .. DR0 |  |  |  |  |  |  |  |

### 4.5 SERIAL COMMUNICATIONS INTERFACE (SCI)

### 4.5.1 Introduction

The Serial Communications Interface (SCI) offers a flexible means of full-duplex data exchange with external equipment requiring an industry standard NRZ asynchronous serial data format.

### 4.5.2 Main Features

- Full duplex, asynchronous communications
- NRZ standard format (Mark/Space)
- Independently programmable transmit and receive baud rates up to 250K baud.
- Programmable data word length (8 or 9 bits)
- Receive buffer full, Transmit buffer empty and End of Transmission flags
- Two receiver wake-up modes:
- Address bit (MSB)
- Idle line
- Muting function for multiprocessor configurations
- Separate enable bits for Transmitter and Receiver
- Three error detection flags:
- Overrun error
- Noise error
- Frame error
- Five interrupt sources with flags:
- Transmit data register empty
- Transmission complete
- Receive data register full
- Idle line received
- Overrun error detected


### 4.5.3 General Description

The interface is externally connected to another device by two pins (see Figure 1):

- TDO: Transmit Data Output. When the transmitter is disabled, the output pin returns to its I/O port configuration. When the transmitter is enabled and nothing is to be transmitted, the TDO pin is at high level.
- RDI: Receive Data Input is the serial data input. Oversampling techniques are used for data recovery by discriminating between valid incoming data and noise.

Through this pins, serial data is transmitted and received as frames comprising:

- An Idle Line prior to transmission or reception
- A start bit
- A data word (8 or 9 bits) least significant bit first
- A Stop bit indicating that the frame is complete.

SERIAL COMMUNICATIONS INTERFACE(Cont'd)
Figure 5. SCI Block Diagram


## SERIAL COMMUNICATIONS INTERFACE(Cont'd)

### 4.5.4 Functional Description

The block diagram of the Serial Control Interface, is shown in Figure 1. It contains 4 dedicated registers:

- Two control registers (CR1 \& CR2)
- A status register (SR)
- A baud rate register (BRR)

Refer to the register descriptions inSection 0.1.5 for the definitions of each bit.

### 4.5.4.1 Serial Data Format

Word length may be selected as being either 8 or 9 bits by programming the M bit in the CR1 register (see Figure 1).

The TDO pin is in low state during the start bit. The TDO pin is in high state during the stop bit.
An Idle character is interpreted as an entire frame of " 1 "s followed by the start bit of the next frame which contains data.
A Break character is interpreted on receiving " 0 "s for some multiple of the frame period. At the end of the last break frame the transmitter inserts an extra "1" bit to acknowledge the start bit.
Transmission and reception are driven by their own baud rate generator.

Figure 6. Word length programming


## SERIAL COMMUNICATIONS INTERFACE(Cont'd)

### 4.5.4.2 Transmitter

The transmitter can send data words of either 8 or 9 bits depending on the $M$ bit status. When the $M$ bit is set, word length is 9 bits and the 9th bit (the MSB) has to be stored in the T8 bit in the CR1 register.

## Character Transmission

During an SCI transmission, data shifts out least significant bit first on the TDO pin. In this mode, the DR register consists of a buffer (TDR) between the internal bus and the transmit shift register (see Figure 1).

## Procedure

- Select the M bit to define the word length.
- Select the desired baud rate using the BRR register.
- Set the TE bit to assign the TDO pin to the alternate function and to send a idle frame as first transmission.
- Access the SR register and write the data to send in the DR register (this sequence clears the TDRE bit). Repeat this sequence for each data to be transmitted.

Clearing the TDRE bit is always performed by the following software sequence:

1. An access to the SR register
2. A write to the DR register

The TDRE bit is set by hardware and it indicates:

- The TDR register is empty.
- The data transfer is beginning.
- The next data can be written in the DR register without overwriting the previous data.

This flag generates an interrupt if the TIE bit is set and the I bit is cleared in the CC register.
When a transmission is taking place, a write instruction to the DR register stores the data in the TDR register and which is copied in the shift register at the end of the current transmission.
When no transmission is taking place, a write instruction to the DR register places the data directly in the shift register, the data transmission starts and the TDRE bit is immediately set

When a frame transmission is complete (after the stop bit or after the break frame) the TC bit is set and an interrupt is generated if the TCIE is set and the I bit is cleared in the CC register.
Clearing the TC bit is performed by the following software sequence:

1. An access to the SR register
2. A write to the DR register

Note: The TDRE and TC bits are cleared by the same software sequence.

## Break Characters

Setting the SBK bit loads the shift register with a break character. The break frame length depends on the $M$ bit (see Figure 2).

As long as the SBK bit is set, the SCl send break frames to the TDO pin. After clearing this bit by software the SCI insert a logic 1 bit at the end of the last break frame to guarantee the recognition of the start bit of the next frame.

## Idle Characters

Setting the TE bit drives the SCI to send an idle frame before the first data frame.
Clearing and then setting the TE bit during a transmission sends an idle frame after the current word.
Note: Resetting and setting the TE bit causes the data in the TDR register to be lost. Therefore the best time to toggle the TE bit is when the TDRE bit is set i.e. before writing the next byte in the DR.

## SERIAL COMMUNICATIONS INTERFACE(Cont'd)

### 4.5.4.3 Receiver

The SCI can receive data words of either 8 or 9 bits. When the M bit is set, word length is 9 bits and the MSB is stored in the R8 bit in the CR1 register.

## Character reception

During a SCI reception, data shifts in least significant bit first through the RDI pin. In this mode, DR register consists in a buffer (RDR) between the internal bus and the received shift register (see Figure 1).

## Procedure

- Select the M bit to define the word length.
- Select the desired baud rate using the BRR register.
- Set the RE bit, this enables the receiver which begins searching for a start bit.
When a character is received:
- The RDRF bit is set. It indicates that the content of the shift register is transferred to the RDR.
- An interrupt is generated if the RIE bit is set and the I bit is cleared in the CC register.
- The error flags can be set if a frame error, noise or an overrun error has been detected during reception.
Clearing the RDRF bit is performed by the following software sequence done by:

1. An access to the SR register
2. A read to the DR register.

The RDRF bit must be cleared before the end of the reception of the next character to avoid an overrun error.

## Break Character

When a break character is received, the SCI handles it as a framing error.

## Idle Character

When a idle frame is detected, there is the same procedure as a data received character plus an interrupt if the ILIE bit is set and the I bit is cleared in the CC register.

## Overrun Error

An overrun error occurs when a character is received when RDRF has not been reset. Data can not be transferred from the shift register to the TDR register as long as the RDRF bit is not cleared.

When a overrun error occurs:

- The OR bit is set.
- The RDR content will not be lost.
- The shift register will be overwritten.
- An interrupt is generated if the RIE bit is set and the I bit is cleared in the CC register.
The OR bit is reset by an access to the SR register followed by a DR register read operation.


## Noise Error

Oversampling techniques are used for data recovery by discriminating between valid incoming data and noise.
When noise is detected in a frame:

- The NF is set at the rising edge of the RDRF bit.
- Data is transferred from the Shift register to the DR register.
- No interrupt is generated. However this bit rises at the same time as the RDRF bit which itself generates an interrupt.
The NF bit is reset by a SR register read operation followed by a DR register read operation.


## Framing Error

A framing error is detected when:

- The stop bit is not recognized on reception at the expected time, following either a de-synchronization or excessive noise.
- A break is received.

When the framing error is detected:

- the FE bit is set by hardware
- Data is transferred from the Shift register to the DR register.
- No interrupt is generated. However this bit rises at the same time as the RDRF bit which itself generates an interrupt.
The FE bit is reset by a SR register read operation followed by a DR register read operation.


## SERIAL COMMUNICATIONS INTERFACE(Cont'd)

### 4.5.4.4 Baud Rate Generation

The baud rate for the receiver and transmitter ( Rx and Tx) are set independently and calculated as follows:

$$
T x=\frac{f_{C P U}}{(32 * P R) * T R} \quad R x=\frac{f_{C P U}}{(32 \star P R) \star R R}
$$

with:
$P R=1,3,4$ or 13 (see SCP0 \& SCP1 bits)
$\mathrm{TR}=1,2,4,8,16,32,64,128$
(see SCT0, SCT1 \& SCT2 bits)
$R R=1,2,4,8,16,32,64,128$
(see SCR0,SCR1 \& SCR2 bits)
All these bits are in the BRR register.
Example: If $\mathrm{f}_{\mathrm{CPU}}$ is 8 MHz and if $\mathrm{PR}=13$ and $T R=R R=1$, the transmit and receive baud rates are 19200 baud.
Note: the baud rate registers MUST NOT be changed while the transmitter or the receiver is enabled.

### 4.5.4.5 Receiver Muting and Wake-up Feature

In multiprocessor configurations it is often desirable that only the intended message recipient
should actively receive the full message contents, thus reducing redundant SCI service overhead for all non addressed receivers.
The non addressed devices may be placed in sleep mode by means of the muting function.
Setting the RWU bit by software puts the SCI in sleep mode:
All the reception status bits can not be set.
All the receive interrupt are inhibited.
A muted receiver may be awakened by one of the following two ways:

- by Idle Line detection if the WAKE bit is reset,
- by Address Mark detection if the WAKE bit is set.

Receiver wakes-up by Idle Line detection when the Receive line has recognised an Idle Frame. Then the RWU bit is reset by hardware but the IDLE bit is not set.
Receiver wakes-up by Address Mark detection when it received a " 1 " as the most significant bit of a word, thus indicating that the message is an address. The reception of this particular word wakes up the receiver, sets the RWU bit and sets the RDRF bit, which allows the receiver to receive this word normally and to use it as an address word.

## SERIAL COMMUNICATIONS INTERFACE(Cont'd)

### 4.5.5 Register Description

STATUS REGISTER (SR)
Read Only
Reset Value: 11000000 (C0h)
7 0

| TDRE | TC | RDRF | IDLE | OR | NF | FE | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7 = TDRE Transmit data register empty.
This bit is set by hardware when the content of the TDR register has been transferred into the shift register. An interrupt is generated if the TIE $=1$ in the CR2 register. It is cleared by a software sequence (an access to the SR register followed by a write to the DR register).
0 : Data is not transferred to the shift register
1: Data is transferred to the shift register
Note: data will not be transferred to the shift register as long as the TDRE bit is not reset.

Bit 6 = TC Transmission complete.
This bit is set by hardware when transmission of a frame containing Data, a Preamble or a Break is complete. An interrupt is generated if TCIE=1 in the CR2 register. It is cleared by a software sequence (an access to the SR register followed by a write to the DR register).
0 : Transmission is not complete
1: Transmission is complete
Bit 5 = RDRF Received data ready flag.
This bit is set by hardware when the content of the RDR register has been transferred into the DR register. An interrupt is generated if $\mathrm{RIE}=1$ in the CR2 register. It is cleared by hardware when $R E=0$ or by a software sequence (an access to the SR register followed by a read to the DR register).
0 : Data is not received
1: Received data is ready to be read
Bit 4 = IDLE Idle line detect.
This bit is set by hardware when a Idle Line is detected. An interrupt is generated if the ILIE=1 in the CR2 register. It is cleared by hardware when $R E=0$ by a software sequence (an access to the SR register followed by a read to the DR register). 0 : No Idle Line is detected
1: Idle Line is detected

Note: The IDLE bit will not be set again until the RDRF bit has been set itself (i.e. a new idle line occurs). This bit is not set by an idle line when the receiver wakes up from wake-up mode.

Bit 3 = OR Overrun error.
This bit is set by hardware when the word currently being received in the shift register is ready to be transferred into the RDR register while $R D R F=1$. An interrupt is generated if RIE=1 in the CR2 register. It is cleared by hardware when $R E=0$ by a software sequence (an access to the SR register followed by a read to the DR register).
0 : No Overrun error
1: Overrun error is detected

Note: When this bit is set RDR register content will not be lost but the shift register will be overwritten.
Bit $2=$ NF Noise flag.
This bit is set by hardware when noise is detected on a received frame. It is cleared by hardware when $R E=0$ by a software sequence (an access to the SR register followed by a read to the DR register).
0 : No noise is detected
1: Noise is detected
Note: This bit does not generate interrupt as it appears at the same time as the RDRF bit which itself generates an interrupt.

## Bit 1 = FE Framing error.

This bit is set by hardware when a de-synchronization, excessive noise or a break character is detected. It is cleared by hardware when $\mathrm{RE}=0$ by a software sequence (an access to the SR register followed by a read to the DR register).
0 : No Framing error is detected
1: Framing error or break character is detected
Note: This bit does not generate interrupt as it appears at the same time as the RDRF bit which itself generates an interrupt. If the word currently being transferred causes both frame error and overrun error, it will be transferred and only the OR bit will be set.

Bit $0=$ Reserved, forced by hardware to 0 .

## SERIAL COMMUNICATIONS INTERFACE(Cont'd) CONTROL REGISTER 1 (CR1)

Read/Write
Reset Value: Undefined
7

7 | R8 | T8 | 0 | M | WAKE | 0 | 0 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Bit $7=\mathbf{R 8}$ Receive data bit 8 .
This bit is used to store the 9th bit of the received word when $\mathrm{M}=1$.

Bit $6=\mathbf{T 8}$ Transmit data bit 8 .
This bit is used to store the 9th bit of the transmitted word when $\mathrm{M}=1$.

Bit $5=$ Reserved, forced by hardware to 0 .
Bit $4=\mathbf{M}$ Word length.
This bit determines the data length. It is set or cleared by software.
0: 1 Start bit, 8 Data bits, 1 Stop bit
1: 1 Start bit, 9 Data bits, 1 Stop bit

Bit 3 = WAKE Wake-Up method.
This bit determines the SCI Wake-Up method, it is set or cleared by software.
0 : Idle Line
1: Address Mark

Bit 2:0 = Reserved, forced by hardware to 0.

## CONTROL REGISTER 2 (CR2)

Read/Write
Reset Value: 00000000 (00h)
7 0

| TIE | TCIE | RIE | ILIE | TE | RE | RWU | SBK |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7 = TIE Transmitter interrupt enable.
This bit is set and cleared by software.
0 : interrupt is inhibited
1: An SCl interrupt is generated whenever TDRE=1 in the SR register.

Bit $6=$ TCIE Transmission complete interrupt enable

This bit is set and cleared by software.

0 : interrupt is inhibited
1: An SCl interrupt is generated whenever TC=1 in the SR register

Bit $5=$ RIE Receiver interrupt enable.
This bit is set and cleared by software.
0 : interrupt is inhibited
1: An SCI interrupt is generated whenever OR=1 or $\mathrm{RDRF}=1$ in the SR register

Bit 4 = ILIE Idle line interrupt enable.
This bit is set and cleared by software.
0 : interrupt is inhibited
1: An SCl interrupt is generated whenever IDLE=1 in the SR register.

## Bit 3 = TE Transmitter enable.

This bit enables the transmitter and assigns the TDO pin to the alternate function. It is set and cleared by software.
0 : Transmitter is disabled, the TDO pin is back to the I/O port configuration.
1: Transmitter is enabled
Note: during transmission, a " 0 " pulse on the TE bit ("0" followed by " 1 ") sends a preamble after the current word.

Bit $2=\mathbf{R E}$ Receiver enable.
This bit enables the receiver. It is set and cleared by software.
0: Receiver is disabled, it resets the RDRF, IDLE, OR, NF and FE bits of the SR register.
1: Receiver is enabled and begins searching for a start bit.

Bit 1 = RWU Receiver wake-up.
This bit determines if the SCI is in mute mode or not. It is set and cleared by software and can be cleared by hardware when a wake-up sequence is recognized.
0: Receiver in active mode
1: Receiver in mute mode
Bit $0=$ SBK Send break.
This bit set is used to send break characters. It is set and cleared by software.
0: No break character is transmitted
1: Break characters are transmitted
Note: If the SBK bit is set to " 1 " and then to " 0 ", the transmitter will send a BREAK word at the end of the current word.

## SERIAL COMMUNICATIONS INTERFACE(Cont'd)

 DATA REGISTER (DR)Read/Write
Reset Value: Undefined
Contains the Received or Transmitted data character, depending on whether it is read from or written to.

| 7 |  |  |  |  |  | 0 |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DR7 | DR6 | DR5 | DR4 | DR3 | DR2 | DR1 | DR0 |  |

The Data register performs a double function (read and write) since it is composed of two registers, one for transmission (TDR) and one for reception (RDR).
The TDR register provides the parallel interface between the internal bus and the output shift register (see Figure 1).
The RDR register provides the parallel interface between the input shift register and the internal bus (see Figure 1).

## BAUD RATE REGISTER (BRR)

Read/Write
Reset Value: 00xx xxxx (XXh)

7
0


Bit 7:6= SCP[1:0] First SCI Prescaler
These 2 prescaling bits allow several standard clock division ranges:

| PR Prescaling factor | SCP1 | SCP0 |
| :---: | :---: | :---: |
| 1 | 0 | 0 |
| 3 | 0 | 1 |
| 4 | 1 | 0 |
| 13 | 1 | 1 |

Bit 5:3 = SCT[2:0] SCI Transmitter rate divisor
These 3 bits, in conjunction with the SCP1 \& SCP0 bits define the total division applied to the bus clock to yield the transmit rate clock in conventional Baud Rate Generator mode.

| TR dividing factor | SCT2 | SCT1 | SCT0 |
| :---: | :---: | :---: | :---: |
| 1 | 0 | 0 | 0 |
| 2 | 0 | 0 | 1 |
| 4 | 0 | 1 | 0 |
| 8 | 0 | 1 | 1 |
| 16 | 1 | 0 | 0 |
| 32 | 1 | 0 | 1 |
| 64 | 1 | 1 | 0 |
| 128 | 1 | 1 | 1 |

Bit 2:0 $=$ SCR[2:0] SCI Receiver rate divisor.
These 3 bits, in conjunction with the SCP1 \& SCP0 bits define the total division applied to the bus clock to yield the receive rate clock in conventional Baud Rate Generator mode.

| RR dividing factor | SCR2 | SCR1 | SCR0 |
| :---: | :---: | :---: | :---: |
| 1 | 0 | 0 | 0 |
| 2 | 0 | 0 | 1 |
| 4 | 0 | 1 | 0 |
| 8 | 0 | 1 | 1 |
| 16 | 1 | 0 | 0 |
| 32 | 1 | 0 | 1 |
| 64 | 1 | 1 | 0 |
| 128 | 1 | 1 | 1 |

SERIAL COMMUNICATIONS INTERFACE (Cont'd)
Table 17. SCI Register Map and Reset Values

| Address <br> (Hex.) | Register <br> Name | $\mathbf{7}$ | $\mathbf{6}$ | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ |
| :---: | :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 30 | SR | TDRE | TC | RDRF | IDLE | OR | NF | FE | 0 |
|  | Reset Value | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |
| 31 | DR | DR7 | DR6 | DR5 | DR4 | DR3 | DR2 | DR1 | DR0 |
|  | Reset Value | $x$ | $x$ | $x$ | $x$ | $x$ | $x$ | $x$ | $x$ |
| 32 | BRR | SCP1 | SCP0 | SCT2 | SCT1 | SCT0 | SCR2 | SCR1 | SCR0 |
|  | Reset Value | 0 | 0 | $x$ | $x$ | $x$ | $x$ | $x$ | $x$ |
| 33 | CR1 | R8 | T8 | 0 | $M$ | WAKE | 0 | 0 | 0 |
|  | Reset Value | $x$ | $x$ | 0 | $x$ | $x$ | 0 | 0 | 0 |
| 34 | CR2 | TIE | TCIE | RIE | ILIE | TE | RE | RWU | SBK |
|  | Reset Value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

### 4.6 PWM/BRM GENERATOR (DAC)

### 4.6.1 Introduction

This PWM/BRM peripheral includes two types of PWM/BRM outputs, with differing step resolutions based on the Pulse Width Modulator (PWM) and Binary Rate Multiplier (BRM) Generator technique are available. It allows the digital to analog conversion (DAC) when used with external filtering.

### 4.6.2 Main Features

- Fixed frequency: $\mathrm{f}_{\mathrm{CP}} / 64$
- Resolution: $T_{\text {CPU }}$
- 10-Bit PWM/BRM generator with a step of $\mathrm{V}_{\mathrm{DD}} / 2^{10}\left(5 \mathrm{mV}\right.$ if $\left.\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}\right)$
- 12-bit PWM/BRM generator with step of $\mathrm{V}_{D D} / 2^{12}\left(1.25 \mathrm{mV}\right.$ if $\left.\mathrm{V}_{D D}=5 \mathrm{~V}\right)$.


### 4.6.3 Functional Description

### 4.6.3.1 10-bit PWM/BRM

The 10 bits of the 10 -bit PWM/BRM are distributed as 6 PWM bits and 4 BRM bits. The generator consists of a 12-bit counter (common for all channels), a comparator and the PWM/BRM generation logic.

## PWM Generation

The counter increments continuously, clocked at internal CPU clock. Whenever the 6 least significant bits of the counter (defined as the PWM counter) overflow, the output level for all active channels is set.
The state of the PWM counter is continuously compared to the PWM binary weight for each channel, as defined in the relevant PWM register, and when a match occurs the output level for that channel is reset.
This Pulse Width modulated signal must be filtered, using an external RC network placed as close as possible to the associated pin. This provides an analog voltage proportional to the average charge passed to the external capacitor. Thus for a higher mark/space ratio (High time much greater than Low time) the average output voltage is higher. The external components of the RC network should be selected for the filtering level required for control of the system variable.
Each output may individually have its polarity inverted by software, and can also be used as a logical output.

Figure 27. PWM Generation


## PWM/BRM GENERATOR (Cont’d)

## PWM/BRM Outputs

The PWM/BRM outputs are assigned to dedicated pins.
In these pins, the PWM/BRM outputs are connected to a serial resistor which must be taken into account to calculate the RC filter (seeFigure 2). In any case, the RC filter time must be higher than TCPUx64.

Figure 28. Typical PWM Output Filter


Table 18. 6-Bit PWM Ripple After Filtering

| $\mathbf{C}_{\text {ext }}(\mu \mathbf{F})$ | V RIPPLE (mV) |
| :---: | :---: |
| 0.128 | 78 |
| 1.28 | 7.8 |
| 12.8 | 0.78 |

With RC filter ( $\mathrm{R}=1 \mathrm{~kW}$ ),
$\mathrm{f}_{\mathrm{CPU}}=8 \mathrm{MHz}$
$V_{D D}=5 \mathrm{~V}$
PWM Duty Cycle 50\%
R=Rint+Rext (Rext is optional).

Note: After a reset these pins are tied low by default and are not in a high impedance state.

Figure 29. PWM Simplified Voltage Output After Filtering


## PWM/BRM GENERATOR (Cont'd)

## BRM Generation

The BRM bits allow the addition of a pulse to widen a standard PWM pulse for specific PWM cycles. This has the effect of "fine-tuning" the PWM Duty cycle (without modifying the base duty cycle), thus, with the external filtering, providing additional fine voltage steps.
The incremental pulses (with duration of $T_{C P U}$ ) are added to the beginning of the original PWM pulse. The PWM intervals which are added to are specified in the 4-bit BRM register and are encoded as shown in the following table. The BRM values shown may be combined together to provide a summation of the incremental pulse intervals specified.
The pulse increment corresponds to the PWM resolution.
For example,if

- Data 18h is written to the PWM register
- Data 06h (00000110b) is written to the BRM register
- with a 8 MHz internal clock (125ns resolution)

Then $3.0 \mu \mathrm{~s}$-long pulse will be output at $8 \mu \mathrm{~s}$ intervals, except for cycles numbered $2,4,6,10,12,14$, where the pulse is broadened to $3.125 \mu \mathrm{~s}$.

Note. If 00h is written to both PWM and BRM registers, the generator output will remain at "0". Conversely, if both registers hold data 3Fh and OFh, respectively, the output will remain at " 1 " for all intervals \#1 to \#15, but it will return to zero at interval \#O for an amount of time corresponding to the PWM resolution ( $\mathrm{T}_{\mathrm{CPU}}$ ).
An output can be set to a continuous " 1 " level by clearing the PWM and BRM values and setting POL = " 1 " (inverted polarity) in the PWM register. This allows a PWM/BRM channel to be used as an additional I/O pin if the DAC function is not required.

Table 19. Bit BRM Added Pulse Intervals (Interval \#0 not selected).

| BRM 4 - Bit Data | Incremental Pulse Intervals |
| :---: | :--- |
| 0000 | none |
| 0001 | $\mathrm{i}=8$ |
| 0010 | $\mathrm{i}=4,12$ |
| 0100 | $\mathrm{i}=2,6,10,14$ |
| 1000 | $\mathrm{i}=1,3,5,7,9,11,13,15$ |

Figure 30. BRM pulse addition (PWM > 0)


## PWM/BRM GENERATOR (Cont'd)

Figure 31. Simplified Filtered Voltage Output Schematic with BRM added


Figure 32. Graphical Representation of 4-Bit BRM Added Pulse Positions


## PWM/BRM GENERATOR (Cont'd)

### 4.6.3.2 12-Bit PWM/BRM

The 12 bits of the 12-bit PWM/BRM generator are distributed as 6 PWM bits and 6 BRM bits.

## PWM Generation

The functionality of the PWM generation is equivalent to the PWM generation of the 10-bit PWM/ BRM described in the previous paragraph and so will not be repeated here. Please refer to the previous paragraph for functionality, to be used in conjunction with the Register description.

## BRM Generation

A 6-bit BRM register defining the intervals where an incremental pulse (with duration of $T_{\text {CPU }}$ ) is added to the beginning of the original PWM pulse.

| BRM 6 - Bit Data | Incremental Pulse Intervals |
| :---: | :--- |
| 000000 | none |
| 000001 | $\mathrm{i}=32$ |
| 000010 | $\mathrm{i}=16,48$ |
| 000100 | $\mathrm{i}=8,24,40,56$ |
| 001000 | $\mathrm{i}=4,12,20,28,36,44,52,60$ |
| 010000 | $\mathrm{i}=2,6,10, \ldots 50,54,58,62$ |
| 100000 | $\mathrm{i}=1,3,5,7,9, \ldots 55,59,61,63$ |

### 4.6.3.3 PWM/BRM OUTPUTS

The PWM/BRM outputs are assigned to dedicated pins.
If necessary, these pins can be used in push-pull or open-drain modes under software control.
In these pins, the PWM/BRM outputs are connected to a serial resistor which must be taken into account to calculate the RC filter.

Figure 33. Precision for PWM/BRM Tuning for VOUTEFF (After filtering)


## PWM/BRM GENERATOR (Cont'd)

### 4.6.4 Register Description

### 4.6.4.1 10-bit PWM/BRM REGISTERS

On a channel basis, the 10 bits are separated into two data registers:

- A 6-bit PWM register corresponding to the binary weight of the PWM pulse.
- A 4-bit BRM register defining the intervals where an incremental pulse is added to the beginning of the original PWM pulse. Two BRM channel values share the same register.
Note: The number of PWM and BRM channels available depends on the device. Refer to the device pin description and register map.


## PWM[1:8] REGISTERS

Read/Write
Reset Value 10000000 (80h)


| 1 | POL | P5 | P4 | P3 | P2 | P1 | P0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Bit $7=$ Reserved (read as " 1 ")

Bit $6=$ POL Polarity Bit.
When POL is set, output signal polarity is inverse; otherwise, no change occurs.

Bits 5:0 = P[5:0] PWM Pulse Binary Weight for channel i.

## BRM REGISTERS

BRM21 (Channels $2+1$ )
BRM43 (Channels $4+3$ )
BRM65 (Channels 6 + 5)
BRM87 (Channels 8 + 7)
Read/Write
Reset Value: 00000000 (00h)

$$
\begin{array}{ll}
7 & 0
\end{array}
$$

| B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bits 7:4 = $\mathbf{B}[7: 4]$ BRM Bits (channel $i+1$ )

### 4.6.4.2 12-bit PWM/BRM REGISTERS

The 12 bits are separated into two data registers:

- A 6-bit PWM register corresponding to the binary weight of the PWM pulse.
- A 6-bit BRM register defining the intervals where incremental pulses are added to the beginning of the original PWM pulse.


## PWMO REGISTER

Read/ Write
Reset Value: 10000000 (80h)

$$
\begin{array}{ll}
7 & 0
\end{array}
$$

| 1 | POL | P 5 | P 4 | P 3 | P 2 | P 1 | P 0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Bit 7 = Reserved (read as " 1 ")

Bit 6 = POL Polarity Bit.
When POL is set, output signal polarity is inverse; otherwise, no change occurs.

Bits 5:0 = P[5:0] PWM Pulse Binary Weight

## BRMO REGISTER

Read/ Write
Reset Value: 11000000 (C0h)


Bits 7:6 = Unused

Bits 5:0 $=\mathbf{B}[5: 0]$ BRM Bits

Note: From the programmer's point of view, the PWM and BRM registers can be regarded as being combined to give one data value.
For example (10-bit):


Effective (with external RC filtering) DAC value

| 0 | $P O L$ | $P$ | $P$ | $P$ | $P$ | $P$ | $P$ | $B$ | $B$ | $B$ | $B$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Table 20. PWMA Register Map

| Address <br> (Hex.) | Register Name | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 22 | PWM0 |  | POL | P5 ..P0 |  |  |  |  |  |
| 23 | BRMO |  |  | BRM Channel 0 |  |  |  |  |  |
| 24 | PWM1 |  | POL | P5 ..P0 |  |  |  |  |  |
| 25 | BRM21 | BRM Channel 2 |  |  |  | BRM Channel 1 |  |  |  |
| 26 | PWM2 |  | POL | P5 ..P0 |  |  |  |  |  |
| 27 | PWM3 |  | POL | P5 ..P0 |  |  |  |  |  |
| 28 | BRM43 | BRM Channel 4 |  |  |  | BRM Channel 3 |  |  |  |
| 29 | PWM4 |  | POL | P5 ..P0 |  |  |  |  |  |

### 4.7 8-BIT A/D CONVERTER (ADC)

### 4.7.1 Introduction

The on-chip Analog to Digital Converter (ADC) peripheral is a 8 -bit, successive approximation converter with internal sample and hold circuitry. This peripheral has up to 8 multiplexed analog input channels (refer to device pin out description) that allow the peripheral to convert the analog voltage levels from up to 8 different sources.
The result of the conversion is stored in a 8 -bit Data Register. The A/D converter is controlled through a Control/Status Register.

### 4.7.2 Main Features

- 8 -bit conversion
- Up to 8 channels with multiplexed input
- Linear successive approximation
- Data register (DR) which contains the results
- Conversion complete status flag
- On/off bit (to reduce consumption)

The block diagram is shown inFigure 1.

Figure 34. ADC block diagram


## 8-BIT A/D CONVERTER (ADC)(Cont'd)

### 4.7.3 Functional Description

The high level reference voltage VDA must be connected externally to the $V_{D D}$ pin. The low level reference voltage $\mathrm{V}_{S S A}$ must be connected externally to the $\mathrm{V}_{S S}$ pin. In some devices (refer to device pin out description) high and low level reference voltages are internally connected to the $V_{D}$ and $\mathrm{V}_{\mathrm{SS}}$ pins.
Conversion accuracy may therefore be degraded by voltage drops and noise in the event of heavily loaded or badly decoupled power supply lines.

## Characteristics

The conversion is monotonic meaning the result never decreases if the analog input does not and never increases if the analog input does not.
If input voltage is greater than or equal to $V_{D D}$ (voltage reference high) then results = FFh (full scale) without overflow indication.
If input voltage $\leq \mathrm{V}_{\mathrm{SS}}$ (voltage reference low) then the results $=00 \mathrm{~h}$.

The conversion time is 64 CPU clock cycles including a sampling time of 31.5 CPU clock cycles.

The $A / D$ converter is linear and the digital result of the conversion is given by the formula:

$$
\text { Digital result }=\frac{255 \text { * Input Voltage }}{\text { Reference Voltage }}
$$

Where Reference Voltage is $\mathrm{V}_{\mathrm{DD}}-\mathrm{V}_{\mathrm{Ss}}$.
The accuracy of the conversion is described in the Electrical Characteristics Section.

## Procedure

Refer to the CSR and SR registers Section 0.1.4 for the bit definitions.

The analog input ports must be configured as input, no pull-up, no interrupt. Refer to the «l/O ports» chapter. Using these pins as analog inputs does not affect the ability of the port to be read as a logic input.
In the CSR register:

- Select the CH 2 to CH 0 bits to assign the analog channel to convert. Refer toTable 1.
- Set the ADON bit. Then the A/D converter is enabled after a stabilization time (typically 30 $\mu \mathrm{s})$. It then performs a continuous conversion of the selected channel.
When a conversion is complete
- The COCO bit is set by hardware.
- No interrupt is generated.
- The result is in the DR register.

A write to the CSR register aborts the current conversion, resets the COCO bit and starts a new conversion.

Notes: The A/D converter may be disabled by resetting the ADON bit. This feature allows reduced power consumption when no conversion is needed.

The A/D converter is not affected by WAIT mode. When the MCU enters HALT mode with the A/D converter enabled, the converter is disabled until the HALT mode is exited and the start-up delay has elapsed. A stabilisation time is also required before accurate conversions can be performed.

## 8-BIT A/D CONVERTER (ADC)(Cont'd)

### 4.7.4 Register Description CONTROL/STATUS REGISTER (CSR)

## Read/Write

Reset Value: 00000000 (00h)


Bit $7=$ COCO Conversion Complete.
This bit is set by hardware. It is cleared by software reading the result in the DR register or writing to the CSR register.
0 : Conversion is not complete.
1: Conversion can be read from the DR register.

Bit $6=$ Reserved. Must always be cleared.
Bit 5 = ADON A/D converter On.
This bit is set and cleared by software.
0 : A/D converter is switched off.
1 : A/D converter is switched on.
Note: a typically $30 \mu$ s delay time is necessary for the ADC to stabilize when the ADON bit is set.

Bit 4 = Reserved. Forced by hardware to 0 .
Bit $3=$ Reserved. Must always be cleared.

Bits 2-0: CH2-CH0 Channel Selection.
These bits are set and cleared by software. They select the analog input to convert.

Table 21. Channel Selection

| Pin* | CH 2 | CH 1 | CH 0 |
| :---: | :---: | :---: | :---: |
| AIN0 | 0 | 0 | 0 |
| AIN1 | 0 | 0 | 1 |
| AIN2 | 0 | 1 | 0 |
| AIN3 | 0 | 1 | 1 |
| AIN4 | 1 | 0 | 0 |
| AIN5 | 1 | 0 | 1 |
| AIN6 | 1 | 1 | 0 |
| AIN7 | 1 | 1 | 1 |

(*The number of pins varies according to the device. Refer to the device pinout).

## DATA REGISTER (DR)

Read Only
Reset Value: 00000000 (00h)
7

| AD7 | AD6 | AD5 | AD4 | AD3 | AD2 | AD1 | AD0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Bit 7:0 = AD7-AD0 Analog Converted Value.
This register contains the converted analog value in the range 00h to FFh.
Reading this register reset the COCO flag.

Table 22. ADC Register Map

| Address <br> (Hex.) | Register <br> Name | $\mathbf{7}$ | $\mathbf{6}$ | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $0 B$ | CSR | COCO | - | ADON | 0 | - | $C H 2$ | $C H 1$ | $C H 0$ |
| $0 A$ | DR | AD7 .. ADO |  |  |  |  |  |  |  |

## 5 INSTRUCTION SET

### 5.1 ST7 ADDRESSING MODES

The ST7 Core features 17 different addressing modes which can be classified in 7 main groups:

| Addressing Mode | Example |
| :--- | :--- |
| Inherent | nop |
| Immediate | Id A,\#\$55 |
| Direct | Id A,\$55 |
| Indexed | Id A, $(\$ 55, \mathrm{X})$ |
| Indirect | Id A,([\$55],X) |
| Relative | jrne loop |
| Bit operation | bset byte,\#5 |

The ST7 Instruction set is designed to minimize the number of bytes required per instruction: To do
so, most of the addressing modes may be subdivided in two sub-modes called long and short:

- Long addressing mode is more powerful because it can use the full 64 Kbyte address space, however it uses more bytes and more CPU cycles.
- Short addressing mode is less powerful because it can generally only access page zero (0000h $00 F F h$ range), but the instruction size is more compact, and faster. All memory to memory instructions use short addressing modes only (CLR, CPL, NEG, BSET, BRES, BTJT, BTJF, INC, DEC, RLC, RRC, SLL, SRL, SRA, SWAP)
The ST7 Assembler optimizes the use of long and short addressing modes.

Table 23. ST7 Addressing Mode Overview

| Mode |  |  | Syntax | Destination | Pointer Address (Hex.) | Pointer Size (Hex.) | Length (Bytes) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Inherent |  |  | nop |  |  |  | + 0 |
| Immediate |  |  | Id A,\#\$55 |  |  |  | + 1 |
| Short | Direct |  | Id A,\$10 | 00..FF |  |  | +1 |
| Long | Direct |  | Id A,\$1000 | 0000..FFFF |  |  | + 2 |
| No Offset | Direct | Indexed | Id A,(X) | 00..FF |  |  | + 0 |
| Short | Direct | Indexed | Id $\mathrm{A},(\$ 10, \mathrm{X})$ | 00..1FE |  |  | +1 |
| Long | Direct | Indexed | Id A,(\$1000,X) | 0000..FFFF |  |  | +2 |
| Short | Indirect |  | Id A,[\$10] | 00..FF | 00..FF | byte | + 2 |
| Long | Indirect |  | Id A,[\$10.w] | 0000..FFFF | 00..FF | word | + 2 |
| Short | Indirect | Indexed | Id A,([\$10],X) | 00..1FE | 00..FF | byte | +2 |
| Long | Indirect | Indexed | Id A,([\$10.w],X) | 0000..FFFF | 00..FF | word | +2 |
| Relative | Direct |  | jrne loop | PC+/-127 |  |  | +1 |
| Relative | Indirect |  | jrne [\$10] | PC+/-127 | 00..FF | byte | $+2$ |
| Bit | Direct |  | bset \$10,\#7 | 00..FF |  |  | +1 |
| Bit | Indirect |  | bset [\$10],\#7 | 00..FF | 00..FF | byte | + 2 |
| Bit | Direct | Relative | btjt \$10,\#7,skip | 00..FF |  |  | + 2 |
| Bit | Indirect | Relative | btjt [\$10],\#7,skip | 00..FF | 00..FF | byte | + 3 |

## ST7 ADDRESSING MODES(Cont’d)

### 5.1.1 Inherent

All Inherent instructions consist of a single byte. The opcode fully specifies all the required information for the CPU to process the operation.

| Inherent Instruction | Function |
| :--- | :--- |
| NOP | No operation |
| TRAP | S/W Interrupt |
| WFI | Wait For Interrupt (Low Power <br> Mode) |
| HALT | Halt Oscillator (Lowest Power <br> Mode) |
| RET | Sub-routine Return |
| IRET | Interrupt Sub-routine Return |
| SIM | Set Interrupt Mask |
| RIM | Reset Interrupt Mask |
| SCF | Set Carry Flag |
| RCF | Reset Carry Flag |
| RSP | Reset Stack Pointer |
| LD | Load |
| CLR | Clear |
| PUSH/POP | Push/Pop to/from the stack |
| INC/DEC | Increment/Decrement |
| TNZ | Test Negative or Zero |
| CPL, NEG | 1 or 2 Complement |
| MUL | Byte Multiplication |
| SLL, SRL, SRA, RLC, <br> RRC | Shift and Rotate Operations |
| SWAP | Swap Nibbles |

### 5.1.2 Immediate

Immediate instructions have two bytes, the first byte contains the opcode, the second byte contains the the operand value. .

| Immediate Instruction | Function |
| :--- | :--- |
| LD | Load |
| CP | Compare |
| BCP | Bit Compare |
| AND, OR, XOR | Logical Operations |
| ADC, ADD, SUB, SBC | Arithmetic Operations |

### 5.1.3 Direct

In Direct instructions, the operands are referenced by their memory address.
The direct addressing mode consists of two submodes:

## Direct (short)

The address is a byte, thus requires only one byte after the opcode, but only allows 00 - FF addressing space.

## Direct (long)

The address is a word, thus allowing 64 Kbyte addressing space, but requires 2 bytes after the opcode.

### 5.1.4 Indexed (No Offset, Short, Long)

In this mode, the operand is referenced by its memory address, which is defined by the unsigned addition of an index register ( X or Y ) with an offset.
The indirect addressing mode consists of three sub-modes:

## Indexed (No Offset)

There is no offset, (no extra byte after the opcode), and allows 00 - FF addressing space.

## Indexed (Short)

The offset is a byte, thus requires only one byte after the opcode and allows 00 - 1FE addressing space.

## Indexed (long)

The offset is a word, thus allowing 64 Kbyte addressing space and requires 2 bytes after the opcode.

### 5.1.5 Indirect (Short, Long)

The required data byte to do the operation is found by its memory address, located in memory (pointer).
The pointer address follows the opcode. The indirect addressing mode consists of two sub-modes:

## Indirect (short)

The pointer address is a byte, the pointer size is a byte, thus allowing 00 - FF addressing space, and requires 1 byte after the opcode.

## Indirect (long)

The pointer address is a byte, the pointer size is a word, thus allowing 64 Kbyte addressing space, and requires 1 byte after the opcode.

## ST7 ADDRESSING MODES(Cont'd)

### 5.1.6 Indirect Indexed (Short, Long)

This is a combination of indirect and short indexed addressing modes. The operand is referenced by its memory address, which is defined by the unsigned addition of an index register value (X or Y) with a pointer value located in memory. The pointer address follows the opcode.
The indirect indexed addressing mode consists of two sub-modes:

## Indirect Indexed (Short)

The pointer address is a byte, the pointer size is a byte, thus allowing 00 - 1FE addressing space, and requires 1 byte after the opcode.

## Indirect Indexed (Long)

The pointer address is a byte, the pointer size is a word, thus allowing 64 Kbyte addressing space, and requires 1 byte after the opcode.
Table 24. Instructions Supporting Direct, Indexed, Indirect and Indirect Indexed Addressing Modes

| Long and Short <br> Instructions | Function |
| :--- | :--- |
| LD | Load |
| CP | Compare |
| AND, OR, XOR | Logical Operations |
| ADC, ADD, SUB, SBC | Arithmetic Additions/Sub- <br> stractions operations |
| BCP | Bit Compare |


| Short Instructions Only | Function |
| :--- | :--- |
| CLR | Clear |
| INC, DEC | Increment/Decrement |
| TNZ | Test Negative or Zero |
| CPL, NEG | 1 or 2 Complement |
| BSET, BRES | Bit Operations |
| BTJT, BTJF | Bit Test and Jump Opera- <br> tions |
| SLL, SRL, SRA, RLC, <br> RRC | Shift and Rotate Operations |
| SWAP | Swap Nibbles |
| CALL, JP | Call or Jump subroutine |

### 5.1.7 Relative mode (Direct, Indirect)

This addressing mode is used to modify the PC register value, by adding an 8-bit signed offset to it.

| Available Relative Direct/ <br> Indirect Instructions | Function |
| :--- | :--- |
| JRxx | Conditional Jump |
| CALLR | Call Relative |

The relative addressing mode consists of two submodes:

## Relative (Direct)

The offset is following the opcode.

## Relative (Indirect)

The offset is defined in memory, which address follows the opcode.

### 5.2 INSTRUCTION GROUPS

The ST7 family devices use an Instruction Set consisting of 63 instructions. The instructions may
be subdivided into 13 main groups as illustrated in the following table:

| Load and Transfer | LD | CLR |  |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Stack operation | PUSH | POP | RSP |  |  |  |  |  |
| Increment/Decrement | INC | DEC |  |  |  |  |  |  |
| Compare and Tests | CP | TNZ | BCP |  |  |  |  |  |
| Logical operations | AND | OR | XOR | CPL | NEG |  |  |  |
| Bit Operation | BSET | BRES |  |  |  |  |  |  |
| Conditional Bit Test and Branch | BTJT | BTJF |  |  |  |  |  |  |
| Arithmetic operations | ADC | ADD | SUB | SBC | MUL |  |  |  |
| Shift and Rotates | SLL | SRL | SRA | RLC | RRC | SWAP | SLA |  |
| Unconditional Jump or Call | JRA | JRT | JRF | JP | CALL | CALLR | NOP | RET |
| Conditional Branch | JRxx |  |  |  |  |  |  |  |
| Interruption management | TRAP | WFI | HALT | IRET |  |  |  |  |
| Code Condition Flag modification | SIM | RIM | SCF | RCF |  |  |  |  |

## Using a pre-byte

The instructions are described with one to four opcodes.
In order to extend the number of available opcodes for an 8 -bit CPU (256 opcodes), three different probate pockets are defined. These prebytes modify the meaning of the instruction they precede.
The whole instruction becomes:
PC-2 End of previous instruction
PC-1 Prebyte
PC opcode
PC+1 Additional word (0 to 2) according to the number of bytes required to compute the effective address

These prebytes enable instruction in Y as well as indirect addressing modes to be implemented. They precede the opcode of the instruction in X or the instruction using direct addressing mode. The prebytes are:
PDY 90 Replace an X based instruction using immediate, direct, indexed, or inherent addressing mode by a Y one.
PIX 92 Replace an instruction using direct, direct bit, or direct relative addressing mode to an instruction using the corresponding indirect addressing mode.
It also changes an instruction using X indexed addressing mode to an instruction using indirect $X$ indexed addressing mode.
PIY 91 Replace an instruction using $X$ indirect indexed addressing mode by a $Y$ one.

INSTRUCTION GROUPS (Cont'd)

| Mnemo | Description | Function/Example | Dst | Src | H | 1 | N | Z | C |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADC | Add with Carry | $A=A+M+C$ | A | M | H |  | N | Z | C |
| ADD | Addition | $A=A+M$ | A | M | H |  | N | Z | C |
| AND | Logical And | $\mathrm{A}=\mathrm{A} . \mathrm{M}$ | A | M |  |  | N | Z |  |
| BCP | Bit compare A, Memory | tst (A.M) | A | M |  |  | N | Z |  |
| BRES | Bit Reset | bres Byte, \#3 | M |  |  |  |  |  |  |
| BSET | Bit Set | bset Byte, \#3 | M |  |  |  |  |  |  |
| BTJF | Jump if bit is false (0) | btjf Byte, \#3, Jmp1 | M |  |  |  |  |  | C |
| BTJT | Jump if bit is true (1) | btjt Byte, \#3, Jmp1 | M |  |  |  |  |  | C |
| CALL | Call subroutine |  |  |  |  |  |  |  |  |
| CALLR | Call subroutine relative |  |  |  |  |  |  |  |  |
| CLR | Clear |  | reg, M |  |  |  | 0 | 1 |  |
| CP | Arithmetic Compare | tst(Reg - M) | reg | M |  |  | N | Z | C |
| CPL | One Complement | A = FFH-A | reg, M |  |  |  | N | Z | 1 |
| DEC | Decrement | $\operatorname{dec} \mathrm{Y}$ | reg, M |  |  |  | N | Z |  |
| HALT | Halt |  |  |  |  | 0 |  |  |  |
| IRET | Interrupt routine return | Pop CC, A, X, PC |  |  | H | 1 | N | Z | C |
| INC | Increment | inc X | reg, M |  |  |  | N | Z |  |
| JP | Absolute Jump | jp [TBL.w] |  |  |  |  |  |  |  |
| JRA | Jump relative always |  |  |  |  |  |  |  |  |
| JRT | Jump relative |  |  |  |  |  |  |  |  |
| JRF | Never jump | jrf * |  |  |  |  |  |  |  |
| JRIH | Jump if ext. interrupt = 1 |  |  |  |  |  |  |  |  |
| JRIL | Jump if ext. interrupt $=0$ |  |  |  |  |  |  |  |  |
| JRH | Jump if H = 1 | $\mathrm{H}=1$ ? |  |  |  |  |  |  |  |
| JRNH | Jump if $\mathrm{H}=0$ | $\mathrm{H}=0$ ? |  |  |  |  |  |  |  |
| JRM | Jump if $\mathrm{I}=1$ | $\mathrm{I}=1$ ? |  |  |  |  |  |  |  |
| JRNM | Jump if $\mathrm{I}=0$ | $\mathrm{I}=0$ ? |  |  |  |  |  |  |  |
| JRMI | Jump if $\mathrm{N}=1$ (minus) | $\mathrm{N}=1$ ? |  |  |  |  |  |  |  |
| JRPL | Jump if $\mathrm{N}=0$ (plus) | $\mathrm{N}=0$ ? |  |  |  |  |  |  |  |
| JREQ | Jump if $\mathrm{Z}=1$ (equal) | $\mathrm{Z}=1$ ? |  |  |  |  |  |  |  |
| JRNE | Jump if $\mathrm{Z}=0$ (not equal) | $\mathrm{Z}=0$ ? |  |  |  |  |  |  |  |
| JRC | Jump if $\mathrm{C}=1$ | $\mathrm{C}=1$ ? |  |  |  |  |  |  |  |
| JRNC | Jump if $\mathrm{C}=0$ | $\mathrm{C}=0$ ? |  |  |  |  |  |  |  |
| JRULT | Jump if $\mathrm{C}=1$ | Unsigned < |  |  |  |  |  |  |  |
| JRUGE | Jump if $\mathrm{C}=0$ | Jmp if unsigned >= |  |  |  |  |  |  |  |
| JRUGT | Jump if ( $\mathrm{C}+\mathrm{Z}=0$ ) | Unsigned > |  |  |  |  |  |  |  |

INSTRUCTION GROUPS (Cont'd)

| JRULE | Jump if ( $C+Z=1$ ) | Unsigned <= |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| LD | Load | dst <= src | reg, M | M, reg |  |  | N | Z |  |
| MUL | Multiply | $\mathrm{X}, \mathrm{A}=\mathrm{X}$ * A | A, X, Y | X, Y, A | 0 |  |  |  | 0 |
| NEG | Negate (2's compl) | neg \$10 | reg, M |  |  |  | N | Z | C |
| NOP | No Operation |  |  |  |  |  |  |  |  |
| OR | OR operation | A $=\mathrm{A}+\mathrm{M}$ | A | M |  |  | N | Z |  |
| POP | Pop from the Stack | pop reg pop CC | $\begin{aligned} & \mathrm{reg} \\ & \mathrm{CC} \end{aligned}$ | $\begin{aligned} & \mathrm{M} \\ & \mathrm{M} \end{aligned}$ | H | 1 | N | Z | C |
| PUSH | Push onto the Stack | push Y | M | reg, CC |  |  |  |  |  |
| RCF | Reset carry flag | C = 0 |  |  |  |  |  |  | 0 |
| RET | Subroutine Return |  |  |  |  |  |  |  |  |
| RIM | Enable Interrupts | $\mathrm{I}=0$ |  |  |  | 0 |  |  |  |
| RLC | Rotate left true C | $\mathrm{C}<=\mathrm{A}<=C$ | reg, M |  |  |  | N | Z | C |
| RRC | Rotate right true C | C $=>$ A $=>$ C | reg, M |  |  |  | N | Z | C |
| RSP | Reset Stack Pointer | S = Max allowed |  |  |  |  |  |  |  |
| SBC | Substract with Carry | A $=\mathrm{A}-\mathrm{M}-\mathrm{C}$ | A | M |  |  | N | Z | C |
| SCF | Set carry flag | $C=1$ |  |  |  |  |  |  | 1 |
| SIM | Disable Interrupts | $\mathrm{I}=1$ |  |  |  | 1 |  |  |  |
| SLA | Shift left Arithmetic | $\mathrm{C}<=\mathrm{A}<=0$ | reg, M |  |  |  | N | Z | C |
| SLL | Shift left Logic | $\mathrm{C}<=\mathrm{A}<=0$ | reg, M |  |  |  | N | Z | C |
| SRL | Shift right Logic | $0 \Rightarrow>A \Rightarrow C$ | reg, M |  |  |  | 0 | Z | C |
| SRA | Shift right Arithmetic | A7 $=>$ A $=>$ C | reg, M |  |  |  | N | Z | C |
| SUB | Substraction | $\mathrm{A}=\mathrm{A}-\mathrm{M}$ | A | M |  |  | N | Z | C |
| SWAP | SWAP nibbles | A7-A4 <=> A3-A0 | reg, M |  |  |  | N | Z |  |
| TNZ | Test for Neg \& Zero | tnz lbl1 |  |  |  |  | N | Z |  |
| TRAP | S/W trap | S/W interrupt |  |  |  | 1 |  |  |  |
| WFI | Wait for Interrupt |  |  |  |  | 0 |  |  |  |
| XOR | Exclusive OR | A = A XOR M | A | M |  |  | N | Z |  |

## 6 ELECTRICAL CHARACTERISTICS

### 6.1 ABSOLUTE MAXIMUM RATINGS

This product contains devices to protect the inputs against damage due to high static voltages, however it is advisable to take normal precaution to avoid application of any voltage higher than the specified maximum rated voltages.

For proper operation it is recommended that Y and $\mathrm{V}_{\mathrm{O}}$ be higher than $\mathrm{V}_{S S}$ and lower than $\mathrm{V}_{\mathrm{DD}}$. Reliability is enhanced if unused inputs are connected to an appropriate logic voltage level (VD or $\mathrm{V}_{\mathrm{SS}}$ ).

Power Considerations.The average chip-junction temperature, $\mathrm{T}_{\mathrm{J}}$, in Celsius can be obtained from:
$T_{J}=\quad T A+P D \times R t h J A$
Where: $T_{A}=$ Ambient Temperature.
RthJA = Package thermal resistance (junction-to ambient).
$P_{D}=P_{\text {INT }}+P_{\text {PORT }}$
$P_{\text {INT }}=I_{D D} \times V_{D D}$ (chip internal power).
PPORT $=$ Port power dissipation (determined by the user).

| Symbol <br> $\mathrm{V}_{\mathrm{DD}}$ | Supply Voltage | Value <br> -0.3 to 6.0 | Unit <br> V |
| :---: | :--- | :---: | :---: |
| $\mathrm{V}_{\mathrm{DDA}}$ | Analog Reference Voltage | -0.3 to 6.0 | V |
| $\mathrm{~V}_{\mathrm{I}}$ | Input Voltage | $\mathrm{V}_{\mathrm{SS}}-0.3$ to $\mathrm{V}_{\mathrm{DD}}+0.3$ | V |
| $\mathrm{~V}_{\mathrm{O}}$ | Output Voltage | $\mathrm{V}_{\mathrm{SS}}-0.3$ to $\mathrm{V}_{\mathrm{DD}}+0.3$ | V |
| $\mathrm{IV}_{\mathrm{DD}}$ | Total Current into $\mathrm{V}_{\mathrm{DD}}$ (source) | TBD | mA |
| $\mathrm{IV}_{\mathrm{SS}}$ | Total Current out of $\mathrm{V}_{\mathrm{SS}}$ (sink) | TBD | mA |
| $\mathrm{T}_{J}$ | Junction Temperature | 150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\text {STG }}$ | Storage Temperature | -60 to 150 | ${ }^{\circ} \mathrm{C}$ |

Note: Stresses above those listed as "absolute maximum ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

### 6.2 RECOMMENDED OPERATING CONDITIONS

| Symbol <br> $\mathrm{T}_{\mathrm{A}}$ | Parameter <br> Operating Temperature | Test Conditions <br> 1 Suffix Version | Value |  |  | Unit ${ }^{\circ} \mathrm{C}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min. <br> 0 | Typ. | Max. <br> 70 |  |
| $V_{D D}$ | Operating Supply Voltage | $\begin{aligned} & \hline \mathrm{f}_{\mathrm{CPU}}=8 \mathrm{MHz} \\ & \mathrm{f}_{\mathrm{CPU}}=4 \mathrm{MHz} \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline 4.5 \\ & 4.0 \end{aligned}$ |  | $\begin{aligned} & 5.5 \\ & 5.5 \\ & \hline \end{aligned}$ | V |
| ${ }^{\text {fosc }}$ | Oscillator Frequency | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=4.0 \mathrm{~V} \\ & \mathrm{~V}_{\mathrm{DD}}=4.5 \mathrm{~V} \end{aligned}$ | 0 |  | $\begin{aligned} & 12 \\ & 24 \end{aligned}$ | MHz |

Figure 35. Maximum Operating Frequency (Fmax) Versus Supply Voltage (KD)


Note: The shaded area is outside the recommended operating range; device functionality is not guaranteed under these conditions.

### 6.3 DC ELECTRICAL CHARACTERISTICS

( $\mathrm{T}_{\mathrm{A}}=0$ to $+70^{\circ} \mathrm{C}$ unless otherwise specified)

| Symbol | Parameter | Test Conditions | Value |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min. | Typ. | Max. |  |
| $\mathrm{V}_{\text {IL }}$ | Input Low Level Voltage All Input pins |  |  |  | $V_{D D} \times 0.3$ | V |
| $\mathrm{V}_{\mathrm{IH}}$ | Input High Level Voltage All Input pins |  | $V_{D D} \times 0.7$ |  |  | V |
| $\mathrm{V}_{\mathrm{HYS}}$ | Hysteresis Voltage ${ }^{1)}$ All Input pins | $V_{D D}=5 \mathrm{~V}$ | TBD |  |  | V |
| $\mathrm{V}_{\text {OL }}$ | Low Level Output Voltage All Output pins | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{I}_{\mathrm{OL}}=+10 \mu \mathrm{~A} \\ & \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{I}_{\mathrm{OL}}=+1.6 \mathrm{~mA} \\ & \hline \end{aligned}$ |  |  | $\begin{aligned} & \hline 0.1 \\ & 0.4 \end{aligned}$ | V |
|  | Low Level Output Voltage High Sink I/O pins | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{I}_{\mathrm{OL}}=+10 \mu \mathrm{~A} \\ & \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{I} ; \mathrm{I}_{\mathrm{OL}}=+1.6 \mathrm{~mA} \\ & \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{I}_{\mathrm{OL}}=+10 \mathrm{~mA} \end{aligned}$ |  |  | $\begin{aligned} & 0.1 \\ & 0.4 \\ & 1.5 \end{aligned}$ |  |
| $\mathrm{V}_{\mathrm{OH}}$ | High Level Output Voltage All Output pins | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{I}_{\mathrm{OL}}=-10 \mu \mathrm{~A} \\ & \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{I}_{\mathrm{OL}}=1.6 \mathrm{~mA} \end{aligned}$ | $\begin{gathered} 4.9 \\ 4 \end{gathered}$ |  |  | V |
| I/L$\mathrm{IIH}^{\text {a }}$ | Input Leakage Current All Input pins but RESET | $\mathrm{V}_{\text {IN }}=\mathrm{V}_{\text {SS }}$ (No Pull-Up configured) $\mathrm{V}_{\mathrm{IN}}=\mathrm{V}_{\mathrm{DD}}$ |  | 0.1 | 10 | $\mu \mathrm{A}$ |
|  | Input Leakage Current RESET pin | $\begin{aligned} & V_{I N}=V_{S S} \\ & V_{I N}=V_{D D} \end{aligned}$ |  | -50 | 10 |  |
| IDD | Supply Current in RUN Mode ${ }^{2)}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} \\ & \mathrm{f}_{\mathrm{OSC}}=12 \mathrm{MHz}, \mathrm{f}_{\mathrm{CPU}}=4 \mathrm{MHz} \\ & \mathrm{f}_{\mathrm{OSC}}=24 \mathrm{MHz}, \mathrm{f}_{\mathrm{CPU}}=8 \mathrm{MHz} \end{aligned}$ |  | $\begin{gathered} \text { TBD } \\ 14 \end{gathered}$ | $\begin{gathered} \text { TBD } \\ 18 \end{gathered}$ | mA |
|  | Supply Current in SLOW Mode ${ }^{3}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} \\ & \mathrm{f}_{\mathrm{OSC}}=12 \mathrm{MHz}, \mathrm{f}_{\mathrm{CPU}}=2 \mathrm{MHz} \\ & \mathrm{f}_{\mathrm{OSC}}=24 \mathrm{MHz}, \mathrm{f}_{\mathrm{CPU}}=4 \mathrm{MHz} \end{aligned}$ |  |  | TBD | mA |
|  | Supply Current in WAIT Mode ${ }^{3}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} \\ & \mathrm{f}_{\mathrm{OSC}}=12 \mathrm{MHz}, \mathrm{f}_{\mathrm{CPU}}=4 \mathrm{MHz} \\ & \mathrm{f}_{\mathrm{OSC}}=24 \mathrm{MHz}, \mathrm{f}_{\mathrm{CPU}}=8 \mathrm{MHz} \end{aligned}$ |  | $\begin{gathered} \text { TBD } \\ 12 \end{gathered}$ | $\begin{gathered} \text { TBD } \\ 18 \end{gathered}$ | mA |
|  | Supply Current in HALT Mode | $\begin{aligned} & I_{\text {LOAD }}=0 \mathrm{~mA} \\ & \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} \end{aligned}$ |  | 250 | 500 | $\mu \mathrm{A}$ |

## Notes:

1. Hysteresis voltage between switching levels
2. CPU running with memory access.
3. All peripherals in stand-by

### 6.4 A/D CONVERTER CHARACTERISTICS

( $\mathrm{T}_{\mathrm{A}}=0$ to $+70^{\circ} \mathrm{C}$ unless otherwise specified)

| Symbol | Parameter | Test Conditions | Value |  |  | Unit |
| :---: | :--- | :--- | :--- | :---: | :---: | :---: |
|  |  |  |  | Min. | Typ. |  |
|  |  |  |  |  |  |  |
| Res | Resolution |  |  | 8 |  | Bit |
| DLE <br> ILE | Differential linearity error <br> Integral linearity error | $\mathrm{f}_{\mathrm{OSC}}=24 \mathrm{MHz}$ |  | $\pm 0.3$ | $\pm 0.5$ | LSB |
| $\mathrm{t}_{\mathrm{C}}$ | Conversion Time | $\mathrm{f}_{\mathrm{CPU}}=8 \mathrm{MHz}$ |  |  | 8 |  |
| 1 |  |  |  |  |  |  |

Note: Noise at $\mathrm{AV}_{\mathrm{DD}}, \mathrm{AV}_{\mathrm{SS}}<10 \mathrm{mV}$

### 6.5 PWM (DAC) CHARACTERISTICS

| PWM/BRM Electrical and Timings |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
|  | Repetition rate | $\mathrm{T}_{\text {CPU }}=125 \mathrm{~ns}$ |  | 125 |  | kHz |
| Res | Resolution | $\mathrm{T}_{\text {CPU }}=125 \mathrm{~ns}$ |  | 125 |  | ns |
| S | Output step | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}, 10$ bits |  | 5 |  | mV |
|  |  | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}, 12$ bits |  | 1.25 |  | mV |
| $\mathrm{R}_{\mathrm{S}}$ | Serial resistor | - |  | 700 | 1000 | Ohms |

### 6.5.1 I2C CHARACTERISTICS

| I2C Electrical specifications |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Symbol | Unit | Standard mode I2C |  | Fast mode I2C |  |
|  |  |  | Min | Max | Min | Max |
| Hysteresis of Schmitt trigger inputs Fixed input levels $V_{D D}$-related input levels | $\mathrm{V}_{\mathrm{HYS}}$ | V | $\begin{aligned} & \mathrm{N} / \mathrm{A} \\ & \mathrm{~N} / \mathrm{A} \end{aligned}$ | $\begin{aligned} & \mathrm{N} / \mathrm{A} \\ & \mathrm{~N} / \mathrm{A} \end{aligned}$ | $\begin{aligned} & 0.2 \\ & 0.05 \mathrm{~V}_{\mathrm{DD}} \end{aligned}$ |  |
| Pulse width of spikes which must be suppressed by the input filter | $\mathrm{T}_{\text {SP }}$ | ns | N/A | N/A | 0 ns | 50 ns |
| Output fall time from VIH min to VIL max with a bus capacitance from 10 pF to 400 pF with up to 3 mA sink current at VOL1 with up to 6 mA sink current at VOL2 | $\mathrm{T}_{\text {OF }}$ | ns | N/A | $\begin{aligned} & 250 \\ & \text { N/A } \end{aligned}$ | $\begin{aligned} & 20+0.1 \mathrm{Cb} \\ & 20+0.1 \mathrm{Cb} \end{aligned}$ | $\begin{aligned} & 250 \\ & 250 \end{aligned}$ |
| Input current each I/O pin with an input voltage between 0.4 V and $0.9 \mathrm{~V}_{\mathrm{DD}} \max$ | I | $\mu \mathrm{A}$ | - 10 | 10 | -10 | 10 |
| Capacitance for each I/O pin | C | pF |  | 10 |  | 10 |

N/A = Not Applicable
$\mathrm{Cb}=$ Capacitance of one bus in pF

| I2C Bus Timings |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Standard I2C |  | Fast I2C |  | Symbol | Unit |
| Bus free time between a STOP and START condition | $\begin{gathered} \text { Min } \\ 4.7 \end{gathered}$ | Max | $\begin{aligned} & \text { Min } \\ & 1.3 \end{aligned}$ | Max | $\mathrm{T}_{\text {BUF }}$ | ms |
| Hold time START condition. After this period, the first clock pulse is generated | 4.0 |  | 0.6 |  | THD:STA | $\mu \mathrm{s}$ |
| LOW period of the SCL clock | 4.7 |  | 1.3 |  | TLOW | $\mu \mathrm{s}$ |
| HIGH period of the SCL clock | 4.0 |  | 0.6 |  | $\mathrm{T}_{\text {HIGH }}$ | $\mu \mathrm{s}$ |
| Set-up time for a repeated START condition | 4.7 |  | 0.6 |  | TSU:STA | $\mu \mathrm{s}$ |
| Data hold time | $0{ }^{1)}$ |  | $0{ }^{17}$ | $0.9{ }^{2)}$ | THD:DAT | ns |
| Data set-up time | 250 |  | 100 |  | TSU:DAT | ns |
| Rise time of both SDA and SCL signals |  | 1000 | 20+0.1 Cb | 300 | TR | ns |
| Fall time of both SDA and SCL signals |  | 300 | 20+0.1-6b | 300 | TF | ns |
| Set-up time for STOP condition | 4.0 |  | 0.6 |  | TSU:STO | ns |
| Capacitive load for each bus line |  | 400 |  | 400 | Cb | pF |

## Notes:

1. The device must internally provide a hold time of at least 300 ns for the SDA signal in order to bridge the undefined region of the falling edge of SCL.
2. The maximum hold time of the START condition has only to be met if the interface does not stretch the low period of SCL signal.

## 7 GENERAL INFORMATION

### 7.1 EPROM ERASURE

EPROM version devices are erased by exposure to high intensity UV light admitted through the transparent window. This exposure discharges the floating gate to its initial state through induced photo current.
It is recommended that the EPROM devices be kept out of direct sunlight, since the UV content of sunlight can be sufficient to cause functional failure. Extended exposure to room level fluorescent lighting may also cause erasure.

An opaque coating (paint, tape, label, etc...) should be placed over the package window if the product is to be operated under these lighting conditions. Covering the window also reduces bd in power-saving modes due to photo-diode leakage currents.
An Ultraviolet source of wave length $2537 \AA$ yielding a total integrated dosage of $15 \mathrm{Watt}-\mathrm{sec} / \mathrm{cm}^{2}$ is required to erase the device. It will be erased in 15 to 20 minutes if such a UV lamp with a $12 \mathrm{~mW} / \mathrm{cm}^{2}$ power rating is placed 1 inch from the device window without any interposed filters.

### 7.2 PACKAGE MECHANICAL DATA

Figure 36. 42-Pin Shrink Plastic Dual In-Line Package, 600-mil Width


Figure 37. 42-Pin Shrink Ceramic Dual In-Line Package, 600-mil Width


Figure 38. 56-Pin Shrink Plastic Dual In Line Package, 600-mil Width

|  | Dim. | mm |  |  | inches |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Typ | Max | Min | Typ | Max |
|  | A |  |  | 6.35 |  |  | 0.250 |
| $\pm$ | A1 | 0.38 |  |  | 0.015 |  |  |
|  | A2 | 3.18 |  | 4.95 | 0.125 |  | 0.195 |
|  | b |  | 0.41 |  |  | 0.016 |  |
|  | b2 |  | 0.89 |  |  | 0.035 |  |
|  | C | 0.20 |  | 0.38 | 0.008 |  | 0.015 |
| ¢¢¢ | D | 50.29 |  | 53.21 | 1.980 |  | 2.095 |
| P | E |  | 15.01 |  |  | 0.591 |  |
|  | E1 | 12.32 |  | 14.73 | 0.485 |  | 0.580 |
|  | e |  | 1.78 |  |  | 0.070 |  |
| Lemp midic w | eA |  | 15.24 |  |  | 0.600 |  |
|  | eB |  |  | 17.78 |  |  | 0.700 |
|  | L | 2.92 |  | 5.08 | 0.115 |  | 0.200 |
|  |  | Number of Pins |  |  |  |  |  |
|  | N | 56 |  |  |  |  |  |

Figure 39. 56-Pin Shrink Ceramic Dual In-Line Package, 600-mil Width


Figure 40.64-Pin Thin Quad Flat Package


### 7.3 ORDERING INFORMATION

Each device is available for production in user programmable version (OTP) as well as in factory coded version (ROM). OTP devices are shipped to customer with a default blank content FFh, while ROM factory coded parts contain the code sent by customer. There is one common EPROM version for debugging and prototyping which features the maximum memory size and peripherals of the subfamily. Care must be taken to only use resources available on the target device.
Contact sales office for further ordering information and availablity.

### 7.3.1 Transfer Of Customer Code

Customer code is made up of the ROM contents and the list of the selected options (if any). The ROM contents are to be sent on diskette, or by electronic means, with the hexadecimal file generated by the development tool. All unused bytes must be set to FFh.

The selected options are communicated to SGS-THOMSON using the correctly completed OPTION LIST appended.
The SGS-THOMSON Sales Organization will be pleased to provide detailed information on contractual points.

Figure 41. ROM Factory Coded Device Types


Figure 42. OTP User Programmable Device Types


Note: The ST72E37J4D0 (42-pin ceramic SDIP) and ST72E671N4D0 (56-pin ceramic SDIP) are used as the EPROM versions for the above devices. The EPROM devices are tested for operation at $25^{\circ} \mathrm{C}$ only.

## ST72371/ST72372 MICROCONTROLLER OPTION LIST

Customer
Address

Contact
Phone No
Reference

SGS-THOMSON Microelectronics references

| Device: | [ ] ST72371/ST72372 |
| :--- | :--- | :--- |
| Package: | [ ] Dual in Line Plastic [ ] Thin Quad Flat Pack: |
| [ ] Standard (Stick) |  |

Comments:
Supply Operating Range in the application:
Oscillator Frequency in the application:
Notes
Signature
Date

## Notes

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[^0]:    This is preliminary information on a new product in development orundergoing evaluation. Detail sare subject to change without notice.

[^1]:    Stack Higher Address $=01$ FFh
    Stack Lower Address $=0100 \mathrm{~h}$

